## TECHNICAL MANUAL

COMBINED<br>OPERATIONS AND MAINTENANCE INSTRUCTIONS<br>WITH<br>PARTS BREAKDOWN<br>(ORGANIZATIONAL AND INTERMEDIATE)<br>RADIO RECEIVER<br>TYPE<br>R-2174(P)/URR<br>AND<br>R-2174A(P)/URR

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## CHANGE 2 A

## WARNING HIGH VOLTAGE

is used in the operation of this equipment
DEATH ON CONTACT
may result if personnel fail to observe safety precautions.
Learn the area containing the high voltage within the equipment.
Be careful not to contact high voltage connections when installing or operating this equipment.
Before working inside the equipment, turn power off and ground points of high potential before touching them.

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## INTRODUCTION

This technical manual provides operation and maintenance instructions for the R-2174(P)/URR and R-2174A(P)/URR Radio Receiver, along with an Illustrated Parts Breakdown and Schematic Diagrams. The manual is prepared in accordance with specifications MIL-M-38798B and MIL-M-38807 and other specifications and publications as called for by these two specifications. The Operation, Maintenance and Schematic Diagrams were prepared in accordance with MIL-M38798B while the Illustrated Parts Breakdown was prepared in accordance with MIL-M-38807. The manual consists of eight chapters along with a cross reference index and alphabetical index.

The original R-2174 contains 11 major modules numbered A1 through A10 with the A5 and A6A1 being optional. The edgelighting configuration in the R-2174 utilizes incandescent bulbs. Also in this model, two configurations for the A7 board exist.

The improved R-2174A version contains the same 11 modules, however the A2 (First Mixer Assembly) and A8 (Second LO/BFO Synthesizer Assembly) boards have been redesigned and assigned new part numbers. Additionally, the edgelighting configuration now includes the use of LEDs. Changes in the R-2174A version are directly interchangeable with the original R-2174 version. This manual includes both the R-2174 and the R-2174A versions.

Chapter 1 presents general information about the Receiver, which includes its physical and operating characteristics. Chapter 2 contains two sections with the first section providing information on installation logistics. Section 11 provides procedures for installing the Receiver and for checking its performance. Preparation for use and reshipment is presented in Chapter 3. This information includes procedures for unpacking the equipment and preparing it for installation. Procedures are also included in this chapter for repacking the unit and preparing it for storage or reshipment. Chapter 4 contains information on operation of the Receiver. This information is designed to acquaint the technician with the unit, and to describe its controls indicators and operating characteristics. Step by step procedures are also provided for operating the Receiver in various modes.

The theory of operation is presented in two ways in Chapter 5. The first method, contained in Section I, describes the system function while the second method provides a detailed circuit analysis and is contained in Section II. Chapter 6 provides information for all functions of maintenance to be performed at organizational and intermediate levels. An Illustrated Parts Breakdown is provided in Chapter 7 which includes a listing of all component parts, contained in the Receiver. Chapter 8 contains schematic diagrams, for all the electronic circuits of the Receiver. An alphabetical index as well as a cross reference index is included at the end of the manual to provide the user with a convenient method for locating information.

## CHANGE 2 xi



Figure 1-1. Radio Receiver R-2174 (P)/URR, Overall View

## CHAPTER 1 GENERAL INFORMATION

1-1. INTRODUCTION. Chapter I provides general information on the R-2174(P)/URR Radio Receiver. This information describes the equipment, gives its purpose, leading particulars, capabilities and limitations. Figure 1-1 shows an overall view of the R-2174(P)/URR Radio Receiver. A list of the equipment supplied, test equipment and technical manuals related to the Receiver are also provided in this chapter.

1-2. DESCRIPTION AND PURPOSE. The R-2174(P)/URR Radio Receiver is designed to receive and demodulate HF signals within frequency range from 0.5 to 29.999 MHz with reception capabilities for AM (A3), FM (F3), CW (AI), LSB/USB (A3J) and optionally ISB (A3B). Reception mode codes are defined in Table 1-1. The unit is a fully synthesized, solid state, microcomputer controlled Receiver that can be operated from its front panel or optionally from a remote location. The Receiver can be tuned over its frequency range by selecting incremental digits from the tens MHz to 1 Hz . Convenient front panel control of most receiver functions is achieved through pushbutton keypads and an easy spin tuning knob with readouts presented on two liquid crystals displays.
a. Selectivity. Receiver bandwidth is selectable through plug-in filters. Any number of filters, from 1 to 7 , of various bandwidths may be installed in the Receiver with selection of a particular bandwidth (filter) made through Receiver control. Fifteen different filters are available that cover bandpass from 400 Hz to 16 kHz including two sideband filters for upper, lower and optional independent sideband modes. The optional available filters are listed in Table 1-4. Additional special filters may be ordered from the factory.
b. Receiver outputs. In addition to a front panel audio PHONES jack for audio monitoring, audio, IF, reference frequencies, diversity AGC and fault indicating outputs are available at the rear panel.
(1) Audio. The front panel PHONES jack provides a nominal 10 milliwatts into 600 ohms and is adjustable with the front panel AF GAIN control. A nominal 1 milliwatt, 600 ohm balanced line and a nominal 1 Watt, 8 ohm line is output to the rear panel. When the receiver is equipped with the optional ISB circuit card, two additional 1 milliwatt, 600 ohm balanced lines are output to the rear panel for sideband audio.
(2) IF Output. A 455 kHz IF output of -10 dBm nominal into 50 ohms is available through a separate connector on the rear panel.
(3) Reference Frequencies. When the receiver is operated in the internal mode (using the internal reference oscillator), one of three selectable reference frequencies ( $1 \mathrm{MHz}, 5 \mathrm{MHz}$ or 10 MHz ) of (approximately) 0 dBm into 50 ohms is available through a separate connector on the rear panel.
(4) Diversity AGC. Automatic gain control outputs from the MAIN IF-AGC circuits, and when installed, from the ISB-AGC circuits are routed to the rear panel for diversity AGC.
(5) Fault Indication Output. In addition to a fault indicator on the front panel a fault indicating circuit that provides a TTL level, related to a fault condition, is output to the rear panel.
c. Receiver Control Capabilities. The Receiver may be operated from its front panel or, with an optional plug-in circuit card, it may be operated from a remote location.
(1) Front Panel (local) Control. Two sets of keypads provide for receiver tuning, mode selection, AGC control, bandwidth selection, RF/AF metering and Built In Test Equipment (BITE) functions. A tuning knob is used to tune BFO offset frequency or may be used as an alternate means for receiver tuning. An IF GAIN control can be used to set AGC threshold levels or may be disabled with fully automatic AGC. An AF GAIN control and two line level controls are used to control audio output levels. Two liquid crystal displays monitor and display receiver functions.
(2) Remote Control. With the optional A6A1 circuit card installed, the Receiver can be controlled from a remote location using serial asynchronous (ASCII-7 bit) character oriented data at one of 16 selectable baud rates from 50 baud to 19.2 kilobaud. Receiver parameters that can be controlled through the remote option include; receiver frequency, BFO, bandwidth, detection mode, AGC, RF/IF manual gain, system status, fault status and BITE control. When remote equipment is in place and operating there is both remote and front panel indication of receiver status when controlled either remotely or locally, with remote indications being sent upon request from the remote controller.
d. Built in Test Equipment (BITE). The Receiver contains a built in test system that can find and isolate faults to a circuit card. The BITE, when initiated locally or remotely runs a self test sequence on various receiver parameters and provides a coded fault readout when faults are detected. The coded readout can then be used to determine the nature of the fault and its location.

1-3. FUNCTIONAL DESCRIPTION. Figure 1-2 shows a simplified block diagram of the Receiver which will aid in understanding this functional description. The description is divided into six functions that best describe the functions of the Receiver: primary signal flow, oscillator synthesizers, automatic gain control, receiver control, built in test equipment and receiver power.
a. Primary Signal Flow. Input signals from an antenna are connected directly to a modular RF lowpass filter which rejects signals above the 30 MHz upper range of the Receiver. The RF output of the filter is routed to the first mixer circuit card and mixed with the synthesized first local oscillator signal. The oscillator is variable between 40.955 and 70.454999 MHz and when mixed with the RF signal yields a difference frequency of 40.455 MHz . This first IF signal is coupled through a bandpass filter, that provides a 20 kHz bandpass, to an AGC controlled amplifier. This AGC level controlled IF signal is then routed to the second mixer circuit card. This circuit card provides additional AGC control and bandpass filtering before coupling the signal to the second mixer. The second mixer with a 40 MHz input from the second local oscillator along with the 40.455 MHz first IF input yields the difference frequency of 455 kHz which is used as the second IF. The second IF signal is filtered and coupled to an IF amplifier to restore gain and for coupling to the plug-in bandpass filters on the main IF/AF circuit card. Any number (up to seven) filters may be plugged into the filter slots which are then selected through receiver control to provide selected bandwidth to the IF signal. The output of filter slot FL1 can be linked to the optional ISB circuit card or linked to the common selected output of the six other filter slots. This output is coupled to an AGC controlled IF amplifier which provides gain and filtering to the IF signal with a selected bandwidth. A portion of this signal is routed through an IF output amplifier to the rear panel. The IF signal is also routed to two detectors along with the BFO oscillator frequency. In CW and sideband modes the BFO, through receiver control, is routed through a limiting amplifier for carrier reinsertion at the product detector. In the AM mode the AM carrier is routed through the same path. In the FM mode the carrier is removed by the FM detector. Receiver control selects the output from the appropriate detector and routes it through an audio lowpass filter to a crosspoint switch. This switch, through receiver control, directs this audio along with ISB audio, when present, through level controls to audio output amplifiers. In non ISB modes three outputs are available; the front panel PHONES jack and a loud speaker output at the rear panel, both controlled by the AF GAIN control and a balanced 600 ohm 1 milliwatt MAIN LINE
rear panel output controlled by the MAIN LINE LEVEL control. In the ISB mode two additional balanced 600 ohm 1 milliwatt lines are output to the rear panel. The outputs contain the USB (line 1) and the LSB (line 2) and are controlled by the MAIN and I-LSB LINE LEVEL controls respectively. The MAIN LINE, PHONES and loudspeaker outputs may contain either the USB or LSB as selected through receiver control. In this ISB mode the output of IF bandpass filter FLI is linked to the input of the optional ISB circuit card and processed in the same manner as the main IF except the BFO is routed through an amplifier directly to the product and synchronous detector and the audio output routed through an emitter follower to the crosspoint switch on the MAIN IF/AF circuit card. This signal is then processed as described above.
b. Oscillator Synthesizers. The Receiver contains a first local oscillator synthesizer for frequency input to the first mixer, a second local oscillator synthesizer for frequency input to the second mixer and a beat frequency oscillator synthesizer for carrier reinsertion to the product detectors in CW and sideband modes.
(1) First Local Oscillator Synthesizer. The first local oscillator output, variable between 40.955 and 70.454999 MHz , is used as the input to the first mixer. Its frequency, controlled through receiver control, determines reception frequency when mixed with the RF signal, to always yield. a difference frequency of 40.455 MHz as the first IF signal. The voltage controlled oscillators frequency is adjustable through a phase lock loop which is controlled through conversion of digital inputs from receiver control.
(2) Second Local Oscillator Synthesizer. The second local oscillators output, fixed at 40 MHz , is used as the input to the second mixer. This signal is mixed with the first IF signal of 40.455 MHz and yields a difference frequency of 455 kHz which forms the second IF signal. The voltage controlled oscillator operates at 20 MHz through a phase lock loop which keeps the oscillator locked on frequency. The 20 MHz output is coupled through a frequency doubler circuit to provide 40 MHz to the second mixer. The output of this oscillator is also used to provide reference frequencies, through divide by circuits, to the first local oscillator, ( 1 MHz ) the BFO ( 1 MHz ) and to the rear panel of the Receiver ( 1,5 , or 10 MHz ). The second local oscillator receives its reference frequency either from the internal reference ( 5 MHz ) or from an external reference.
(3) Beat Frequency Oscillator. The beat frequency oscillator (BFO) output, variable between 447 and 463 kHz , is used as carrier reinsertion at the product detectors for CW and sideband operation. The oscillator operates at a center frequency of $22.75 \mathrm{MHz}, 50$ times above the BFO output frequency. The oscillator is variable through a phase lock loop that is controlled through conversion of digital inputs from receiver control. The output of the oscillator is routed through a divide by 50 circuit to the product detectors on the main IF/AF circuit card and to the optional ISB circuit card when installed.
c. Automatic Gain Control. The Receiver contains Automatic Gain Control (AGC) circuitry for maintaining audio and IF output levels with large variations in the incoming RF signals. The AGC operates in one of three selectable modes; automatic, manual, or automatic with a manually set threshold. The AGC receives a portion of the second IF signal, detects it, provides three different hang times (long, medium or short) provides for automatic or manual threshold control, filters and distributes the gain control signal to both the first and second IF amplifiers. Two similar AGC circuits, just described, are contained; one on the MAIN IF/AF circuit card, the second on the optional ISB circuit card, but with their outputs integrated with each other. In non ISB modes the system operates as described above through the MAIN IF-AGC circuits. In ISB modes, the stronger of the two signals, from upper and lower sidebands, control the first IF amplifiers while the second IF amplifiers are controlled independently with their respective AGC circuits.
d. Receiver Control. The Receiver is controlled through a pre-programmed microcomputer contained on a circuit card within the Receiver. Receiver control instructions, are routed through
computer controlled interface circuits to the microcomputer. These receiver control instructions are processed by the microcomputer which in turn sends instructions through interface circuits to the appropriate receiver function to be controlled and to indicating circuits. Front panel (local) instructions are routed through the receiver control circuit card while remote instructions are routed through an optional interface circuit card. In either remote or local mode the microcomputer instructions are routed through the receiver control circuit card to receiver functions and to two liquid crystal displays on the front panel. Receiver status formation is stored in the microcomputer and can be routed to a remote controller, on request through the optional interface circuit card. The microcomputer processed instructions control; receiver frequency (first local oscillator), BFO (BFO oscillator), bandwidth (filter selection), AGC mode and level (AGC circuits) receiver mode (detector selection) audio output (MAIN/ISB selection), built in test equipment (BITE), and readouts for the above functions along with AF/RF metering.
e. Built In Test Equipment. The built in test equipment (BITE) is a system of microcomputer programs that reads receiver parameters to determine performance of various circuits. The BITE sequence controls receiver functions to read parameters under various modes, frequencies, bandwidths, etc. The tests performed by BITE include; readability of the microcomputer RAMs (random access memory), lock condition and settling time of all three frequency synthesizers, locating and measuring the bandwidth of installed IF bandpass filters, AGC operation including ISB AGC (if installed), product detector operation and metering circuit operation.
f. Receiver Power. The receiver contains a modular power supply which can be operated from 100, 120, 220 or 240 Volt, 48 to 420 Hz , single phase power. The power supply converts the ac input into six different dc voltages for distribution throughout the receiver. Four of the dc voltages $(+5,+15,-15$ and +20 ) are regulated while two ( +5 and +15 ) are non regulated.

## TABLE 1-1. RECEPTION MODE CODES

Complete Designation Example:


1-4. MECHANICAL DESCRIPTION. A rigid, die-cast, full width chassis is used as the base for the mainframe of the Receiver. Mounted within compartments on the underside of this chassis are the mixer boards and the frequency generation system. The input lowpass filter, (A1), main IF/AF, (A4), optional ISB IF/AF (A5) and power supply modules (A 10) are located on the top surface of the die-cast chassis while the control (A9) and digital I/O modules (A6A1 and A6A2) are attached to the Receiver mainframe. All modules are accessible for maintenance and can be removed or replaced using simple hand tools without the use of a soldering iron.


Figure 1-2. Radio Receiver R-2174(P)/URR, Simplified Block Diagram
a. Manual controls and indicators for operation of the Receiver are contained on the front panel while input/output jacks and connectors are provided on the rear panel. A PHONES jack, for audio connection to optional headphones, is contained on the front panel for convenient access. A primary power fuse is accessible from the rear panel.

1-5. LEADING PARTICULARS. Table 1-2 lists the leading particulars for the Receiver. These particulars include official nomenclature, common name, physical descriptions, storage and cabling requirements.

1-6. CAPABILITIES AND LIMITATIONS. Table 1-3 describes functional characteristics for the Receiver along with individual component characteristics and ambient conditions for operation. The left column of the table lists the Receiver function while the right column describes the capability or limitation.

1-7. EQUIPMENT SUPPLIED. Table 14 lists the equipment supplied with the Receiver. Equipment shown as optional may or may not be supplied with the Receiver.

1-8. EQUIPMENT REQUIRED BUT NOT SUPPLIED. All equipment required for normal installation and operation of the Receiver is supplied and listed ir Table 1-4. No other equipment is required to place the unit in operation.

1-9. OPTIONAL ITEMS. Optional filters, assemblies, and other items that are available for use with the Receiver are listed in Table 1-5.

1-10. TEST EQUIPMENT. Test equipment required for intermediate level maintenance is listed in Table 1-6. No other tools or test equipments are required for organizational level maintenance.

1-11. RELATED TECHNICAL MANUALS. A list of the technical manuals related to the Receiver are presented in Table
1-7. The publication number, title, and equipment nomenclature for each manual is shown.
1-12. MNEMONICS. Table 1-8 lists mnemonics and acronyms used throughout this manual.
TABLE 1-2. LEADING PARTICULARS

| Leading Particular | Description |
| :--- | :--- |
| Nomenclature | Receiver, Radio R-2174(P)/URR. |
| Common Name | Receiver |
| Type | Superheterodyne HF, Solid State, RF and Logic Modular Circuitry. <br> Modular circuit card assemblies, packaged in an aluminum frame using <br> a die-cast internal support and housing. |
| Package | $100,120,220$ or $240 \pm 10 \%$ Volts and 43 to 420 Hz at 50 Watts (nominal) <br> consumption. |
| Requirements Power | Printed circuit card switch for selection of one of 4 above listed volt- <br> ages. |
| Selection | Chassis size: Height, 5.1 inches high; 16.6 inches wide; 18.4 inches deep. <br> Fhysical Size <br> handles: 20.0 inches. <br> 30 Pounds. |
| Weight |  |

TABLE 1-2. LEADING PARTICULARS (Cont.)

| Leading Particular | Description |
| :--- | :--- |
| Mounting | Designed for standard rack mounting with two front panel handles for ease <br> and safety in handling during installation or removal. <br> Storage <br> Requirements |
| May be stored in ambient $-40^{\circ}$ to $+70^{\circ} \mathrm{C}$ with humidity from 10 to 95 <br> percent non-condensing. |  |
| Cabling | See Chapter 2, Installation. |

TABLE 1-3. CAPABILITIES AND LIMITATIONS

| Receiver Function | Capability or Limitation |
| :---: | :---: |
| Frequency Range | 500 kHz to 29.999999 MHz . |
| Frequency Selection | 1 Hz increment. |
| Frequency Tuning | By keyboard entry or continuous tuning with selectable rate, FAST ( 1 kHz increments), SLOW ( 30 Hz increments) and FINE ( 1 Hz increments). BFO continuous in 10 Hz increments. |
| Frequency Indication | (a) 8 digit electronic readout of tuned frequency to 1 Hz . <br> (b) 3 digit and sign readout of BFO relative to IF center $\pm 8 \mathrm{kHz}$ |
| Frequency Stability | The stability of the internal frequency standard shall be at least $\pm 5$ parts in $10^{\circ}$ per day for any $10^{\circ} \mathrm{C}$ increment within the temperature range $0^{\circ} \mathrm{C}$ to $50^{\circ} \mathrm{C}$. |
| Modes of Operation | USB/A3J Upper Side Band; LSB/A3J Lower Side Band; ISB/A3B Independent Side Band (optional); CW/AI Continuous Wave; AM/A3 Amplitude Modulation; FM F3 Frequency Modulation. |
| Input Impedance | 50 Ohms nominal, 2.1 VSWR; type N connector, and unbalanced input |
| AGC Range | The change in IF or AF output level for a change in input level of -100 dBm to -10 dBm is less than 6 dB . |
| AGC Time Constants | The attack time of the AGC system is less than 15 msecs. The decay time for the three selectable time constants are: |
| Image/Spurious Rejection | $80 \mathrm{~dB}, 25 \mathrm{kHz}$ or more removed from tuned frequency. |
| Internal Spurious Responses | Not greater than -123 dBm except for one spurious in each MHz which may be as high as -116 dBm . |
| Sensitivity $\begin{array}{r} 500 \mathrm{kHz}-29.999 \\ \mathrm{MHz} \end{array}$ | Single Sideband: -113 dBm (.5uv) for $10 \mathrm{db}(\mathrm{S}+\mathrm{N}) / \mathrm{N}$ ratio in a 3.24 kHz bandwidth. <br> AM: -99 dBm (2.5 uv) for $10 \mathrm{db}(\mathrm{S}+\mathrm{N}) / \mathrm{N}$ ratio in a 6 kHz bandwidth. |

TABLE 1-3. CAPABILITIES AND LIMITATIONS (Cont.)

| Receiver Function | Capability or Limitation |
| :---: | :---: |
| Reciprocal Mixing | The apparent noise at the input of the Receiver with a 3.24 kHz IF bandwidth will be less than - 13 dBm when the Receiver is subjected to the following interfering signal conditions. <br> (a) $\pm 15 \mathrm{kHz}$ offset from the Receiver tuned frequency at a level of -23 dBm. <br> (b) $\pm 500 \mathrm{kHz}$ offset from the Receiver tuned frequency at a level of $\pm 2 \mathrm{dBm}$. |
| Noise Figure | 16 dB from 0.5 MHz to 1.499999 MHz ; 13 dB from 1.5 MHz to 29.99999 MHz . |
| Intermodulation (In Band) | Better than -50 dB for two -36 dBm input signals when measured at the IF or line AF output. |
| Intermodulation | Third order input intercept point greater than +20 dBm . Second order input intercept point greater than +60 dBm . |
| Cross Modulation | The cross modulation appearing on a - 45 dBm unmodulated tuned signal due to a $30 \%$ AM signal which is no closer to the tuned signal than $\pm 100$ kHz will be less than $10 \%$ if the level of the interfering signal is +15 dBm . |
| Test Tone to Total Distortion (TTDR) Ratio | The receiver TTDR is greater than 33 dB for an AM signal modulated at 1 kHz to a depth of $75 \%$, or any level between -83 dBm and -10 dBm , when a 6.8 kHz IF bandwidth (optional Special Filter No. 4) is used. |
| Outputs | (a) Phone output to front panel. 10 mW nominal into 600 Ohm load. <br> $3 \%$ distortion. AM signal modulated $30 \%$ at lkHz . <br> (b) Line output at rear panel. 1 mW nominal at 600 Ohms. $2 \%$ distortion at 1 mW . Adjustable from front panel. AM signal modulated $30 \%$ at 1 kHz . <br> (c) Loudspeaker output at rear panel. I Watt nominal at 8 Ohms. 3\% distortion. AM signal modulated $30 \%$ at l kHz . <br> (d) 455 kHz IF output at rear panel. BNC connector. -10 dBm nominal at 50 ohms. |
| Rear Panel Connectors | Antenna Input Connector (N). IF Output Connector (BNC). REF Input/ Output Connector (BNC). Power Input Connector. Ground Terminal. Digital I/O Connector. AF/AGC outputs (D Connector). |
| Remote Control (Option) | Full remote control of the following receiver parameters using serial asynchronous, (ASCII-7 bit) character oriented data at one of 16 selectable baud rates from 50 baud to 19.2 kilobaud. <br> Tuned Frequency, BFO Tuning - BFO ON/OFF, IF Bandwidth, Detection Mode, AGC Time Constants - AGC ON/OFF, RF/IF Gain, FAULT Status, BITE Control, system status. |
| Status Indication | Front panel indication of status under remote control, remote indication of status under local or remote control. |
| RFI/EMI | Meets specification MIL-STD-46 1. |

TABLE 1-4. EQUIPMENT SUPPLIED

| Quantity <br> Per Unit | Nomenclature |  |
| :--- | :--- | :--- |
| 1 | R-2174(P)/URR | Description |
| 1 | W18 | Cable W18, Primary Power Input |
| 1 | P3 | Mating Connector for J3 |

TABLE 1-5. OPTIONAL ITEMS

| Nomenclature | Description |  |  |
| :---: | :---: | :---: | :---: |
| ISB Circuit | Printed circuit board for lower sideband signal path. (Optional). <br> Printed circuit board for remote control of the Receiver. Supplied with mating connector (Optional). |  |  |
| Board A5 |  |  |  |
| Interface Circuit |  |  |  |
| Board A6A1 |  |  |  |
|  |  |  |  |
| 0.4 kHz B.P | Filter \#1 | 454800 to 455200 Hz Band Pass Filter | General Purpose/1 |
| $1.2 \mathrm{kHz} \mathrm{B.P}$. | 2 | 454400 to 455600 Hz Band Pass Filter $\}$ |  |
| 6.8 kHz B.P. | 3 | 451600 to 458400 Hz Band Pass Filter |  |
| 3.24 kHz B.P. | 4 | 455350 to 458050 Hz Band Pass Filter | General PurposeOffset Sideband/ 1 |
| $3.24 \mathrm{kHz} \mathrm{B.P}$. | 5 | 451950 to 454650 Hz Band Pass Filter |  |
| $0.4 \mathrm{kHz} \mathrm{B.P}$. | 6 | 454800 to 455200 Hz Band Pass Filter | Dind |
| 1.2 kHz B.P. | 7 | 454400 to 455600 Hz Band Pass Filter |  |
| 3.24 kHz B.P. | 8 | 453380 to 456620 Hz Band Pass Filter > | Defined Delay/2 |
| 6.8 kHz B.P. | 9 | 451600 to 458400 Hz Band Pass Filter |  |
| 16.0 kHz B.P. | 10 | 447000 to 463000 Hz Band Pass Filter |  |
| 0.5 kHz B.P. | 11 | 454750 to 455250 Hz Band Pass Filter |  |
| 1.0 kHz B.P. | 12 | 454500 to 455500 Hz Band Pass Filter | Linear Phase/3 |
| $2.0 \mathrm{kHz} \mathrm{B.P}$. | 13 | 454000 to 456000 Hz Band Pass Filter |  |
| 3.0 kHz B.P. | 14 | 453500 to 456500 Hz Band Pass Filter |  |
| 6.0 kHz B.P. | 15 | 452000 to 458000 Hz Band Pass Filter |  |

*All filters are optionally procured, so that some filters listed may not be available as stock items.
NOTE: Some filters may be assigned as fixed items to the Receiver, otherwise seven filter slots are available for plug-in filters to the A4 Module.

## General Purpose Filters

General purpose filters (filters \#1 thru \#5) provide coverage of reception modes A1, A2, A3, A3H, A3J and A3B (with ISM module A5, installed).

This group of filters provides the specified optional standard intermediate frequencies (IF) as follows:

| Filter <br> Number | 3 dB BW <br> Hz Min | 60 dB BW <br> Hz Max |
| :--- | :--- | :--- |
|  |  |  |
| 1 | 400 | 2480 |
| 2 | 1200 | 7200 |
| 3 | 6800 | 22440 |

The general purpose offset sideband filters conform to the requirements of MIL-STD-188C Figure 18.

| Filter |  | Frequency Corners of Template Referred to AF |  |  |
| :--- | :---: | :---: | :---: | :---: |
| Number | Lower 3 dB | Upper 2 dB | Lower 60 db | Upper 60 db |
|  | -350 | -3050 | +300 | -4000 |
| 4 (LSB) | -350 | +3050 | -300 | +4000 |

For both filters, envelope delay distortion is maintained within 800 usecs between the frequencies 750 to 2750 Hz .

## NOTE

It is essential to install filter \#4 in slot FLI .
/2 Defined Delay Filters
This group of filters form part of the specified optional special IF bandwidths. Some similarity exists between the bandwidths and those in the general purpose group, but shape factors are much lower, and the envelope delay is now specified. All are symmetrical about the intermediate frequency.

Defined delay filters are described below:

| Filter <br> Number | 1 dB BW <br> Hz Min | 3 dB BW <br> Hz Min | 60 dB Bw <br> Hz Max | Envelope Delay <br> usecs over Hz BW |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| 6 |  | 400 | 1000 | 2000 | 300 |
| 7 | - | 1200 | 2400 | 900 | 1200 |
| 8 | 3100 | 3240 | 4300 | 1000 | 2800 |
| 9 | 6400 | 6800 | 13600 | 1000 | 6400 |
| 10 | 12000 | 1600 | 32000 | 60 | 12000 |

## Change 1 1-10

General purpose coverage of reception modes A1, A2, A3, A3H, A3J is provided with added capability of F4, F1 and F4 emissions.

Filter \#8 can be used for the reception SSB signals. If it is plugged into the FL1 slot, the receiver BITE routine will recognize it as a "symmetrical sideband" filter, and when LSB or USB are selected, the adjustment to the first and last local oscillators will be made automatically.

Linear Phase Filters
Filters 11 thru 15 complete the range of specified optional special IF bandwidths. All have minimum delay variation over their specified bandwidths.

| Filter <br> Number | 3 dB BW <br> Hz Min | 60 dB BW <br> Hz Max | Envelope Delay <br> usecs over Hz BW |  |
| :--- | :--- | :--- | :--- | :--- |
| 11 | 500 | 3000 | 80 | 500 |
| 12 | 1000 | 6000 | 40 | 1000 |
| 13 | 2000 | 12000 | 20 | 2000 |
| 14 | 3000 | 18000 | 20 | 3000 |
| 15 | 6000 | 36000 | 20 | 6000 |

These filters become essential where preservation of pulse slopes are important. The bandwidth is dependent upon the pulse rate and rise time.

TABLE 1-6. TEST EQUIPMENT LIST (INTERMEDIATE LEVEL MAINTENANCE ONLY)

| $\begin{aligned} & \hline \text { TYPE } \\ & \text { DESIGNATION } \end{aligned}$ | MFR CODE OR NAME/ADDRESS | FIGURE AND INDEX NO. | NOMENCLATURE | USE |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 4960PT3T | 28480 |  | Spectrum Analyzer, | Para. 6-9 | Intermediate |
| 010-6105-03 | 80009 |  | X10 Oscilloscope Probe | $\begin{array}{\|c\|} \hline \text { Para. 6-9 } \\ \hline \text { Table 6-3 } \\ \hline \end{array}$ | Intermediate |
| 1121A | 28480 |  | HI-Impedance Probe, Spectrum | $\begin{array}{\|c\|} \hline \text { Para. 6-9 } \\ \hline \text { Table 6-3 } \\ \hline \end{array}$ | Intermediate |
| 465M | 80009 | 6-18 | Oscilloscope, Dual Trace, Portable AN/USM-425(V)1 | Para. 6-9 | Intermediate |
| 8040A-01 | 89536 |  | Digital Voltmeter | Para. 6-9 |  |
| 8640B-003 | 28480 | 6-18 | SG-1093/U AM/FM Signal Generator | Para. 2-11.c. <br> Para. 6-9 | Intermediate Intermediate |

TABLE 1-7. RELATED TECHNICAL MANUALS

| Publication Number | Publication Title | Equipment Nomenclature |
| :--- | :--- | :--- |
| 31R2-2URR-256WC-1 | Scheduled Periodic Inspection <br> Work Cards | Receiver, Radio, R-2174(P)/URR |

TABLE 1-8. MNEMONICS AND ACRONYMS

| Mnemonic or | Description |
| :--- | :--- |
| Acronym | Accumulator |
| ACC | Acknowledge |
| ACK | Analog to Digital |
| A/D | Arithmetic Logic Unit |
| ALU | American Standard Code for Information Interchange |
| ASR | Automatic Send and Receive |
| BCD | Binary Coded Decimal |
| BITE | Built In Test Equipment |
| BFO | Beat Frequency Oscillator |
| BPS | Bits Per Second |
| BSC | Binary Synchronous Communication |
| CAM | Contents Addressable Memory |
| CLK | Clock |
| CMOS | Complementary Metal Oxide Semiconductor |
| CPS | Characters Per Second |
| CPU | Central Processor Unit |
| CRC | Cyclic Redundancy Check |
| CROM | Control Read Only Memory |
| CRT | Cathode Ray Tube |
| CRTC | Cathode Ray Tube Controller |
| CTS | Clear to Send |
| D/A | Digital to Analog |
| DMA | Direct Memory Access |
| DTL | Diode-Transistor Logic |
| Dø-D7 | Data Lines Zero thru 7 |
| EBCDIC |  |
| ECLtended Binary Coded Decimal Information Code |  |
| EOR | Emitter Coupled Logic |
| EPROM | Exclusive OR |
|  | Erasable Programmable Read Only Memory |
| FE | Framing Error |
| FET | Field Effect Transistor |
| FF | Flip-Flop |
| FIFO | First In First Out |
| GPIB | General Purpose Interface Bus |
|  |  |

TABLE 1-8. MNEMONICS AND ACRONYMS (cont'd.)

| Mnemonic or | Description |
| :--- | :--- |
| Acronym | Int |
| IC | Integrated Circuit |
| INT | Interrupt |
| I/O | Input - Output |
| IOCS | Input Output Control System |
| IRQ | Interrupt Request |
| ISB | Independent Sideband |
| LCD | Liquid Crystal Display |
| LED | Light Emitting Diode |
| LIFO | Last In First Out |
| LOC | Loop On-Line Control |
| LSB | Lower Sideband |
| LSD | Least Significant Digit |
|  |  |
| MOS | Metal Oxide Semiconductor |
| MPU | Microprocessor Unit |
| MSD | Most Significant Digit |
| MUX | Multiplexer |
|  | Overload Error |
| OE | Out of Lock |
| OOL | Overflow |
| OV | Printed Circuit |
| PC | Priority Interrupt Control |
| PIC | Phase Locked Loop |
| PLL | Programmable Read Only Request |
| PROM | Read Only Memory |
| ROM | Read/Write and Clock |
| ROMC | Reprasive OR |
| RPROM | Reprogrammable Programmable Read Only Memory |
| RPT | Repeat |
| RST | Restart |
| RTC | Real Time Clock |
| RTS | Request to Send |
| R/W | Read/Write |
| SMI | Static Memory Interface |
| S/N | Signal to Noise |
| SR | Service Request |
| STB | Strobe |
| TDSR | Transmit Data Service Request |
| TTL | Transistor - Transistor Logic |
|  | Universal Asynchronous Receiver Transmitter |
| UART | Upper Sideband |
| USB | Universal Synchronous Receiver Transmitter |
| UST |  |

## CHAPTER 2 <br> INSTALLATION

2-1. INTRODUCTION. This chapter provides information for receiving, handling and installation of the R-2174(P)/URR Radio Receiver. Section I describes the logistics involved in unpacking, receiving, mounting requirements, ventilating requirements and cable connections. Section II of Chapter 2 describes manpower requirements, the step by step instructions for installing the options in the Receiver and the initial checkout procedure to ensure proper installation.

## Section I. INSTALLATION LOGISTICS

2-2. UNLOADING AND UNPACKING. The carton containing the Receiver is 3.6 cubic feet and weighs approximately 34 pounds. No special unloading equipment or procedures are required, except to handle the carton with the normal care given to any shipping carton containing electronic equipment. Any options that may be used with the receiver are normally supplied in a separate package. The packaging details for the options are dependent on the individual site requirements. To unpack the Receiver, refer to Figure 2-1 and perform the following procedures:
a. Open the top of the shipping carton and fold back the flaps.
b. Lift out the top foam cushion.
c. Carefully lift out the wrapped box containing the Receiver. Because of the weight of the receiver (approximately 30 pounds) it is recommended that two people lift out the Receiver.
d. Remove the barrier wrapping from the interior container.
e. Open the interior container and carefully lift out the Receiver.
f. Place the Receiver on a convenient work bench and visually inspect the Receiver for any damage.
g. Replace all packaging material back in the shipping carton. Save all material in the event that the Receiver must be reshipped.

2-3. HOUSING. The unit may be housed (stored) from inclement weather in any structure that will sustain a temperature between $-40^{\circ}$ and $+70^{\circ} \mathrm{C}$ and a relative humidity of 10 to 95 percent. The unit has an indefinite shelf life stored under the above conditions except for the nickel cadmium battery contained on circuit card assembly A6A2. Shelf life is reduced to 3 years if the battery is left fitted to the circuit board.

2-4. RECEIVING DATA. Table 2-1 shows dimensions and weights of the cartoned Receiver as well as the Receiver. Figure 2-1 illustrates the shipping carton for the Receiver.

TABLE 2-1. DIMENSIONS AND WEIGHTS

| Item | Dimensions | Weight |
| :--- | :--- | :--- |
| Cartoned Receiver | 10.5 inches high; 23.5 inches wide; 25.5 inches <br> deep. Cubage: 3.64 cubic feet. | Approximately <br> 34 Pounds. |
| Receiver | Rack size: 5.1 inches high; 16.6 inches wide; <br> 18.4 inches deep. Front Panel: 5.25 inches <br> high; 19.0 inches wide. Overall depth with <br> handles; 20.0 inches. Cubage: 0.9 cubic feet. | 30 Pounds. |

TABLE 2-2. ITEM LIST REQUIRED FOR INSTALLATION

| Item | Description/Purpose |
| :--- | :--- |
| 1. | Receiver, Radio R-2174(P)/URR. |
| 2. | Power Cable, W18. Connects Receiver to primary power source. |
| IF Bandwidth Filter. The Receiver may be operated with from one to seven |  |
| d. | different bandwidth filters with a restriction of 5 maximum symmetrical |
| filters. A total of 15 different filters are available. |  |
| 4. | Mating Connector. Mates with Receiver rear panel AF OUT connector J3. <br> 5. (Optional) |
| ISB Circuit Card Assembly, A5. Permits the Receiver to operate in the ISB |  |
| mode (receiving both upper and lower sidebands simultaneously). |  |
| Remote Control Circuit Card Assembly, A6AI. Permits the receiver to be |  |
| operated from a remote control device. A mating connector is supplied with |  |
| this option. |  |

2-5. INSTALLATION REQUIREMENTS. The Receiver is designed to be mounted in a standard 19 inch rack. The sides of the Receiver have been drilled and tapped to accept Jonathan type 110QD-16-2 slides. The use of slides, however, is optional and is dependent on the individual site requirements. If slides are installed, it is recommended that cable retractors be used to simplify extending the Receiver out of the rack. If the unit is rack mounted without the slides, access to the rear panel must be provided for connection of cables and test equipment. Refer to Table 2-1 for space dimensions required for mounting the equipment and to Table 2-2 for a list of the items required for installing the Receiver.

## WARNING

Prior to installation of slide rails, visual inspection of receiver shall be performed to prevent injury to personnel and damage to equipment.
a. Remove top cover (see page 7-12 Fiq. 7-1, Index 1-1). Observe routing of AC Line Switching Cable, W20 (see page 7-15, Fig. 7-1, Index 1-64). If cable W20 is located opposite (predrilled and tapped holes) cable should be bent 90 degrees at AC Connector J2. This should remove cable from opposite (predrilled and tapped holes). Use only 10-32 x $3 / 8$ inch screws to mount slides. MIL-STD-454G states that turns of screws exposed should not be more than $11 / 2$ times the diameter of the thread (past drilled and tapped holes).

## Change 1 2-2



1. Exterior Contalner - 25-1/2" long by 23-1/2" wide by $10-1 / 2^{\prime \prime}$ deep.
2. Top and Bottom Cushions - $25-5 / \mathbf{8}^{\prime \prime}$
long by $23-5 / 8^{\prime \prime}$ wide by $5-1 / 4^{\prime \prime}$ deep.
3. Interior Container - 21 " long by 19-1/4" wide by 5-1/4" deep.
4. Front Panel Cushion - 19 " long by 1-1/2" wide by $5-1 / 4^{\prime \prime}$ deep.
5. Side Cushions - $19^{" 1}$ long by 4-1/2" wide by $3 / 4^{\prime \prime}$ doep.
6. Rear Panel Cushion - 16-1/4" lons by 4.5/8" wide by $1.5 / 8^{\prime \prime}$ deep.
7. Intimate Wrap.
8. Barrier Wrap.
9. Dericcant.
10. R-2174(P)/URR Radio Recetver

Figure 2-1. Packaging Details, Receiver

## CAUTION

The top and bottom covers on the Receiver, as well as the heat sinks on the rear panel, must be unobstructed to permit proper air circulation. The power dissipation of the Receiver is approximately 50 Watts with half of the power dissipated as heat. In most installations, special cooling will not he required.
b. The options associated with the Receiver include plug-in IF bandwidth filters, independent sideband operation, and operating the Receiver through a remote control device. One or more of the plug-in filters must be installed before the receiver can operate, The ISB and remote control capabilities, however, are optional. Because of the different possible filter combinations and the options, it is recommended that a Receiver configuration chart or log book be maintained for each Receiver. The configuration chart or log book would list the IF bandwidth filters currently in the Receiver and the operating options installed.
c. All system interface connections to the Receiver are made through the rear panel connectors listed in Table 2-3.

TABLE 2-3. REAR PANEL CONNECTORS

| Reference <br> Designation |  | Nomenclature |
| :--- | :--- | :--- |

## Section II. INSTALLATION PROCEDURES

2-6. MANPOWER REQUIREMENTS. To install the options in the Receiver, install the Receiver in the system, and perform the tests required to ensure operational performance requires one man for one hour.

2-7. INSTALLATION. The installation procedures for the Receiver are divided into three main steps. The first step is to install the plug-in filters and any options that may be associated with the Receiver. The second step is to perform an initial check-out of the Receiver, and the third step is to make the required system connections to the Receiver. Paragraphs 2-8, 2-9 and 2-10 contain detailed instructions for performing these steps.

2-8. INSTALLATION OF RECEIVER OPTIONS. The three primary operating options associated with the Receiver include:

- Operation with different bandwidth options through the use of plug-in bandidth filters.
- Operation of the Receiver with a remote device.
- Operation of the Receiver in the ISB mode.

The following paragraphs detail the steps necessary to install these options.
a. IF Bandwidth Filter Installation. The Receiver may be supplied with from 1 to 7 plug-in filters. Table 1-5 lists the filters available with the initial procurement. Additional filters may be available through later procurements, refer to Table 1-5. The filters are mounted on the A4 board, accessible from the top of the Receiver. To gain access to the A4 board, loosen the six quarter-turn fasteners holding the top cover to the Receiver and remove the top cover. An RF shield is mounted over the filter sockets on the A4 board. Remove the three screws holding this shield to the chassis and remove the shield. Figure 2-2 Illustrates the seven filter slots on the A4 board.
(1) If a Receiver configuration chart has been prepared for this particular Receiver (as recommendedin Paragraph 2-5a.) refer to the chart and determine if the Receiver is to be operated with the ISB option. If a Receiver configuration chart was not prepared, determine if an ISB option is to be included by checking the shipping data or the station manual, as appropriate. If the Receiver is to be operated with the ISB option, both upper and lower sideband filters must be installed in the Receiver.

## NOTE

## If LSB is specified for installation, it must be installed in filter slot FL1.

If the Receiver is not to be operated with the ISB option, either a lower sideband filter or symmetrical sideband filter may be installed in the FL1 position. If a lower sideband filter is installed, the companion upper sideband filter must also be installed in one of the remaining filter positions. If a symmetrical filter is to be used to provide sideband operation, the Receiver will use the filter installed in the FL1 position. This is for both sidebands by putting a 1.8 kHz in the appropriate frequency offset in the first and last local oscillators. (This assumes a 3.2 kHz symmetrical filter will be used for this application.)
(2) The remaining filters may be installed in any sequence in filter positions FL2 through FL7. (NOTE: only 5 symmetrical filters allowed.) However, in order to simplify system operation and troubleshooting, it is recommended that a format be established and used for all Receivers at a particular site. A typical format would be to insert the USB filter (if used) in position FL2 and insert filters with increasing bandwidths in filter positions 3 through 7.
(3) Once the filter complement and arrangement has been determined, the following procedure should be used to insert the filters into the Receiver. Refer to Figures 2-2 and 2-3.
(a) Working from the front of the Receiver, position the filter to be used for LSB operation over filter position FLI (the filter position closest to the rear of the Receiver). Make certain that the large pins are aligned with the large sockets and the smaller pins are aligned with the smaller sockets.
(b) Carefully push down on the filter to insert the pins into the sockets. Relatively light pressure is required to insert the pins into the sockets. If the filter does not easily slide into place, recheck the pin/socket alignments.
(c) Insert the appropriate filters into filter positions FL2 through FL7 (as required), using the procedures described in steps 1 and 2.
(d) After all filters have been inserted, visually inspect the filters to insure that they are poperly seated. The bottom of the filters should be flat against the surface of the A4 board.


Figure 2-2. Location of IF Bandwidth Filter Slots


Figure 2-3. Location of Filters and Jumpers, A4 circuit card assembly.
(e) Replace the RF shield over the filters and secure the shield in place by tightening the three screws.

## NOTE

If the Receiver is equipped to be operated from a remote control device or in the ISB mode, continue with the procedures described in Paragraphs 2-8b and 2-8c. If the Receiver is not to be operated with these two options, replace the top cover on the Receiver chassis and refer to the initial checkout procedure described in Paragraph 2-9.
b. Remote Control Interface Installation. If the receiver is to be operated with the remote control option the A6A1 circuit card assembly must be installed in the Receiver. In order to install this card the A6A2 circuit card assembly must be removed from the Receiver and mated with the A6A1 card. The two cards are then inserted back into the Receiver as an assembly. The following procedure details the steps necessary to install the cards.
(1) Working from the front of the Receiver, disconnect the A9W1 cable assembly from the A6A2 card. This cable assembly connects to A6A2J1 located on the front of the A6A2 card. Figure 2-4 illustrates the A6A2 and A6A1 circuit card assemblies installed in the Receiver and the location of A9W.
(2) Remove the three screws located along the top edge of the A6A2 card and carefully remove the card from the Receiver.

## CAUTION

Do not place the A6A2 circuit card assembly on any conductive material. Failure to comply may result in shorting the battery contained on this card.
(3) Check the rear panel of the Receiver. A blank plate may be covering upthe hole (located on left hand side of the rear panel) for the remote control interface socket. If the blank plate is present, remove the two screws and lock washers holding the plate and remove the plate, saving the screws and lock washers.
(4) A hard-wired fink, designated LK1, must be removed from the A6A2 card when the remote control option is used. The link is physically located between U1 and U2 (the two 40 pin LSI chips located near the J2 connector). Use a pair of cutters (or a low wattage soldering iron) to remove the link. Refer to Figure 2-5 for location of the link.
(5) There are six different mechanical links located on the A6A1 circuit card assembly that must be installed according to the external data protocol to be used. Table 2-4 lists the mechanical links required for each interface. Use a low wattage soldering iron to install or remove the links as required for data protocols. Refer to Figure 2-6 for location of the links.
(6) Align the 50 pin male connector on the A6A1 circut card assembly with the 50 pin female connector on the A6A2 circuit card assembly as shown in Figure 2-7. Mate the two connectors.
(7) Carefully place the two cards into the receiver chassis. Make certain the remote control interface connector (A6A1J1) clears the hole in the rear panel.


Figure 2-4. Receiver Options Installation Detail


Figure 2-5. A6A2 Circuit Card Assembly, Location of Jumper


Figure 2-6. Optional A6A1 Circuit Card Assembly, Location of Jumpers


Figure 2-7. Mating A6A1 and A6A2 Circuit Card Assemblies

## CAUTION

If a nylon washer is supplied as part of the hardware kit for A6A1 this must be used on center-top of A6A1 circuit card with attached screw to prevent the mounting post from shorting the circuit card.
(8) Position the two cards in the card guide attached to the receiver chassis. Be sure the cards are fitted firmly in the bottom of the card guides. Secure the cards to the side of the chassis with 6 screws ( 3 screws on each card) and nylon washer on center post of A6A1 if required. Secure the interface connector to the rear panel with two screws and lock washers saved from blanking plate. Figure 2-4illustrates the two cards installed in the Receiver.
(9) Reconnect the A9WI cable assembly (removed in step 1) to connector A6A2J1.
c. Independent Sideband Option Installation If the Receiver is to be operated in the ISB mode the AS circuit card assembly must be installed in the Receiver. The A5 card is physically located towards the rear of the Receiver, between the A1 assembly and the A4 circuit card assembly; as shown in Figure 2-4. The following procedure details the steps necessary to install the AS circuit card assembly. Refer to Figures 2-3 and 2-4 for the location of cables and connectors.
(1) Place the metal baseplate shield (curved edge upwards) on the four standoffs and secure with four screws into the standoffs using screws and washers supplied. (NOTE: Flat washer against plate.)
(2) Position the A5 circuit card assembly so that the ribbon cable is opposite the J 8 connector on the A 4 circuit card assembly. Secure the A5 circuit card assembly to the metal baseplate with 4 screws and washers with flat washers next to circuit card.
(3) Plug in the ribbon cable (A5W1) from the A5 card into J8 on the A4 board. Connect coaxial cable W10 between A4J6 and A5J3. Connect coaxial cable W11 between A4J3 and A5J1.
(4) The mechanical jumper on the A4 circuit card assembly must be properly positioned for ISB/SSB operation. The jumper, designated LK1, is physically located to the right of filter position FL1 when looking from the front of the Receiver. Refer to Figure 2-3. With the AS circuit card assembly installed, the jumper must be connected across the two terminals designated ISB.

TABLE 2-4. JUMPER OPTIONS, REMOTE CONTROL INTERFACE

| Link Designation | 188C/232C/423 Interface | RS442 Interface |
| :---: | :--- | :--- |
|  | Install | Install |
| LK1 | Remove | Remove |
| LK2 | Install | Remove |
| LK3 | Install | Remove |
| LK4 | Install | Remove |
| LK5 | Remove | Install |

2-9. INITIAL CHECKOUT PROCEDURE. The following procedure is designed to insure that the receiver is in an operational condition. Before attempting this procedure, make certain that all steps described in Paragraph 2-8 have been completed as required.
a. Place the Receiver on a convenient work table.
b. Verify that the front panel POWER switch is set to the OFF position.
c. Connect a ground strap between the GROUND lug on the rear panel and a good electrical grond.
d. Determine the power source, in terms of voltage and frequency, that will supply the primary power for the receiver. Ensure that frequency is between 43 and 420 Hz and the voltage is either 100, 120, 220, or 240 Volts. Change fuse to $1 / 2 \mathrm{amp}$ for 240 -Volt operation.
e. On the rear of the Receiver, slide the transparent plastic cover to the left, remove the printed circuit card switch and reinsert it so that the primary power voltage ( $10 \%$ ), to be connected to the Receiver, is visible after reinsertion of the circuit card switch. Slide the plastic cover to the right.
f. Connect the primary power cable W18 plug P2 to connector A10J1 on the rear panel of the Receiver.
g. Connect Plug P1 of primary power cable W18 to the primary power source.
h. Place the front panel POWER switch to the ON position.
i. The Receiver will probably indicate a random display since the Receiver has not been initialized. To initialize the Receiver, depress the LOCK pushbutton and while still holding the LOCK in, depress the AM pushbutton; then release in the opposite order they were depressed. This will cause the microprocessor to initiate the BITE sequence, to check and record the filters and other operating parameters of the Receiver. During the initialization sequence (which takes approximately 1 minute) the Receiver display will change to indicate that the sequence is in process. At the completion of the sequence, the Receiver will display the last entered data. If the sequence fails, refer to Chapter 6 for maintenance procedures.
j. To verify that the microprocessor and the front panel control logic are operating properly, depress the TUNE RATE pushbutton and turn the frequency selection knob. Verify that the FREQUENCY display increases and decreases accordingly. Depress the USB, LSB, then CW and BWI through BW5 pushbuttons in sequence. Verify that the correct filter value is displayed. Depress the ENTER pushbutton and depress pushbuttons $0,1,2,3,4,5,6$, and 7 . Verify that the digits are displayed in the FREQUENCY display.
k. After verifying frequency display select all AGC modes, all detector modes and in turn verify the operation and display on the mode LCD, also verify the BFO by stepping to CW then selecting BFO CENTER, BFO and tuning the BFO (tuning knob).
I. Set the POWER switch to the off position and disconnect the power cord.
m. This completes the initial checkout procedure for the Receiver. Continue with the procedures detailed in Paragraph 2-10 to install the Receiver in the system.

2-10. SYSTEM CONNECTIONS. System connections for the Receiver are based on the individual site requirements and the options associated with a Receiver. The following paragraphs describe in detail all of the procedures required to install the Receiver. If a particular paragraph does not apply it should be ignored. Additionally, the individual site requirements will determine the most effective method of installing the Receiver. In some installations, it may be easier to pre-wire an entire equipment rack and then install the Receivers. In other cases it may be easier to install the Receiver and then add the wiring. Figure 2-8 illustrates the connectors located on the rear panel of the Receiver.


Figure 2-8. Rear Panel Connectors and Controls
a. RF IN, Connector J1. The RF input from the antenna to the Receiver is made through the rear panel connector, J 1. This connector is a standard N -type female connector and will mate with any standard N -type male connector. The Receiver is shipped with a plastic dust cover over connector J 1 , which must be removed before the antenna connection can be made. The input impedance at the connector is 50 ohms, unbalanced, with a VSWR of less than 2:1 over the operating frequency range of the Receiver.
b. IF OUT, Connector J2. The second IF output signal at 455 kHz is brought through this female BNC-type connector; which will mate with any standard male BNC-type connector. The plastic dust cover must be removed before the connection can be made. The output signal level at the connector is approximately $-!0 \mathrm{dBm}$ into 50 Ohms.
c. AF OUT, Connector J3. The AF OUT connector on the rear panel is a 25 pin D-type connector providing audio, AGC, and fault status outputs. Table 2-5 lists the signal outputs on each pin which are further described in the following paragraphs:
(1) If the Receiver is not equipped for ISB operation (A5 circuit card assembly is not installed) the Line 1 and Line 2 outputs-are not used. The Monitor Line output and the Loud-speaker output, along with the front panel PHONES jack, provide the audio output for the Receiver for all operating modes.
(2) If the Receiver is equipped for and is to be operated in the ISB mode, the Line 1 output will contain the USB audio and the Line 2 output will contain the LSB audio. The Monitor Line output, the Loudspeaker output, and the front panel PHONES jack will provide either USB or LSB audio; as controlled by the front panel ISB U/L pushbutton.
(3) If the Receiver is equipped for ISB operation but is not operating in the ISB mode, audio output will be the same as step 2.
(4) The Line 1, Line 2, and Monitor Line provide a 1 mW balanced output at 600 Ohms. The amplitude level is adjustable from the front panel MAIN and I-LSB LINE LEVEL controls. The Loudspeaker output is I Watt into 8 Ohms, controlled from front panel AF GAIN control.
(5) The Main IF and ISB-LSB AGC Monitor/Input terminals may be used to control the AGC for diversity combining or to monitor the AGC voltage. For diversity combining, the Main IF AGC signal (on pin 21 ) of one receiver may be connected to pin 21 of the second Receiver. If the Receiver is equipped for ISB operation, pin 22 of one Receiver may be connected to pin 22 of the second Receiver. The AGC circuits in both receivers will automatically respond to the proper signal.
(6) The Fault Indicator output will be high (logical $1,+2$ Volts to +5 Volts) when a fault is not present in the receiver. This output will go low (logical 0,0 Volts to +0.8 Volts) when one of the synthesizer circuits is out of lock.
(7) A mating connector (Cannon Type DB-25P) with a connector shell is supplied with the receiver. Use the following procedure, and the pin number/signal designation information in Table 2-5 to wire the connector.
(a) Slide the connector shell over the cable to be used.
(b) Solder the cable wires to the comector as required (refer to Table 2-5). The connector pin numbers are indicated on the front of the connector.
(c) Slide the connector back into the connector shell. Place one of the spring clips on top of the connector with the curved edge pointing upwards. Secure the spring clip and the connector to the connector shell with the self-tapping screw. Repeat this procedure for the other side of the connector.
(d) Connect the wired cable to the rear panel connector $\mathrm{J3*}^{*}$ and secure with the two connector spring latches.

## TABLE 2-5. AF OUT J3 PIN CONNECTIONS

| Pin Number | Signal Designation |
| :---: | :---: |
| 1 | Output Line 1 Output (Used only with ISB Option). Provides USB |
| 2 | Center Tap $\}$ output during ISB operation; AM/FM/CW/USB output dur- |
| 14 | Output $\int$ ing non-ISB operation. |
| 3 | Output Line 2 Output (Used only with ISB Option). Provides LSB |
| 16 | Center Tap $\} \quad$ output during ISB operation or SSB operation. |
| 15 | Output $\}$ |
| 4 | Output $\}$ Monitor Line Output. Provides AM/FM/CW/SSB output dur- |
| 5 | Center Tap $\} \quad$ ing non-ISB operation; provides switch controlled selection |
| 17 | Output of USB or LSB during ISB operation. |
| $18$ | Output ${ }^{\text {Signal Ground }}$ \} Loudspeaker Output. Same as Monitor Line Output. |
| $6$ | Signal Ground |
| 7, 8, 9 | Ground (Cable Shield) |
| $\begin{aligned} & 10,11,12,13 \\ & 19,20,24,25 \end{aligned}$ | Not connected |
| 21 | Main IF Diversity AGC Monitor/Input |
| 22 | ISB Lower Sideband Diversity AGC Monitor/Input (used only with ISB option) |
| 23 | Fault Indicator (Low Indicates Fault) |

d. REF IN/OUT, Connector J7. This rear panel connector is used in conjunction with rear panel switch S2. With the switch set to the EXT position the Receiver will accept an external 1,5, or 10 MHz reference input through connector J7. With switch S2 set to the INT position, the Receiver will supply a 1,5 , or 10 MHz reference output through connector J7*. Connector J7 mates with any standard male BNC-type connector. The input circuit has an input impedance of 50 Ohms and will operate with peak-to-peak signal levels of 1.0 Volts, +0.5 Volts.
e. Remote Control, Connector A6A1W1J1. The Receiver is supplied with a mating connector for the remote control interface connector Table 2-6 lists the pin numbers and signal designations for this connector. Note that the function on the first nine pins ( A through J) vary according to the interface to be used. The remaining pins ( K through c ) are the same for all interfaces. Table 2-7 lists the pin number/baud rate selection for the receiver, and Table 2-4 lists the A6A1 circuit card assembly jumpers for the different interfaces. (NOTE: These pins must be connected to ground or left unconnected to provide correct address, baud rate and parity for correct operation.)
*Refer to Figure 6-13A and Figure 8-11 for frequency selection link options.
(1) Slide the mating connector shell over the cable to be used for the remote control.
(2) Connect the pins on the wires, and then using the special insertion tool provided, insert the pins into connector housing in accordance with Table 2-6 for the interface to be used. The connector pins are designated on the front of the connector.
(3) After carefully checking all wiring, slide the connector into the connector shell and secure the cable clamp. Attach the connector to A6A1WIJ1 and secure.
(4) Visually inspect the A6A1 circuit card assembly to insure that all jumpers are installed in accordance with Table 2-4.

## WARNING

## The Voltages associated with the power input to the Receiver can be dangerous to personnel. Use caution when connecting or disconnecting the power cable.

f. Power Cable, Connector A10J1. The power cable, W18, attaches to the rear panel connector A10J1. Before attaching this cable, verify that the printed wiring card switch is properly inserted (the operating voltage designation is visible). Access to the card is made by sliding the transparent plastic cover to the left. Additionally, the receiver should be grounded by attaching a suitable ground wire to the rear panel GROUND lug before the ac power cable is attached.

2-11. INITIAL OPERATING PROCEDURE. The following procedure provides an overall check on the Receiver operation and the system interface connection. Before attempting this procedure, verify that all appropriate options have been installed in accordance with Paragraph 2-8 and that all appropriate system interface connections have been made in accordance with Paragraph 2-10.
a. Turn the Receiver on by setting the front panel POWER switch to the ON position.
b. A random display may be present on the Receiver. To correct this, the Receiver must be initialized by running the BITE sequence (Built In Test Equipment). This is accomplished by depressing the LOCK pushbutton and, while still holding LOCK in, press the AM pushbutton; then release in the opposite order they were depressed. The Receiver will automatically run through the entire BITE sequence. If no errors are detected, the Receiver will return to the last entered setting upon completion of the BITE sequence. If an error is detected, the BITE sequence will stop and display a number. If this occurs, refer to the fault isolation procedures contained in Chapter 6(table 6-2) of this manual for further action. If the BITE sequence successfully completes but the resulting display continues to be erroneous, it may be due to a loss of memory back-up. If this occurs, ensure that the receiver is in LOCAL Mode and exercise all pushbutton front panel functions including MAIN and BFO Tuning. If this does not produce a coherent display, refer to Chapter 6 for further fault isolation. After verifying display select all AGC modes, all detector modes and in turn verify the operation and display on the mode LCD; also verify BFO by selecting CW and BFO CENTER, BFO and then tuning the BFO.
c. Connect the signal generator, listed in Table 1-5, to connector J1 RF IN. Adjust the signal generator for a 3.5000000 MHz output at -97 dBm . Set Receiver frequency to 03.5 MHz , CW mode, BFO offset to plus 1 kHz and short AGC time constant.
d. Press the METER RF/AF pushbutton for AF indication in the upper left portion of the liquid crystal display (LCD). The AF should indicate 0 dB . If indication is not 0 dB , use a screwdriver to adjust the MAIN-LINE LEVEL adjust on the front panel for 0 dB indication on the LCD.
e. If the Receiver is equipped with the ISB option, depress the ISB U/L pushbutton to obtain an I-LSB indication on the display (the pushbutton may have to be depressed twice). NOTE: The Receiver frequency must be adjusted to 1.8 kHz higher than the generator frequency to obtain

TABLE 2-6. REMOTE CONTROL INTERFACE CONNECTOR SIGNAL DESIGNATION

| Pin <br> Designation | $\begin{aligned} & \hline \text { MIL STD 188C } \\ & \text { Interface (See Note) } \\ & \hline \end{aligned}$ | E1A-RS232C/RS423 Interface (See Note) | $\begin{aligned} & \text { E1A-RS422 } \\ & \text { Interface (See Note) } \end{aligned}$ |
| :---: | :---: | :---: | :---: |
| A | System Ground | System Ground | System Ground |
| B | Not Used | Data Out A-Mark Neg. | Data Out A-Mark Neg. |
| C | Data Out Ground | Data Out Ground | Not Used |
| D | Data Out | Not Used | Data Out B-Mark Pos. |
| E | Jumper to Pin F | Not Used | Not Used |
| F | Jumper to Pin E | Data in A-Mark Neg. | Data In A-Mark Neg. |
| G | Data In Ground | Data In Ground | Not Used |
| H | Data In | Jumper to Pin J | Data In B-Mark Pos. |
| J | Not Used | Jumper to Pin H | Not Used |
| K | Receiver Address D1-1* | Receiver Address D1-1* | Receiver Address D1-1* |
| L | Receiver Address D1-2* | Receiver Address D1-2* | Receiver Address D1-2* |
| M | Receiver Address D1-4* | Receiver Address D1-4* | Receiver Address D1-4* |
| N | Receiver Address D1-8* | Receiver Address D1-8* | Receiver Address D1-8* |
| P | Receiver Address D2-1* | Receiver Address D2-1* | Receiver Address D2-1* |
| R | Receiver Address D2-2* | Receiver Address D2-2* | Receiver Address D2-2* |
| S | Receiver Address D2-4* | Receiver Address D2-4* | Receiver Address D2-4* |
| T | Receiver Address D2-8* | Receiver Address D2-8* | Receiver Address D2-8* |
| U | /Parity Select* | /Parity Select* | /Parity Select* |
| V | Even/Odd Parity* | Even/Odd Parity* | Even/Odd Parity* |
| W | Baud Rate B4** | Baud Rate B4** | Baud Rate B4** |
| X | Baud Rate B3** | Baud Rate B3** | Baud Rate B3** |
| Y | Baud Rate B2** | Baud Rate B2** | Baud Rate B2** |
| Z | Baud Rate B1** | Baud Rate B1** | Baud Rate B1** |
| a | Ground | Ground | Ground |
| b | Ground | Ground | Ground |
| c | System Ground | System Ground | System Ground |

NOTE: Circuit Card A6AI must be linked as shown in Table 2-4 for each interface configuration. D1=LSD, D2=MSD.
*Logic $1=$ open circuit; Logic $0=$ ground for Address, Parity option and Baud rates.
**Refer to Table 2-7 for Baud rates.
TABLE 2-7. BAUD RATE SELECTION

| Pin | Baud Rate |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Desig. | 50 | 75 | 110 | 134.5 | 150 | 300 | 600 | 1200 | 1800 | 2000 | 2400 | 3600 | 4800 | 7200 | 9600 | 19200 |
| W | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| X | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |
| Y | 0 | 0 | 1 | 1 | 0 | 0 | 1 | 1 | 0 | 0 | 1 | 1 | 0 | 0 | 1 | 1 |
| z | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 |

an audio output. With the meter set to display $A F$, verify that a 0 dB indication is present. If a 0 dB indication is not present, use a screwdriver to adjust the I-LSB adjustment on the front panel to obtain a reading of 0 dB .
f. Remove the signal generator and connect an antenna to J1 RF IN on the rear panel.
g. Connect a headphone to the front panel PHONES jack. Depress the AM pushbtton and tune the Receiver to a local AM broadcast frequency. The Receiver may be tuned either with the front panel frequency selection control or set directly to the desired frequency through the use of the numeric keypad. When using the tuning knob, depress the TUNE RATE pushbutton to select the desired rate. The display will indicate FAST or SLOW under TUNING. If no display is present the Receiver will be in the fine tuning rate mode. When using the numeric keypad, depress the ENTER pushbutton and then enter the desired frequency with the numeric keys, including leading zeros.
h. Press the LOCK pushbutton and rotate the tuning dial in both directions. Frequency indication should not change and station selected should not change.
i. Select the desired bandwidth by depressing the appropriate bandwidth pushbutton (BW1 through BW5). Adjust the audio level in the headphone with the front panel AF GAIN control.
j. Depress the MAN, SHORT, MED, and LONG pushbuttons in sequence. Verify that the correponding indication is present in the display. Note that these pushbuttons are push on/push off and that the manual mode may be used in conjunction with the short, medium, or long AGC modes.
k. The manual mode may be used to set the threshold level on any of the AGC modes. To set the threshold, press the MAN pushbutton and set the meter to display RF by depressing the METER RF/AF pushbutton. Adjust the front panel IF GAIN control to obtain the desired threshold level on the meter. With none of the AGC modes indicated the Receiver will operate in the manual mode. With any one of the AGC modes selected the receiver will operate in the AGC mode with the threshold level as set in the manual mode.
I. Press the BW1 through BW5 pushbuttons in sequence and verify that the proper bandwidth is indicated in the display and the audio quality in the PHONES varies with the bandwidth selected.
m. Press the CW and BFO pushbuttons. Verify that the BFO display varies as the front panel tuning knob is turned. Press the BFO CENTER pushbutton and verify that the BFO display returns to zero. Set the front panel POWER switch to the off position.
n. This completes the initial operating check of the Receiver. Refer to the operating instructions contained in Chapter 4 for additional information on the operation of the Receiver.

## CHAPTER 3 <br> PREPARATION FOR USE AND RESHIPMENT

3-1. INTRODUCTION. This Chapter provides information required for preparing the R-2174(P)/URR Radio Receiver for use from a shipped or stored condition and for preparing it from an operating condition for storage or shipment. Section I describes the information required for preparing the unit from a stored or shipped condition, and installation and checkout procedures required for preparing the Receiver for use. Section II describes the procedures required for removing the unit from an operational rack and then packing it for storage-or shipment.

## Section I. PREPARATION FOR USE

3-2. UNPACKING. To unpack the Receiver and prepare it for installation and operation, refer to Paragraph 2-2 in Chapter 2 of this manual and perform the procedures as spelled out in that paragraph. Table 2-2 lists the items required to install the Receiver.

3-3. INSTALLATION AND PERFORMANCE CHECK. To prepare the receiver for operation requires the installation of all options associated with the Receiver and the performance of an initialization sequence. Paragraph 2-8 in Chapter 2 of this manual provides detailed instructions for installing these options and Paragraph 2-9 contains an initial checkout procedure. After the required system interface connections are made, as described in Paragraph 2-10, the Receiver performance may be checked by performing the procedures described in Paragraph 2-11

## Section II. PREPARATION FOR RESHIPMENT

3-4. REMOVAL. The Receiver must be disconnected and removed from its mounting rack before being prepared for reshipment. The following procedures describe the recommended sequence.
a. Ensure Receiver is turned off, then disconnect primary power cable W18 from its primary power source. Disconnect the other end of the cable from A10J1 on the rear panel of the Receiver.
b. Disconnect the antenna cable from J1-RF IN on the rear panel.
c. Disconnect the ground strap from the GROUND lug space located on the rear panel.
d. Remove any other equipment or cables connected to connectors or jacks on the rear panel.
e. Remove headphones, if they are inserted, from the PHONES jack located on the front pmel.
f. Remove the four mounting screws from the front panel securing the Receiver to the mounting rack.

## WARNING

The Receiver weighs approximately 30 pounds. Be careful as you slide the unit out of the rack. Have a firm grip on the handles as the weight leaves the rack so that it does not drop causing injury to legs or feet.
g. Grasp the Receiver by the handles on the front panel and slide the unit out of the rack. Place the unit on a bench.
h. This concludes the removal procedures. The unit may now be packed for storage or reshipment.

3-5. PACKING. To prepare the Receiver for storage or reshipment, it must be carefully wrapped and packaged as described in the following procedures.
a. Refer to Figure 2-1 in Chapter 2 and make sure that the shipping carton and all packing materials are in good condition.
b. Visually inspect the Receiver for damage and that all parts are present and secure. Record any damage or missing parts before packing.
c. Wrap the Receiver in the intimate wrap and place the unit in the interior container with cushions and desiccant as shown in Figure 2-1.
d. Close the interior carton and wrap with the barrier wrap.
e. Place the bottom cushion in the exterior carton and carefully place the interior carton into the cup of the cushion. Place the top cushion over the top of the interior carton.
f. Close and seal the exterior carton.

## CHAPTER 4 <br> OPERATION

4-1. INTRODUCTION. This chapter provides detailed instructions for operation of the R-2174(P)/URR Radio Receiver. The information is presented in three sections within the chapter. Section I describes the controls and indicators, Section II presents detailed operating instructions while Section III provides instructions for emergency operation.

## Section I. CONTROLS AND INDICATORS

4-2. GENERAL. The controls and indicators of the Receiver are shown in Figure 4-1 and listed and described in Tables 4-1 and 4-2. Both tables are printed in two columns. The left hand column, headed with NOMENCLATURE, lists the control or indicator exactly as it is spelled out on the Receiver front panel in upper case letters. If an item is not designated with nomenclature, such as the Tuning Knob, then it is listed in upper and lower case letters. The right hand column, headed with DESCRIPTION/FUNCTION, describes each control or indicator and gives its function.

TABLE 4-1. FRONT PANEL CONTROLS

| Nomenclature | $\quad$ Description/Function |
| :--- | :--- |
| POWER-ON | Double pole, single throw toggle switch. Provides on-off control for <br> Receiver by controlling primary power source to power supply. <br> Round black knob - optically coupled to a tuning encoder. Provides <br> selection of Receiver operating frequency or BFO frequency. (See <br> TUNE RATE and BFO pushbutton controls.) The rate of change <br> of the frequency depends on the speed the tuning knob is rotated and <br> the rate of tune selected through the TUNE RATE switch. Clockwise <br> rotation increases frequency, counterclockwise rotation decreases the <br> frequency. This knob is disabled by the LOCK pushbutton. |
| Left Keyboard | Pushbutton switches. |
| Mode Selection |  |
| AM, CW, USB, |  |
| LSB, ISB U/L, FM | These pushbuttons are used to select the operating mode of the receiver. <br> These pushbuttons are the push-on type. That is, when a pushbutton is <br> depressed the receiver will remain in that mode until a different mode is |
| selected. The ISB U/L pushbutton will be enabled only if the Receiver |  |
| is equipped with the ISB option. If the Receiver is equipped with this |  |
| option, depressing the pushbutton will activate both the USB and LSB |  |
| channels simultaneously. If the front panel display indicates I-USB the |  |
| PHONES jack will be connected to the upper sideband channel. Depress- |  |
| ing the ISB U/L pushbutton a second time will change the display to indi- |  |
| cate I-LSB and connect the PHONES jack to the LSB channel. The Moni- |  |
| tor Line and Loudspeaker outputs, through rear panel connector J3, will |  |
| also be switched by use of the ISB U/L pushbutton. |  |

TABLE 4-1. FRONT PANEL CONTROLS (Cont.)

| Nomenclature | $\quad$ Description/Function |
| :--- | :--- |
| Filter Selection | These pushbuttons are used to select specific bandwidths within the <br> BW1, BW2, BW3, <br> BW4, BW5 |
| Receiver. The Receiver is capable of accepting up to 7 plug-in IF <br> bandwidth filters. Receivers are usually configured with separate <br> upper and lower sideband filters, leaving a capacity of up to five sym- <br> metrical filters. The upper/lower sideband filters will be automatically <br> selected when the USB or the LSB mode is selected. The remaining <br> filters (up to a total of 5) are selected by depressing the appropriate <br> bandwidth selection pushbutton. BWI pushbutton will select the nar- <br> rowest bandwidth, BW2 will select the next-wider bandwidt, and so <br> on. Like the mode selection pushbuttons, these pushbuttons are the <br> push-on type, so that a new bandwidth is selected by depressing a dif- <br> ferent pushbutton. If only three symmetrical filters are installed, then <br> only pushbuttons BWI through BW3 will be in operation. |  |
| METER RF/AF | This pushbutton is used to cause the front panel meter to display either <br> an RF scale or an AF scale. The receiver will always display either <br> scale. Depressing the pushbutton will cause the display to switch either <br> from RF to AF or AF to RF. |
| AGC Mode Selection | The Receiver is designed to operate with three different gain control <br> MAN, SHORT, <br> modes: Manual, Automatic, and Automatic with a selectable threshold. <br> Depress the MAN pushbutton to cause the MAN indication to appear in <br> the display. If an automatic indication (SHORT, MED, or LONG) is |
| present in the display it may be deleted by depressing the corresponding |  |
| pushbutton. The Receiver is now in the manual mode with the gain |  |
| under control of the front panel IF GAIN control. When the Receiver is |  |



Figure 4-1. Front Panel Controls and Indicators, Radio Receiver R-2174()/URR

TABLE 4-1. FRONT PANEL CONTROLS (Cont.)

| Nomenclature | Description/Function |
| :---: | :---: |
| LOCK | This pushbutton is the push-on type. When the lock mode is activated, as indicated by the display, the front panel tuning knob will be disabled. The lock mode can be disabled by depressing the TUNE RATE or BFO pushbutton. |
| BFO, <br> BFO CENTER | These pushbuttons are enabled only if the Receiver is operated in the CW mode. Depressing the BFO pushbutton will permit selection of the BFO frequency through the front panel frequency knob. Depressing the BFO CENTER pushbutton will set the BFO to zero. Depressing the BFO CENTER pushbutton a second time will return the BFO to the previously selected frequency. The BFO pushbutton is disabled if the Receiver is in the BFO CENTER mode. |
| LOCAL/REMOTE | This push-on-push-off type pushbutton will set the receiver to operate either in the local mode (control of the receiver is through front panel) or the remote mode. Note that the Receiver must be equipped with the optional A6A1 remote control interface circuit card assembly to be operated in the remote control mode, however, if A6A1 is not installed selection of REMOTE will disable front panel control. |
| IF GAIN | This front panel control is used to establish the local Receiver IF GAIN control when the Receiver is operated in the manual mode; and to set the threshold level when the Receiver is operated in the automatic mode with a manually set threshold. |
| AF GAIN | This front panel control is used to control the level of the Receiver unbalanced audio outputs. |
| MAIN LINE LEVEL | This screwdriver set potentiometer is used to adjust the 600 ohm balanced line level when the Receiver is operated in the AM, FM, CW, or SSB modes. If the receiver is equipped to operate in the ISB mode (optional A5 circuit card assembly is installed) this control will also adjust the line level of the upper sideband when the receiver is operated in the ISB mode. |
| I-LSB LINE LEVEL | This screwdriver set potentiometer is used only when the Receiver is equipped with the optional A5 circuit card assembly. It is used to adjust the line level of the lower sideband when the receiver is operated in the ISB 600 ohm balanced. |

TABLE 4-2. FRONT PANEL INDICATORS

| Nomenclature | Description/Function |
| :--- | :--- |
| FAULT | Red LED indicator will be illuminated if any synthesizer is out of lock. |
| Right Display | Liquid Crystal Display (LCD) U3 |
| FREQUENCY-MHz <br> BFO kHz | Indicates the tuned frequency, in MHz, of the Receiver. <br> Indicates selected BFO frequency in kHz with plus or minus sign to <br> indicate direction of offset from IF frequency. |

TABLE 4-2. FRONT PANEL INDICATORS (Cont.)

| Nomenclature | Description/Function |
| :--- | :--- |
| Left Display | Liquid Crystal Display (LCD) U4 |
| $-100+10$ AF |  |
| 024681012 RF | Located upper left corner. Indicates audio level in dBm (upper scale), for <br> 600 ohm balanced monitor output line, RF in tens of dBs above 1 uVolt <br> (lower scale). Scale selected through METER-RF/AF pushbutton switch. |
| BW-kHz * | Located upper right corner. Indicates the IF bandwidth selected by <br> BW1 to BW5 pushbuttons switches, in kHz. |
| AGC, MAN, SHORT, | Located lower left comer. The AGC display is always present. The MAN <br> display will be present when the receiver is being operated in the manual <br> gain control mode. The AGC display (SHORT, MED, LONG) indicates <br> the type of automatic gain control being used by the receiver. If both <br> the MAN display and one of the AGC displays are present, it indicates <br> that the receiver is being operated in an automatic gain control mode <br> with a manually set threshold. <br> Located lower middle. The MODE display is always present, the re- |
| maining displays indicate the operating mode of the receiver. Note that |  |
| the I-display, associated with the LSB and USB displays, is used only |  |
| when the Receiver is equipped with the ISB option (A5 circuit card |  |
| assembly is installed). AUX is used when the forced bandwidth set-up |  |
| is in operation. |  |
| Located lower right. The TUNING display is always present. The FAST, |  |

## Section II. OPERATING INSTRUCTIONS

4-3. INTRODUCTION. The Receiver is supplied as a basic unit with one or more options supplied separately. All options associated with a particular Receiver must be installed in the basic unit before the Receiver can be operated. Paragraphs 2-7 through 2-11 describe the installation of the Receiver options, an initial check-out procedure, and the system interconnections associated with the Receiver.
a. The options associated with the Receiver include plug-in IF bandwidth filters, independent sideband operation, and operating the Receiver through a remote control device. One or more of the plug-in filters must be installed before the Receiver can operate. The ISB and remote control capabilities, however, are optional. Because of the different possible filter combinations and the options, some portions of the operating procedures will be different for different Receivers. In order to simplify the operation of the Receiver, it is recommended that a Receiver configuration chart or $\log$ book be maintained for each Receiver. The configuration chart or log book would list the IF bandwidth filters currently installed in the Receiver, the operating options, the type of remote control interface, etc.
b. Before attempting to operate the receiver, verify that all options associated with the Receiver have been installed. Additionally, the front panel controls and indicators described in Tables 4-1 and 4-2 should be reviewed before operating the receiver,

## CAUTION

Do not use sharp instruments to depress the keypads since this may damage the equipment.
4-4. LOCAL OPERATING PROCEDURES. The following paragraphs describe the basic operating procedures associated with the Receiver.
a. Initialization. The Receiver must be initialized after any plug in filters have been changed, any options have been installed or removed, or if the receiver memory data may have been disturbed. To initialize the Receiver, set the POWER switch to ON and depress the LOCK push-button and while still holding the LOCK in depress the AM pushbutton then release in the opposite order they were depressed. The Receiver will now perform an automatic initialization sequence to update the microprocessor memory with the current receiver configuration. At the completion of the initialization sequence (approximately 1 minute) the Receiver will return to its previous operating condition. If an error is detected during the initialization sequence the micro-processor will stop the sequence and display a two digit number in the frequency display. If this occurs, refer to the fault isolation procedures contained ir Chapter 6 of this manual.
b. Mode Selection. The Receiver may be operated in the AM, FM, CW, USB, or LSB modes. Additionally, if the Receiver is equipped with the optional A5 circuit card assembly, in the ISB mode. Selection of the desired operating mode is made by depressing the appropriate pushbutton on the left keyboard.
c. Frequency. The operating frequency in any of the modes must be selected and entered into the Receiver. The frequency may be selected by depressing the TUNE RATE pushbutton and adjusting the indicated frequency with the tuning knob or by depressing the ENTER key, then the appropriate numeral keys. Once the desired frequency is entered either by the keypad or tuning knob it may be finely adjusted by selecting the "fine" rate with the TUNE RATE key then rotating the tuning knob. To lock the frequency, press the LOCK pushbutton. This disables the tuning knob and prevents inadvertent frequency change.
d. BFO. The internal beat frequency oscillator is tuned by depressing the BFO pushbutton and adjusting the BFO through the tuning knob. The BFO offset is indicated as above (+) or below (-) the IF frequency. Depressing the BFO CENTER pushbutton will cause the BFO to go to zero. Depressing the BFO CENTER pushbutton a second time will return the BFO to its previous setting.
e. Bandwidth. A total of seven different IF bandwidth filters may be installed in the Receiver. If sideband filters are installed, the Receiver will automatically switch in the appropriate filter when the LSB or USB operating mode is selected. If a symmetrical sideband filter is used for sideband operation the receiver will use the filter installed in the FL1 position for both sidebands by putting an offset in the first and last local oscillators. During the initialization sequence the microprocessor will assign the remaining filters to front panel pushbuttons BW1 through BW5 in an ascending order. That is, depressing BWI will select the narrowest bandwidth, BW2 the next widest bandwidth, and so on.
f. AGC Operation. The Receiver may be operated in any one of three different gain control operating modes: manual, automatic, and automatic with a manually set threshold. To set the Receiver for manual operation, depress the MAN pushbutton and verify that none of the automatic modes (SHORT, MED, and LONG) are shown in the display. If an automatic mode is shown, depress the corresponding pushbutton to delete the mode. Depress the METER RF/AF pushbutton (if required) to display the RF scale. Adjust the front panel IF GAIN control to select the desired gain. To set the receiver for automatic gain control operation, depress the MAN pushbutton. The receiver will switch out of the manual mode to the SHORT AGC mode. The medium and long AGC operating modes may be selected by depressing the MED or LONG pushbuttons. To operate the Receiver in the automatic mode with a manually set threshold, set the Receiver to the manual mode and adjust the threshold as previously described. When the desired level is set, select the AGC mode by depressing the SHORT, MED, or LONG pushbutton. The display will indicate both the MAN indication and the selected automatic mode. To change the automatic mode, depress the selected automatic gain control pushbutton.
g. Local/Remote. Provision is made on the front panel of the Receiver to operate the unit either from the front panel (LOCAL) or from a REMOTE location; provided the receiver is equipped with the optional circuit card assembly A6AI. Refer to Paragraph 4-6 for remote operation.
h. BITE. The Built In Test Equipment (BITE) can be initiated at any time during operation of the Receiver to determine its operating condition. This self test may be initiated either from the front panel or from a remote location. From the front panel BITE is controlled through different combinations of pushbuttons; LOCK with AM, CW or LSB. LOCK and AM will initiate the self test with the front panel indicators indicating progression of the test. LOCK and CW continues the test after a fault is indicated while LOCK and LSB provides for returning to the operating mode from a BITE routine. To apply either of these three routines the LOCK pushbutton must be pressed first and while still holding it depressed press its companion pushbutton (AM, CW or LSB) then release the pushbuttons in the opposite order to which they were depressed. If a fault is encountered, during the self test, the BITE will stop and the frequency display will indicate a two digit number. The fault isolation procedures described in Chapter 6 of this manual list the probable errors corresponding to the two digit numbers. To continue the test after a fault is indicated, press the LOCK and the CW pushbuttons as described above. If it is desired to return to the operating mode while the BITE sequence is in progress, press LOCK and LSB in the manner described. Refer to Paragraph 4-6 for remote operation of BITE.

4-5. TYPICAL OPERATING PROCEDURES. Because of the different operating options associated with the receiver, and the different applications for the receiver, the detailed operating instructions may be different for each site. The following procedure describes a typical operating sequence for selecting a station in the local AM broadcast band.
a. Set the POWER switch to ON and if it is desired to initialize the Receiver, depress the LOCK pushbutton and while still holding the LOCK in depress the AM pushbutton, then release in the opposite order they were depressed.
b. Depress pushbuttons BWI through BW5 and observe the indicated bandwidth in the display. Select the desired bandwidth by depressing the appropriate pushbutton.
c. Depress the AM pushbutton and verify that the AM mode is indicated in the display.
d. Depress the ENTER pushbutton and the numeric keys to enter the frequency of the desired station. Note that the first digit entered will be the ten's MHz digit, the second will be the unit's MHz digit, and so on.
e. Select the desired AGC mode by depressing the appropriate pushbutton.
f. Depress the METER RF/AF pushbutton to display the AF scale.
g. Connect a set of headphones to the front panel PHONES jack and adjust the AF GAIN control to obtain a suitable audio level.
h. Depress the TUNE RATE pushbutton to select either FAST, SLOW, or fine tuning mode. Note that fine tuning rate is indicated by the absence of the FAST and SLOW indication.
i. Turn the front panel tuning knob to obtain the maximum signal strength as indicated by the meter and the audio level in the headphones.
j. Depress the LOCK pushbutton to disable the front panel tuning knob.
k. This completes the typical operating sequence for operation of the Receiver in the AM mode. Operating the Receiver in the remaining modes is similar to the procedure just described.

4-6. REMOTE OPERATION. If the Receiver is equipped with the optional A6A1 circuit card assembly it may be operated from a remote control device. Before attempting remote control operation, make sure the procedures described in Paragraphs 2-8b and 2-1 0 e have been incorporated.
a. The data character used for remote control is the standard ASCII asynchronous format consisting of a start bit, seven data bits (one ASCII character), an optional parity bit, and two stop bits. Each command message to the receiver must be terminated with the carriage return character (an additional line feed character is not required).
b. The remote operation of the receiver, as described in the following paragraphs, may be divided into four main functions: remote control of the receiver, override commands, receiver status, and receiver monitoring. Up to 10 separate receivers may be connected to the remote control device on the same bus. Status and monitor commands may be sent to only one Receiver at a time.
c. Receiver Control. The Receiver will respond to the following control commands:
(1) Receiver Address. The command $\$(N)$ will select a particular Receiver designated by ( N ) where N represents the Receiver number. That is $\$ 7$ will select Receiver number 7. More than one Receiver may be addressed at the same time by inserting commas between the Receiver numbers. For example, $\$ 7,23,8$ would select Receivers 7 , 23 and 8. (Multiple addressing is not allowed for status return request.)
(2) Frequency Selection. The command $\mathrm{F}(\mathrm{N})$ is used to select the Receiver operating frequency designated by ( N ) where N represents the Receiver frequency. The frequency command may specify the desired frequency down to 1 Hz . For example, the command F03.415926 would tune the Receiver to 3.415926 MHz . If an exact frequency is not required, the leading and trailing zeros may be eliminated. For example the command F3.4 would tune the Receiver to 3.400000 MHz . Note that in both cases a decimal point is required to indicate MHz .
(3) Mode Selection. The desired mode is selected by sending one of the following commands:

D1 AM Operation
D2 FM Operation
D3 CW Operation with variable offset, see Paragraph (4)
D4 CW Operation with zero offset
D5 ISB Operation (if ISB option is installed)
D6 LSB Operation
D7 USB Operation

## NOTE

To monitor status on controller refer to Paragraph e (4).
(4) BFO Offset. The BFO offset frequency may be set in the CW mode by sending the command $B(N)$ where $(\mathrm{N})$ indicates the offset frequency in kHz . For example, the command $\mathrm{B}+1.82$ will set the BFO offset 1.82 kHz above the center frequency; B-4.65 will set the offset 4.65 kHz below the center frequency.
(5) Bandwidth Selection. The desired bandwidth is selected by sending the command $\mathrm{I}(\mathrm{N})$ where $(\mathrm{N})$ indicates the filter bandwidth in kHz . For example, the command 13.24 would select the 3.24 kHz filter. If a command is received that does not match the filter in the Receiver, the Receiver will automatically select the closest filter. For example, the command 17 . could select a 6.8 kHz filter. Note that the decimal point is used to indicate kHz .
(6) Gain Control Mode. The desired gain control mode is selected by sending one of the following commands:

M1 Selects short AGC time constant
M2 Selects medium AGC time constant
M3 Selects long AGC time constant
M4 Selects manual gain control, see Paragraph (7).
MS Selects short AGC with preset threshold, see Paragraph (7).
M6 Selects medium AGC with preset threshold, see Paragraph (7).
M7 Selects long AGC with preset threshold, see Paragraph (7).

## NOTE

To monitor status on controller refer to Paragraph e (4).
(7) Manually Set Gain Control. The Receiver will respond to remote commands to set a manual gain. The Receiver IF gain is controlled by adding attenuation, gain controlled in 150 steps over AGC range of approximately 120 dB . Any position can be selected by sending number from 0 to 150 . This feature may also be used in conjunction with the AGC operation to establish a minimum threshold level for the AGC. The command for setting the level is $A(N)$ where ( N ) represents an amount of attenuation, to be added to the circuit. For example, the command A3 would send 3 to the AGC control circuit; A104 would send 104.
d. Status Commands. The status commands provide the remote controller with the status of a particular Receiver. Status commands may be sent to only one Receiver at a time since some of the status commands will result in monitor data from the Receiver back to the remote control device. The Receiver will respond to the following status commands:

```
S1 Set Receiver to local control
S2 Set Receiver to remote control
S3 Initiate BITE self test sequence, see Paragraph (1)
```


## Change 1 4-9

```
S4 Terminate BITE self test sequence, see Paragraph (1)
S5 Send bandwidth of installed IF filter, see Paragraph (1)
S6 Send BITE results, see Paragraph (1)
S7 Force bandwidth setup, see Paragraph (1)
S8 Enable remote AGC dump, see Paragraph (2)
S9 Inhibit remote AGC dump, see Paragraph (2)
```

(1) If the Receiver is in the override mode command S1 will be ignored. If the Receiver successfully completed the BITE self test sequence (command S3), sending S 6 will result in a response OK08 (CR). The carriage return (CR) indicates the end of the message. If the self test was not completed successfully, the Receiver might respond with something like; : 4, 17, 33, END 08 (CR). The colon will be sent at the start of the test. For each step of the test that fails, the Receiver will send the step number (as in the example steps 4 , 17, and 33 ) up to a total of 5 steps. The typical Receiver response for command SS would be L, 1.2, U, .4, , 16, ,(CR) indicating the type or bandwidth of filter installed in filter slot sequency (FL1, FL2, etc.). When no filter is installed digits will not appear between commas in the return information as shown in the example for filter slots FL5 and FL7. The S7 command is used only for maintenance and is described in Chapter 6.
(2) Status commands S8 and S9 are used for AGC dump control. When the Receiver is being operated remotely, these commands will be extremely useful to prevent "signal blasting" during tuning procedures. The S9 command, Inhibit remote AGC dump, will prevent the Receiver microprocessor from effectively causing the AGC to run wide open when remotely tuning in small frequency increments, e.g., 1 to 30 Hertz steps. If the AGC dump were not inhibited, the Receiver would run at full signal gain each time there was a 1 Hertz frequency change, causing high level audio signals to be sent over the voice links to the remote control area. This could result in possible link damage or severe crosstalk in the system. Upon completion of the fine tuning activity, the S8 command restores the AGC dump to microprocessor control. The S 9 functions will not be needed for large frequency increments, typically 1 MHz or greater.
e. Monitor Commands. The remote control device may command a particular Receiver to provide monitor data. The response from the Receiver will be in the following sequence: frequency detector, AGC mode, IF bandwidth, BFO frequency, IF attenuation, and status. All unnecessary information will be eliminated. For example, if the Receiver is in the AM mode, BFO data will not be included. The Receiver will respond to the following monitor commands:

G Receiver will respond with all relevant data
T Receiver will respond with specific data.
(1) The Receiver will respond to a T command according to the data specified in the command. For example, the command TFI will result in frequency and bandwidth response; TFD will result in frequency and detector mode response.
(2) The status data will be sent as the last item in each monitor response from the Receiver. The status data is a one or two digit number representing the following conditions:

| 0 | Receiver is operating in local control mode |
| :--- | :--- |
| 1 | Receiver is operating in remote control mode |
| 2 | Synthesizer is out of lock |
| 4 | Receiver is in the Override mode |
| 8 | Last command sequence has character transmission error |
| 16 | Last command sequence had data error |
| 32 | Lost Data error in last sequence |

(3) If two or more conditions are present the numbers representing the conditions will be added together and transmitted as one number. For example, $17(1+16)$ would indicate that the Receiver is in remote control (1) and last command sequence had a data error (16); $13(1+4+8)$ would indicate remote operation (1), override mode (4) and character transmission error (8).
(4) The Receiver will respond to a G command with a response reflecting all Receiver operating conditions. Let us assume that the Receiver address is 10 and the Receiver is set to 12.34 MHz , CW mode, BFO offset of +1.2 kHz , MAN AGC, with manual gain set with (A20), seeParagraph 4-6p (7), and Bandwidth of 6.8 kHz . The return information to the remote controller will be formatted as follows:

Data Stream to Receiver: $\$ 10 \mathrm{G}$ (CR)
Data Stream from Receiver: F12.34, D3, M4, I6.8, B1.2, A20, S1(CR)
It can be seen by analyzing the Status (S) data that the Receiver is operating in the remote control mode.
f. Override Operation. The remote control device may command a Receiver to switch to an override mode. In the override mode, some of the automatic operating features of the Receiver are disabled. That is, the first local oscillator is always tuned to 40.455 MHz above the entered frequency and the IF slot is always enabled regardless of the type of filter installed. Therefore, the remote controller must decide what the filter should be used for, the type of detection mode to employ, and the BFO offset to receive a signal. For instance, in sideband detection with a symmetrical filter, the first local oscillator and BFO must be properly offset to correctly demodulate the signal. The controller also assumes the responsibility for insuring that the filter is installed in the selected slot, since an empty slot will cause a dead Receiver. The override mode blanks the display except for remote indication. In addition, override signals cannot be handed off to the operator, since the machine has no way of deciding the difference between a sideband signal with virtual carrier offset or a CW signal with a BFO offset. The override mode is invoked when both detector and bandwidth are sent in the same command with the $=$ sign. For example $\mathrm{l}=4$ will select the filter in filter socket 4 instead of a filter at or near 4 kHz as when 14 command is sent. Sending either without the equals stores the = data but leaves the Receiver in the modal command mode. The Receiver will respond to the following override commands: The Receiver requires both $\mathrm{D}=(\mathrm{N})$ and $\mathrm{I}=(\mathrm{N})$ to go into override mode.
(1) Mode Selection. The desired mode is selected by sending one of the following commands:

| $D=1$ | Selects envelope detector |
| :--- | :--- |
| $D=2$ | Selects continuous wave detector |
| $D=3$ | Selects frequency modulation detector |
| $D=4$ | Selects ISB operation |

NOTE
To monitor status on controller refer to Paragraph e (4).
(2) Filter Selection. In the override mode the filters are selected according to the filter socket number instead of the filter bandwidth. The command is $\mathrm{I}=(\mathrm{N})$ where ( N ) corresponds to the filter socket number. For example, the command $\mathrm{I}=3$ will select the filter inserted in filter socket 3 . Override mode is disabled by sending a normal I or D command.

NOTE
To monitor status on controller refer to Paragraph e (4).
(3) Other operating commands (Frequency, AGC, Manual AGC, Attenuation and BFO) can be entered in the normal way.

4-7. SHUTDOWN PROCEDURES. The Receiver is shutdown simply by setting the POWER switch to the off position. When the Receiver is switched back on, it will automatically return it its last setting.

## Section III. EMERGENCY OPERATION

4-8. EMERGENCY OPERATING PROCEDURES. The emergency operating procedures applicable to the Radio Receiver R-2174(P)/URR are dependent on the particular operating mode being used. For example, if the Receiver is being operated in the remote control mode and a failure occurs in the remote control circuitry, it may be switched to the local mode and manually operated with the front panel controls.

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## CHAPTER 5

## THEORY OF OPERATION

5-1. INTRODUCTION. This chapter contains the theory of operation for the R-2174(P)/URR Radio Receiver. The theory describes the primary signal flow as it progresses through the various components of the receiver and explains the functional relationships of each component to the signal flow. The description or theory is divided into three main sections to best describe the receiver function. Section I describes the primary signal flow as related to the receiver's functional operation. Section II describes in detail each circuit card assembly and the components included on each card. Section III describes the functional operation of mechanical assemblies. Simplified functional block diagrams and timing diagrams are used throughout the text to aid the technician in understanding the various functions.

## SECTION I. FUNCTIONAL SYSTEM OPERATION

5-2. GENERAL. Section I provides a description of the functional operation of the Receiver. This description traces the primary signal from the antenna input, through the Receiver, to the audio output and describes secondary functions as they are related to that primary signal flow. To simplify the description and for a basic understanding of the receiver operation the functions are divided into five major divisions. These five divisions, as shown in Figure 5-1, are: primary signal (RF, IF and AF), oscillator synthesizers, Automatic Gain Control (AGC), receiver control, and power supply.
a. The primary signal function consists of circuit card assemblies A1, A2, A3, A4 and A5. This function describes the signal flow from the RF input at the antenna to the audio termination into headphones or speaker. The second major division describes the operation of the oscillator synthesizers (circuit card assemblies A7 and A8). This description includes the first local oscillator signal to the first mixer, the second local oscillator to the second mixer and the beat frequency oscillator function for CW operation. The third major division describes the operation of the Automatic Gain Control (AGC) which is contained primarily on circuit card assembly A4 along with IF and AF components. Some other AGC circuits are located on A3 and A5. The fourth major division describes the function of the receiver control circuitry which is contained on the front panel and circuit card assemblies A6A1, A6A2 and A9. The fifth major functional division of the Receiver is the power supply contained in assembly A 10 . Figure $5-2$ shows a simplified overall functional block diagram of the Receiver and should be followed in reading the description.

5-3. PRIMARY SIGNAL RF, IF AND AF. The primary signal consists of the radio frequency (RF) signal, the intermediate frequency (IF), and the audio signal (AF). The description that follows divides the primary signal into those three functions to more clearly present their operation.
a. RF Signal. The antenna signal is connected through the rear panel of the Receiver to a low-pass filter located on assembly Al. The four section elliptical low-pass filter rejects frequencies above 35 MHz and at the same time prevents local-oscillator and IF frequencies from being radiated back through to the antenna. Each section of the filter contains a coil-capacitor tank circuit, with the coil of each stage adjustable for peaking the tank circuit. This provides optimum rejection at its designed frequency. The output of this first low pass filter is routed to another filter located on circuit card assembly A2. This filter operates much in the same manner as the one just described but also provides for impedance matching to the first mixer stage and to reduce peak to peak ripple on the carrier signal. The output of this filter is connected directly to the first mixer where the RF signal is mixed with the variable frequency from the first local oscillator to form the first IF signal.
b. IF Signal. The first IF signal is developed in the first mixer stage where the RF signal is mixed with the variable output frequency of the first local oscillator. The first local oscillator frequency is varied in direct relation to the RF frequency selected from the control section and varies between 40.955 ( 0.5 MHz selected) and $70.454 \mathrm{MHz}(29.999 \mathrm{MHz}$ selected). This variable frequency produces a difference frequency in the mixer of 40.455 MHz . This difference frequency along with all other resultant frequencies is routed to a filter which rejects all other frequencies except the desired 40.455 MHz carrier with intelligence. The filter has a 20 kHz bandwidth and is coupled to a linear buffer amplifier. This amplifier stage is controlled by a control signal from the AGC circuit to maintain a constant amplitude output.
(1) The AGC controlled IF signal is routed to circuit card assembly A3, where it is connected to a two stage IF amplifier, with additional between stage filtering. These two stages provide additional level control of the IF signal. This AGC controlled IF signal is impedance coupled to a band pass filter consisting of four stages of tuned filter traps. An adjustable amplifier just prior to the filters provides for adjusting the gain through the filters. The IF output from the filter is connected to the second mixer where the IF is mixed with the second local oscillator to form the second IF signal. The second local oscillator output of 40 MHz is mixed with the 40.455 MHz first IF , which provides a difference frequency of 455 kHz . This difference frequency is transformer coupled to a filter network which rejects all frequencies except the desired 455 kHz second IF signal. The output of the filter network is coupled to an IF amplifier to restore gain and for coupling to the plug-in band pass filters, contained on circuit card assembly A4.
(2) The IF signal from A3 is coupled directly to seven plug-in bandwidth filter slots. The number of filters plugged in and the bandwidth of each filter depends on the option of the operator. Any number, up to seven, may be plugged into the card at any one time. The filter bandwidth desired for operation is then automatically switched into the IF circuit when it is selected from the front panel or remotely. When the optional independent sideband (ISB) is used, filter slot FL1 must contain a lower sideband filter. All bandpass filter slots are permanently connected to a diode switch, including FL1. This filter may be connected, through a movable link, from the diode switch to a bus that leads directly to the ISB circuit card assembly A5.
(3) In all modes of operation, including ISB, the diode switch selects the desired filter slot. The control circuitry automatically selects the filters in an ascending bandwidth order regardless of the order in which they are plugged into the sockets; that is, bandwidth I (BWI) selects the narrowest bandwidth and so on with BW5 selecting the widest bandwidth. Two slots are generally reserved for upper sideband (USB) and lower sideband (LSB) which are also automatically selected when that mode is directed from the control section. The IF signal output of the selected bandwidth filter is impedance coupled to a two stage AGC controlled amplifier. The output of this amplifier is then coupled to a bandpass filter for additional filtering of the IF signal. A portion of this signal is routed through a buffer amplifier to J2-IF OUT, on the rear panel, as the IF output signal.
(4) The IF signal is also coupled through a buffer amplifier to one input of an RF switch and to the product and synchronous AM detector. The RF switch also has the beat frequency oscillator (BFO) as an input with the output of the switch coupled directly to the FM detector. The switch has two modes of operation and is switched through the control circuitry. In the AM and FM modes the switch selects the IF signal, in all other modes, the switch automatically couples the BFO to the FM detector and limiting amplifier. Two outputs of the FM detector are then coupled to a detector select switch and to the carrier input of the product and synchronous detector. The output to the product and synchronous detector will be either the demodulated AM signal or BFO depending on the mode selected. The detector select switch is also controlled by the function modes of the control section. In the FM mode the select switch passes only the FM detected signal to an audio filter. In all other modes the detector select switch passes the output of the product and synchronous detector to the same audio filter.


| Circuit Card |  |  |
| :--- | :--- | :--- |
| Assembly | Nomenclature | Function |
| A1 | RF Low Pass Filter | Primary Signal |
| A2 | First IF Mixer | Primary Signal |
| A3 | Second IF Mixer | Primary Signal |
| A4 | Main IF/AF | Primary Signal/AGC |
| A5 (optional) | ISB IF/AF | ISB-Primary Signal |
| A7 | First LO Synthesizer | Oscillator Synthesizer |
| A8 | Second LO/BFO Synthesizer | Oscillator Synthesizer |
| A6A1 (optional) | Serial Asynchronous Interface | Receiver Control |
| A6A2 | Microcomputer | Receiver Control |
| A9 | Front Panel Control | Receiver Control |
| A10 | Power Supply | Receiver Power |

Figure 5-1. Basic Receiver Functional Breakdown
(5) When the ISB is selected from the control section, the LSB portion of the IF signal is linked directly to circuit card assembly A5. The flow of the IF signal through the ISB circuit is very similar to that just described; except that the BFO is connected directly to the ISB detector. The USB portion of the IF signal is routed through the A4 circuit card in the ISB mode.
c. AF Signal. The detected audio signal from either the FM detector or the product detector is selected by the detector select switch through receiver control. The selected audio is routed through a lowpass filter to a crosspoint switch. Through receiver control the crosspoint switch selects the various audio modes available as outputs from the Receiver. When the ISB option is installed, the audio output from that circuit card is also coupled to the switch. In non ISB modes the main audio (A4 card) is selected and routed to two separate audio amplifiers. The first amplifier is volume controlled through the AF GAIN control on the front panel. The output of this amplifier is routed both to the rear panel for loudspeaker output and to the phones jack on the front panel. The second amplifier is level controlled through a variable attenuator placed in the line to the second amplifier by the crosspoint switch. Attenuation is varied through a screwdriver adjustment on the front panel. This amplifier then drives an output transformer which provides the monitor line output to the rear panel. If the ISB circuit card is installed in this non ISB mode, the output of the attenuator will also be routed through circuits on the ISB card and appear on Line 1 output on the rear panel. This circuit is described in the ISB mode which follows.
(1) In the ISB mode either the main (USB) or the ISB (LSB) is selected and routed to the same circuits as described in non ISB mode. In addition, the crosspoint switch couples both the LSB and USB through variable attenuators to their respective amplifiers on the ISB circuit card. The two amplifiers drive output transformers which couple the USB (Line 1 output) and LSB (Line 2 output) to the rear panel. Level control of Line I output is through front panel screwdriver adjustment MAINLINE LEVEL. Line 2 output is controlled through I-LSB LINE LEVEL. These two adjustments vary the attenuators connected in their respective lines.

5-4. OSCILLATOR SYNTHESIZERS. The oscillator synthesizers consist of the first local oscillator (LO) synthesizer, the second LO synthesizer and the beat frequency oscillator (BFO) synthesizer. The three oscillator synthesizers are each independent separate oscillators except that the 1 MHz signal derived from circuitry in the second LO is used as a reference frequency to the other two oscillators.
a. First Local Oscillator Synthesizer. The first local oscillator provides the oscillator frequency to the first mixer where it is mixed with the RF signal to produce the first IF signal. This variable oscillator, located on circuit card assembly A7, is controlled from the receiver control frequency select. It is a voltage controlled single loop synthesizer oscillator with an output frequency variable between 40.955 and 70.454999 MHz in 1 Hz increments.
(1) The voltage controlled oscillator (VCO) generates the basic frequency which is applied to a drive amplifier located on circuit card assembly A2. This same output frequency is applied to a divide by N circuit. The value of N is determined by the digital control logic, also coupled to the divide by N circuit. This digital control logic depends on the RF frequency selection inputs. The output of the divide by N circuit is coupled to a phase comparator to which a 100 kHz reference signal and the digital logic is also connected. The basic output of the phase comparator depends on the phase difference between the reference signal and the divide by N signal. This basic output is combined with the digital control logic, filtered and applied to the dc control amplifier. The output of this dc amplifier controls the frequency of the VCO. The output of the VCO is transformer coupled to a high pass filter, located on circuit card assembly A2, which rejects frequencies below the oscillator range. This filter output is applied to a transistor drive amplifier which routes the signal through an RF wideband transformer to the mixer. The VCO is varied between 40.955 MHz (RF selection of 0.5 MHz ) and 70.454999 MHz (RF selection of 29.999 MHz ). When this oscillator frequency is applied to the mixer and mixed with the RF signal a difference frequency of 40.455 MHz is obtained. It is this difference frequency that is used for the first IF signal.
b. Second Local Oscillator Synthesizer. The second local oscillator provides the oscillator frequency to the second mixer where it is mixed with the first IF signal to produce the second IF signal. The oscillator, located on circuit card assembly A8, is a constant frequency phase locked oscillator driven from a reference signal. The output of the oscillator is 20 MHz which is frequency doubled to provide the 40 MHz signal to the second mixer. A reference signal, either the internal reference signal or an external reference signal, is required for operation of the oscillator synthesizer. The internal reference signal comes from a crystal oscillator which has an output frequency of 5 MHz . An external reference signal may be applied through the REF IN/OUT connection on the rear panel. The REF INT/EXT switch, also located on the rear panel must be in the applicable position as to the reference selected. With the switch in the EXT position, the internal reference oscillator is turned off and the external reference signal is applied through a transistor switch and shaper circuit to a phase comparator. With the switch in the INT position, access to any external reference is turned off, the internal oscillator is turned on and applied through the transistor switch in the same manner as the external reference.
(1) The heart of the oscillator synthesizer is a 20 MHz crystal referenced oscillator whose output is coupled through a buffer amplifier to a divide by 2 circuit which in turn feeds another divide by 2 and a divide by 10 circuit. The resultant $10 \mathrm{MHz}, 5 \mathrm{MHz}$ and 1 MHz output of the circuit is coupled to a data select circuit. The 1 MHz signal is also routed to the BFO oscillator and to circuit card assembly A7, where it is used as a reference signal to the first local oscillator. The reference frequency output of the data select circuit is applied to the phase comparator which phase compares this signal to the internal or external reference described above. The phase difference signal (if any) is then applied through a digital to analog converter to the 20 MHz oscillator. If the oscillator tends to drift off frequency the phase difference between the reference signal and oscillator signal will be detected by the phase detector and the phase difference, applied through the digital to analog converter, will readjust the oscillator.
(2) The output of the oscillator is also coupled through another buffer amplifier to a frequenq doubler. The 40 MHz output signal from this doubler is routed to circuit card assembly A3 and capacitor coupled to the second mixer. The difference frequency between this 40 MHz signal and the 40.455 MHz first IF signal is the second IF signal of 455 kHz .
c. BFO Synthesizer. The BFO provides the fixed and variable beat frequency for reinsertion of the carrier in the sideband and CW modes. The oscillator operates at the second IF signal frequency of 455 kHz and can be varied 8 kHz in either direction. The BFO oscillator, located on circuit card assembly A8, is a voltage controlled variable oscillator with a center frequency of 22.75 MHz . The oscillator output frequency is filtered and applied through a buffer amplifier to a divide by 50 circuit. The output of this divide by 50 circuit provides the variable 447 to 463 kHz BFO signal for reinsertion at the product detectors on circuit card assemblies A4 and A5.
(1) A digital control circuit is used to vary the basic oscillator frequency of 22.75 MHz . The output of this circuit is coupled to a phase comparator, along with a 500 Hz reference signal. This 500 Hz reference signal is derived by applying the 1 MHz reference signal, from the second local oscillator, to a divide by 2000 circuit. The phase comparator compares the output of the digital control circuit and the 500 Hz reference signal and applies the difference signal through a digital to analog converter to the VCO. The oscillator may be varied between 22.35 and 23.15 MHz in 500 Hz increments, which when applied to the divide by 50 circuitry provides a BFO frequency between 447 and 463 kHz variable in 10 Hz increments. The BFO frequency is filtered before being routed to the product detectors.

5-5. AUTOMATIC GAIN CONTROL (AGC) . The AGC circuits provide the receiver with constant level AF output signal with large variations in the incoming RF signal. For example, the change in IF or AF output levels is less than 6 dB for a change in the input level of -100 dBm to -10 dBm . This automatic gain control is accomplished through AGC circuitry, located mainly on circuit card assembly A4. The optional ISB circuit card assembly A5 contains its own AGC for signal gain control in the ISB mode of operation.
a. The receiver may be operated in any one of three different gain control modes: manual, automatic, and automatic with a manually set threshold. In the manual mode the gain is set through a front panel control. In the automatic mode the AGC circuits will compensate for changes in the receiver input signal level. In the automatic/manual mode the front panel control is used to set the operating threshold of the AGC circuits.
b. The AGC operates from a portion of the IF signal taken after the gain controlled IF amplifier stage, thus maintaining a closed AGC loop. The AGC circuitry detects the IF signal, provides three different decay times, provides for automatic threshold control of the output signal or manual threshold control. The output of the AGC circuit is routed to the second IF amplifier on circuit card assembly A4 and to a current amplifier on circuit assembly A3. The current amplifier controls the first IF signal by controlling the gain of two IF amplifier stages; one located on circuit card assembly A3 and the other located on circuit card assembly A2.

5-6. RECEIVER CONTROL. Signals that control the receiver's operational parameters (such as operational mode, receiver frequency, BFO frequency, bandwidth, AGC and BITE sequence) are produced by the microcomputer (A6A2) and routed by the front panel receiver control circuit card assembly (A9). The microcomputer under program control follows instructions it receives from the front panel in LOCAL operation and from the remote controller in REMOTE operation. In both LOCAL and REMOTE, the microprocessor functions essentially the same; however, when being operated from a remote location, the power must be turned on at the front panel and the optional serial asynchronous interface assembly (A6A1) must be installed. The serial asynchronous interface circuit card assembly (A6A1) interfaces the external remote controller (when used) with the microcomputer.
a. Figure 5-4 shows the receiver control circuits and shows the signal flow between the front panel receiver control assembly (A9), the microcomputer circuit card assembly (A6A2) and the serial asynchronous interface circuit card assembly (A6A1). The front panel receiver control assembly (A9) contains the Liquid Crystal Displays (LCD), connects to both sets of keyboard switches and the receiver control circuits and routes data between these units and the microcomputer. The R-2174(P)/URR Radio Receiver interconnection diagrams contained in Chapter 8 show all connections to and from these modules.
b. As shown in Figure 5-2, the microcomputer (A6A2) directs receiver operations by interfacing with circuits on the receiver control assembly (A9) and serial asynchronous interface (A6A1) and sending control signals to various receiver circuits.
c. The receiver control assembly (A9) contains the tuning mode and frequency displays and the tuning control circuits and connects to the frequency select and the modes select keypads. The strobes and selection circuits that route data under microprocessor control are also on the receiver control assembly. The strobe and selection circuits control the transfer of information between the front panel and the microcomputer as well as between the microcomputer and the receiver circuitry.
d. The basic functions performed by the microcomputer (A6A2) include:

1. Initialize circuits following power application.
2. Read local input signals from front panel controls.
3. Update front panel displays.
4. Compute and send receiver tuning and operating data to the appropriate receiver circuits.
5. Receive commands from the remote controller and upon request, return receiver status.
6. Retain memory of receiver setting at power failure or turn-off and restore receiver to the operational modes when power is reapplied.
e. The microcomputer directs receiver operations by executing the control program that it obtains from the Erasable Program Read Only Memory (EPROM). During local operation, the receiver is controlled by the microcomputer as follows:
(1) The frequency and mode setting established by the front panel switches and the tuning control are continuously read at 25 ms intervals by the microcomputer. The actual scanning function of the microcomputer is under program control. As the microcomputer scans the front panel switch and controls, it also stores the current status of each control parameter in the memory (RAM).
(2) The microcomputer, again under program control, uses the stored control parameters that it placed in memory to compute the control signals that it sends to the receiver circuits (1st LO Synthesizer, 2nd LO and BFO, and main IF/AF). These digital control signals are then sent to the receiver circuits through the receiver control assembly (A9) to generate the desired operation.
(3) Periodically, as established by the control program, the microcomputer reads the receiver status that it has stored in memory (RAM) and sends this information to control the front panel displays.
(4) When the microcomputer senses a BITE (Built In Test Equipment) request from the front panel switches, it is directed to perform the BITE test sequence and follows the BITE program which is also contained in memory. During BITE sequence, the processor disables all external and local controls.

## NOTE

The receiver control program and the BITE program have been developed by the manufacturer as part of the receiver design and cannot be changed or updated by the customer for either operational control or maintenance.
f. During Remote operation, the Receiver is controlled by the microcomputer, but in place of instructions from the front panel, the instructions to the microprocessor are obtained from the remote controller:
(1) During the front panel scan, the microprocessor monitors the position of the LOCAL-REMOTE switch. When this switch is in REMOTE, the microprocessor will branch to the remote mode portion of the program so that instead of responding to front panel switches, it will look for command words from the remote interface card (A6A1).
(2) Commands from the remote controller are received by the Serial Asynchronous Interface Assembly A6AI. As each command is received, an interrupt is sent to the microcomputer which directs the microcomputer to branch to specific portions of the program (or subroutine) to carry out the command. After responding to the command instructions, the microcomputer returns to the normal remote mode program until a new set of commands or requests are received from the remote controller.
(3) When the remote controller contains data for receiver control (such as frequency, AGC, or mode selection) the microcomputer stores the data in memory (RAM), and on completion of remote data interrogation transfers the information to the receiver circuits.
(4) When the remote controller command contains a request for receiver status, the microprocessor accesses the corresponding receiver status information stored in memory (RAM) and sends it via the Serial Synchronous Interface Assembly (A6AI) to the Remote Controller.
(5) In addition, the microcomputer under program control periodically (every 25 ms ) reads the receiver parameters from memory (RAM) and continually updates the front panel displays.
g. When a request is received from the remote controller during LOCAL operaton, the microprocessor will respond and return the status of the receiver as described in the above paragraphs. Remote commands received during LOCAL operation will be stored in memory but they will not be acted upon unless the receiver is placed in REMOTE operation.
h. The microcomputer directs all operations and communicates with other receiver control circuits through its 8 bit bi-directional data bus and the write/read and clock (ROMC) lines. The ROMC lines indicate the type of instructions to be performed with the write and clock lines providing the necessary information. The 8 -bit bus provides both bidirectional data and unidirectional address capability (to the receiver). The operating control program is contained in the program memory EPROM (Erasable Program Read Only Memory). Temporary storage for receiver settings and for data computations is provided by the working Random Access Memory (RAMs), which can be written into and read out of by the CPU. These memories are addressed by the CPU, through the static memory interface (SMI). The CPU sends the ROMC, write and clock signals to the SMI. The SMI recognizes the ROMC code calling for a Memory Address operation.
The SMI, in sequence, addresses the EPROM or RAM over the memory address bus. Then it sends a read signal to the EPROM (if it is addressed) or sends a read or write signal to the RAM (if it is addressed). The CPU places the data to be read by the RAM on the data bus or reads the data placed on the data bus by the EPROM or RAM, as appropriate.
i. The microcomputer (A6A2) also contains the RESET and RAM data retention circuitry. The RESET circuitry generates reset signals when power is applied and turned off. This reset signal is applied to the CPU. When power comes on, the CPU initializes all circuitry to its starting condition and causes the program to start at its initial program address.
j. When power is removed, due to power turn-off or power failure, the reset signal to the CPU goes low. The CPU now causes the system to come to an orderly halt. In addition, the memory data retention circuitry (at power turn-off) connects an internal battery to the RAMs so that the receiver settings are retained in this memory. Thus, when power is reapplied, the receiver will be reset to its last operational condition when power was interrupted. Also, at power turn off, the memory retention circuitry places the RAM in a low power drain mode which retains memory but draws a minimal amount of power from the internal battery. When external power is applied, the internal battery is charged by the external power supply.
k. The CPU, in programmed sequence, receives and sends data from and to the front panel displays and controls, and the receiver control circuitry (through A9) via the CPU data bus.
The data is directed between the CPU data bus and the receiver control and front panel circuitry, in the correct program sequence, by the bi-directional, tri-state switch. This switch is controlled by the strobe logic which is driven by the ROMC, write and clock signals from the CPU. The addressing of the various receiver and front panel circuitry, to accept or supply data from or to the data bus, in the prescribed program sequence, is done by the strobe logic and tri-state latched switch. This switch is driven by the CPU data bus and its outputs latched to the input data from the bus, at the prescribed program times, by the strobe logic,
I. The CPU receives and sends data from and to the remote controller via the Serid Asynchronous Interface (module A6A1). Data to and from the CPU and module A6A1 are sent directly over the CPU data bus. Additional control signals between the CPU and the module A6A1 are sent via the CPU I/O (input-output) ports. A UART (Universal Asynchronous Receiver Transmitter) in the A6A1 module interfaces the parallel 8-bit data word on the CPU bus to the serial data streams in and out from and to the remote controller. The UART also generates and sends to the CPU an interrupt request whenever the remote controller sends commands or data. This interrupt causes the microcomputer to orderly stop its present program and jump to a program routine which will service the remote controller commands and data. The UART and interrupt logic on the A6A1 module will supply the CPU with the interrupt routine starting address by placing this interrupt vector address on the CPU data bus when the CPU acknowledges that it will service the interrupt request. The reset signal from the microcomputer module is also sent to the interface module to initialize its circuitry at power turn-on.
m . As indicated in the block diagram Figure 5-2, the front panel receiver control, module A9, connects to the keyboard switch panels 1 and 2. These switches are continually read, in program sequence, by the CPU. The switches are read in groups, with the switches being selected by enabling their associated data select buffers. The data select and strobes circuitry selects the buffers, in the programmed sequence, as directed by the strobe logic and tri-state latched switch outputs from the microcomputer A6A2. The read data goes on to the buffered data bus and then through the data buffers and the bi-directional, tri-state switch to the CPU bus. The digital outputs from the tuning knob encoder and digital data outputs from the IF/AF (A4 module) are also read out the same as the switch data.
n. Both the tuning mode and frequency indicating front panel Liquid Crystal Displays (LCD) are contained on module A9. These displays are continually updated by the microcomputer CPU. The displays are driven by decoderdrivers which are enabled by their respective data select and strobes. The data from the CPU bus is applied to this circuitry, at the correct times in the program sequence, through the data buffers and bi-directional tri-state switch.
o. The operating and tuning data developed by the microprocessor is relayed to the appropriate receiver circuitry through module A9. The buffered data bus goes directly to the main IF/AF module, A4, with the data select and strobes circuitry supplying the strobes to the various circuits in this module. Data to the 1 st LO and 2nd LO/BFO modules are generated, in the proper program sequence, through the data select and strobes circuitry in the A9 module. Data to the I st LO consist of a strobed and clocked serial data stream. The 2nd LO/BFO data consist of a binary coded digital word for the BFO frequency and an ON-OFF signal for the BFO. It should be noted that OOL (Out of Lock) signals from the 1st LO, REF and BFO drive the OOL indicators located on the A9 card. The OOL overall signal, generated by the microcomputer, is sent to the front panel OOL indicator through A9.
p. The serial asynchronous interface module, A6A1, in addition to the UART and interrupt circuitry described earlier, contains circuitry for selecting and generating serial data mode and baud rates and for setting the receiver address when the receiver is used in the remote mode.

5-7. BUILT IN TEST EQUIPMENT (BITE). The BITE system of the Receiver has the ability to perform two major functions. First, it determines, organizes and displays the bandwidth of the IF filters installed in the Receiver. This allows the installation of these filters in any slot, with minimum limitations. Secondly, the BITE performs tests of functionality of the receiver modules. These tests provide overall verification of the operation of the Receiver, and specific verification of the operation of selected modules.

## a. The BITE system performs the following functional tests in the Receiver: Readability of the RAM in the

microcomputer; lock condition and timing of all phase locked loops; settling time in the frequency synthesizers; operation of the IF AGC; operation of the ISB AGC detector if it is installed; and measuring the bandwidth of all of the IF filters. In order for the Receiver to pass the above tests, all modules must have been operating properly; except for the A1 Lowpass Filter, A6A1 Remote Interface, and the audio portions of A4 Main IF/AF and A5 ISB (if installed) which are not interrogated during the BITE sequence.
b. BITE may be controlled by the operator from the front panel through the simultaneous use of two controls, one on each keypad. This minimizes the likelihood of accidental interruption of an operating Receiver. To initiate BITE from the front panel, the operator must press both the LOCK and the AM pushbutton switches, then release AM and then LOCK for initialization. The microcomputer then begins the full BITE sequence. As the microcomputer is executing the BITE program, the front panel display indicates what the Receiver is actually doing. For instance, frequency, BFO and mode data are displayed. Should an error be discovered the front panel frequency display is blanked except for a two digit error code which contains the number of the test that failed. The operator writes down the number of the test that failed then pushes both LOCK and CW and then releases CW and then LOCK, which tells the microcomputer to continue to the next test. The displays are reinstated and the testing proceeds. The displays remain active throughout the tests since the process takes approximately one minute to complete, and if the displays are moving, the operator has confidence that the tests are proceeding. When the test is finished, the Receiver will return to the signal it was monitoring prior to being told to perform the BITE test sequence. Should the operator wish to terminate the BITE cycle at any time, he may press and hold LOCK and LSB then release LSB then LOCK and the Receiver will revert to its normal pretest operation at a point in the sequence where disabling the process is allowed. In addition to being able to originate the test, to continue the test, and to stop the test, there is (for maintenance purposes) a loop facility which can be invoked by using both LOCK and USB pushbuttons (refer to section 6 of this manual). In this mode, the front panel controls, with the exception of LOCK and USB, are disabled but the microcomputer will continuously supply the signals required to perform the failed test so that additional fault isolation procedures, using external ' test equipment, may be employed.
c. The BITE sequence may-also be initiated by-a remote device. The remote control device sends a message to the Receiver telling unit to initiate BITE. BITE will then report up to a maximum of five errors and then report a test complete code upon request. If the test complete code is received with no fault numbers, then the receiver passed the BITE tests. The control device may send a message to-the Receiver asking for the installed IF filter bandwidths: The Receiver would then send the measured bandwidths to the controller showing the filter slot positions in which each filter is installed.

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d. During initialization, BITE is initiated if the microcomputer determines that the memory has been corrupted for some reason. This automatically initiated BITE will determine the receiver filter complement, organize a filter assignment table, and return the Receiver to operation. When performing this test, the Receiver does not stop on errors but completes the testing and restores receiver operation. The Receiver also does not stop on errors when BITE is initiated under remote control, but completes the sequence and stores the fault numbers for remote interrogation. The following discussion describes the actual tests conducted to verify the proper operation of each module.
(1) Microcomputer. A11 256 bytes of the Random Access Memory (RAM) in the microcomputer are tested to ensure that the memory may be written into and read out of correctly. A ones and zeroes memory pattern is used to perform this test.
(2) Second Local Oscillator and BFO. The lock status of both the BFO and the second local oscillator synthesizers is tested. The second local oscillator synthesizer should be locked at all times. The BFO is enabled by placing the receiver in the CW mode. The BFO synthesizer is then tested for a lock condition both at 455 kHz and all 500 Hz steps between and including plus and minus 8 kHz offsets. This dynamic program also checks the switching times of the BFO synthesizer as indicated by the out-of-lock circuitry on the BFO board. A failure in any of these tests is indicated as an error code to the operator or to the remote controller, on request.
(3) First Local Oscillator. Testing of lock, in 500 kHz intervals from 29.999 MHz to 0 MHz , is performed, and the switching time of the synthesizer throughout the band is tested utilizing out-of-lock signals as the indication for reaching lock after each step.
(4) IF Module. The IF module (A4) is checked to determine whether or not the AGC circuits operate properly on signals. The test routine enables the CW detector and BFO, and checks that an audio output is present during the initial filter tests. The manual IF gain attenuation system is tested by observing its control effects on the audio output level during the filter tests.
(5) ISB Module. When the optional ISB board (A5) is installed it is checked for proper AGC action along with the IF gain and the manual IF gain control, in the same way that the main IF board (A4) was checked.
(6) IF Filters. One of the major functions of the BITE is to determine the bandwidths of the IF bandwidth filters installed in the Receiver, and assign them to bandwidth selection switches BW1 through BW5. The filters are assigned in order of increasing bandwidth and allow the operator to select a desired bandwidth. In addition, two different types of single-sideband detection filters may be used and the bandwidth determination routines verify their correct installation. Two sideband filters may be installed in the Receiver for independent or normal sideband operation, one for upper sideband, the other for lower. (When ISB is installed these two filters are required.) If one symmetrical sideband filter is to be used for both sidebands it may be installed in filter slot FL1. Then for lower or upper sideband, the first and second local oscillators are offset by 1.8 kHz to accommodate the symmetrical filter. There is one restriction on the filter complement in the Receiver; the FL1 slot must contain the filter to be used for lower sideband. It may be either the independent offset sideband filter or a symmetrical filter, but in either case it must be the filter used for lower sideband. The BITE bandwidth routine checks the filter in the FL1 slot to determine whether or not it is a center-tuned filter or an offset filter. If center tuned, it uses the filter for both sidebands by putting a 1.8 kHz offset in the first and second local oscillator. If it is an offset filter, the filter slot that the filter is in is labeled for the lower sideband and the rest of the filter complement is searched for the matching USB filter. If a symmetrical filter is found to be installed in the FL1 slot, the offset for the first and second local oscillators is set to 1.8 kHz ; however, if an offset sideband filter is found with a symmetrical filter in FL1 slot an error will occur. If there are less than 5 symmetrical filters installed in the remaining slots
of the Receiver, the symmetrical filter in FL1 is used as a center tuned filter which may be used for the AM, FM, or CW detection mode. If five other filters are present in the system, the symmetrical sideband filter will be used only for sideband reception. The remaining filter slots, FL2 through FL7, are checked for presence of a filter, except for a slot which has previously been identified as the USB slot. If there is a filter present in a slot, its bandwidth is measured. The first LO is scanned, in frequency, from a 10 kHz maximum offset back to the 3 dB point, based on a previously measured center frequency reference level. (NOTE: This reference level can vary from filter to filter.) The frequency difference between the center referenced level and the 3 dB point is designated as half the actual bandwidth. Having measured the bandwidth of all of the filters installed, the filters are sorted in order of increasing bandwidth so that when the BW1 pushbutton switch on the front panel of the receiver is pressed, the narrowest bandwidth available is selected. When the BW2 pushbutton switch is pressed, the next narrowest progressing to the widest at BW5 (if there are less than 5 symmetrical filters installed in slots FL1-FL7 the widest one is assigned a number corresponding to the maximum number of symmetrical filters installed). Future filters which may have different bandwidths from the fifteen presently defined may be used in this system without any change. The bandwidths of the filters installed in the Receiver can be reported to the remote controller upon command. In the remote operating mode, if the remote controller asks for a specific bandwidth, the Receiver selects the bandwidth that is nearest to the bandwidth requested. The Receiver reports the actual bandwidth used to the remote controller. This is an indirect check on the filters.

5-8. POWER SUPPLY. The power supply provides the various dc voltages required throughout the Receiver. The unit, located on assembly A10, contains a step-down transformer, diode rectifiers, filter capacitors, regulators, and an alternate V2 amp 250v fuse. Primary input power is controlled through a POWER ON switch located on the front panel. This primary input power may be either 100,120 , 220 or 240 Volts from 43 to 420 Hz . The proper voltage is selected through a card select switch located on the rear panel of the Receiver. A stepdown transformer provides three different voltages for rectification, filtering and regulation. Six different dc voltages are provided at the output of the power supply. These voltages are +20 Vdc regulated, +15 Vdc regulated, +15 Vdc unregulated, -15 Vdc regulated, +5 Vdc regulated and +5 Vdc unregulated.

## SECTION II. FUNCTIONAL (DETAILED) OPERATION OF ELECTRONIC CIRCUITS

5-9. GENERAL. This section provides a detailed description of the functional operation of all electronic circuits contained in the Receiver. Refer to Figure 5-2 for an overall functional block diagram. Figure 5-1 lists the circuit card assemblies in the order that they are described in the following paragraphs. Simplified functional block diagrams along with timing diagrams are used throughout the text to aid in simplifying the description. Components referred to throughout the text are referenced by their last two reference designators; such as A4C6 is called out in the text as C6. Block diagrams referred to and included throughout the text should be used in conjunction with applicable schematic diagrams in Chapter 8.

5-10. RF LOWPASS FILTER, A1. The incoming RF signal is passed from the RF IN connector J1 to a 50 ohm, 4 section elliptical lowpass filter which has a cut-off frequency of 35 MHz (refer to schematic diagram, Figure 8-2). This filter provides the necessary protection to the Receiver from image signals at frequencies between 81.4 and 111.4 MHz ; and from signals at the first intermediate frequency of 40.455 MHz . The filter also prevents first local oscillator reradiation from the antenna connection. Each section of filter consists of a tank circuit, consisting of a tunable coil (L1 through L4) and capacitor (C2, C4, C6 and C8), connected sequentially in the receiver line with a second capacitor (C1, C3, C5, C7 and C9) connected from each tank circuit parallel to the signal flow. Each tank circuit is tuned to provide a high resistance to a particular frequency while other frequencies are reflected in the parallel capacitor which in turn reflects this signal to the next stage and so on.

5-11. FIRST MIXER, A2. Figure 5-3 is a simplified block diagram of the first mixer module A2. It consists of a signal lowpass filter, first mixer, bandpass filter, first IF amplifier and drive amplifier with its associated filters. The function of this module is to convert the incoming RF signal to the first intermediate frequency of 40.455 MHz , by mixing with the first local oscillator frequency of 40.955 to 70.455 MHz . The schematic diagram for the first mixer is shown in Figure 8-3
a. RF Signal Lowpass Filter and Mixer. The output of the AI module is connected to the first mixer through a two section elliptical lowpass filter, (L15-C23, L11-C24-C25 and L12-C26-C27), which has a cut off frequency of 35 MHz and serves to present a defined impedance to the mixer, U1, RF input port. This filter operates much in the same manner as the RF input filter described under Paragraph 5-10 above, except the first input coil L15 is non-adjustable. It offers a very low impedance to the incoming wanted RF signal, but an increasingly higher impedance to frequencies above 29.999 MHz . The mixer, U1, is used for mixing both the incoming RF signal ( 0.5 to 29.999 MHz ) and the first local oscillator signal ( 40.955 to 70.455 MHz ). The resultant frequencies are taken from the mixer, and filtered to provide a difference frequency of 40.455 MHz to form the first IF signal.
b. First Local Oscillator Input Filter and Drive Amplifier. The filter for the incoming first local oscillator signal is comprised of four sections of a tunable coil, (L1 to L4) connected in series with a capacitor, (C2, C4, C6 and C8) and with the combination of the two connected parallel to signal flow. A second capacitor, (C1, C3, C5 and C7) connected in series with signal flow separates each coil-capacitor. Each coil is tuned for maximum impedance for a desired frequency. The series capacitance acts as a high impedance to undesired frequencies below 40 MHz . The output of this filter is coupled to a common emitter amplifier Q2 through capacitor C9 and resistor R3. The mixer drive amplifier is comprised of transistors Q1, Q2, Q3 and Q4. The local oscillator signal from the filter, which may be monitored at TP2, is coupled to the base of common emitter amplifier, Q2, whose current is regulated by transistor Q1. The voltage at the base of Q1 is set by divider R1 and R2 which in turn sets the potential at the emitter of Q1. Thus the current through R6 is regulated by bias control of Q2 via R6 and L6. The output of Q2 is capacitance-coupled (C21 and C22) to a complementary pair amplifier made up of PNP transistor Q3 and NPN transistor Q4. The output of this pair is applied directly to transformer T1 to drive the LO input to mixer U1. The mixer also receives the RF input from Al as described above.
c. AGC Controlled IF Amplifier. The output of the mixer is coupled to a bandpass filter FL1. This crystal bandpass filter is designed to reject all the resultant mixer frequencies, except the difference frequency of 40.455 MHz with a bandwidth of 16 kHz . This 16 kHz bandwidth provides an additional option in the $\mathrm{AM}, \mathrm{FM}$ or CW modes of operation. This first IF signal is coupled through C44 to an impedance matching network of L16, C32 and R14, and to a linear amplifier consisting of field effect transistor Q5. A dual tapped transformer T3, makes up part of the load circuit of Q5, to which is connected a current controlled AGC signal. This AGC signal, in effect, varies the impedance of the load transformer which in turn varies the gain of Q5. A second signal from the AGC circuit is applied to the gate of Q5 which varies its bias in relation to the AGC signal strength. This results in a high linear AGC controlled first IF signal, for output to the second mixer circuit card assembly A3.
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Figure 5-3. First Mixer, Functional Block Diagram

5-12. SECOND MIXER, A3. Figure 5-4 shows a simplified functional block diagram of the second mixer circuit card assembly A3. It consists of a three stage AGC controlled first IF amplifier, a bandpass filter, a mixer and output second IF amplifier. Input signals to the circuit card include; the first IF signal from A2, AGC signal from A4, and the second local oscillator signal from A8. The output signals consist of AGC output to A2 and the second IF signal to A4. Figure 84 shows the schematic diagram for the second mixer, A3.
a. First IF Amplifier. The first IF amplifier consists of three stages with the second stage gain controlled from the AGC signal. The third stage drives the signal for input to the bandpass filter. The 40.455 MHz signal routed from A2 is coupled to the drain of field effect transistor Q1, through capacitor C1. The grounded gate of this stage provides high gain for input to filter FL1 through capacitor C5. Filter FL1 provides for rejection of all frequencies other than the 40.455 IF signal. The output of the filter is connected through capacitor C6 to the drain of field effect transistor Q2, which also has a grounded gate. The load circuit of Q2 consists of resistors R16 and R18 and a dual tapped transformer T1. The AGC signal is connected to one tap of the transformer and, in effect, varies the impedance of the load transformer. This action varies the gain of the amplifier in relation to the AGC signal. The output of this stage is taken from the second tap on the transformer and coupled to the base of NPN transistor Q3, through capacitor C11 and resistor R20. A variable coil that forms part of the first section of a four section bandpass filter is connected into the load circuit of Q3. The output of Q3 is, therefore, reflected directly into the bandpass filter. A variable resistor R26 in the emitter circuit of Q3 provides for gain adjustment of this stage
b. Bandpass Filter, Mixer and Second IF Amplifier. The bandpass filter consists of four tunable tank circuits (C15L5, C16-L6, C17-L7, and C18-C19-L8), each made up of a tunable coil and a capacitor. Each stage is tuned to resonate at the first IF signal frequency and reflects its output to the next section for finer tuning and so on. The output of this filter is coupled directly to the input of integrated circuit mixer U3. A 40 MHz signal from the second local oscillator is connected to a second input of the same mixer. A difference frequency of 455 kHz between the 40 MHz and the 40.455 MHz IF signal results in the mixer. It is this difference frequency that is used as the carrier for the second IF signal. All other frequencies are rejected through the filter consisting of capacitors C31 and C32 and coil L9. The output of the mixer, U3, is connected to a tapped load transformer T2. The output is taken from that transformer tap and coupled through C24 and the filter, just described, to an integrated amplifier U4. This stage provides amplification for the second IF signal output from A3.
c. AGC Amplifier. A two section AGC amplifier is contained on circuit card assembly A3 which provides for both voltage and current control of a signal from AGC circuits on A4. This controlled AGC signal is applied to two IF signal stages for level control. One of the IF stages controlled is located on circuit card assembly A2 and described in paragraph 5-11c. The second AGC controlled IF amplifier is on A3 and is described in Paragraph 5-12a. An AGC signal from the AGC circuit on A4 is routed through resistor R2 to two separate amplifiers. The first amplifier is a two stage feedback amplifier consisting of integrated operational amplifiers U2B and U2C. The highly regulated output of this amplifier is routed to circuit card assembly A2 and used as the bias control to the gate of that output IF amplifier. The same AGC signal through resistor R2 is routed to an integrated operational amplifier U2A. This amplifier has both its negative and positive inputs regulated through voltage regulating transistors U1A (positive) and U1B (negative). The action of these transistors control the bias voltage to the operational amplifier which in turn controls its output current flow. The output of this amplifier stage is then coupled to transistor U1C which amplifies the signal and applies it to the IF amplifier through resistor R19 and diode CR1. The diode prevents IF signal feedback to the amplifier. This signal then controls the gain of that IF amplifier in relation to the AGC signal from A4. The output of the operational amplifier U2A is also coupled to the base of transistor U1D which buffers the signal for application to the IF amplifier stage on A2.

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Figure 5-4. Second Mixer, Functional Block Diagram

5-13. MAIN IF/AF, A4. The Main IF/AF circuit card assembly A4 contains IF circuits, AF circuits, and AGC circuits. The description of these circuits are divided into those three basic functions and shown in three separate simplified block diagrams in Figures 5-5, 5-6, and 5-7 Receiver control circuits are also contained on A4 and are described under the A9 circuit card. Input signals to the circuit card include: the second IF signal from A3, the BFO signal from A8, audio and AGC signals from A5 and control signals from A9. Output signals from the board include: second IF signal to A5, BFO, audio and AGC signals to A5, audio signals to A9 and to the rear panel, AGC signals to A3, and control signals to A9. The schematic diagram for the A4 circuit card is shown in Figure 8-5.
a. IF Circuits. Figure 5-5 shows a functional block diagram of the IF circuitry as it functions on circuit card assembly A4. This circuitry consists of the plug in bandpass filters, their switching circuitry, a four stage IF amplifier, an IF output amplifier, an RF switch, a limiting amplifier and FM detector, and a product and synchronous AM detector. The second IF signal routed from A3 is connected directly to seven 455 kHz bandpass filter sockets FL1-FL7. These sockets provide for plugging in any number, up to seven, of preselected filters of various bandwidths. These filters optionally available in bandwidths from 0.4 kHz to 16 kHz , may be changed anytime at the customer's discretion (refer to Chapter 4). Selection of a particular installed filter is then accomplished automatically through the receiver control system. Each filter socket is connected to a diode switch which is controlled from the receiver control circuits. The output of the diode switch for filter FL1 must be linked to the common output of all the other filter switches, if it is used in A4 operation. If it is to be used for the ISB operation, then it must be linked to the output for that circuit card assembly A5. The receiver control is programmed to select and switch into the circuit the filters in ascending order of bandwidths, regardless of the order in which they are plugged into the sockets; that is, when BW1 is selected, from receiver control, the narrowest bandwidth contained in the seven sockets (six if ISB is installed) will automatically be selected. BW2 will select the next widest bandwidth and so on with BW5 selecting the widest bandwidth. Two filters are generally reserved for USB and LSB operation, which are also selected automatically when those modes of operation are called for through receiver control.
(1) The common output of the diode switch which consists of CR1 through CR14, R9 through R22 and C21 through C27, is connected through resistor R25 to the base of transistor Q1. This emitter follower stage acts as a buffer between the diode switch and the input to a two stage IF amplifier U8 with AGC control. An incoming AGC signal is applied to each stage of the integrated circuit amplifier and provides for level control of the IF signal. A variable resistor R39 connected between the output of the first stage and the input of the second stage provides for manual adjustment of the gain of the IF signal. Variable resistor R47 is used for adjusting the AGC signal level. The output of the two stage IF amplifier is connected to a filter consisting of capacitors C44, C46 and C47, resistor R50 and tunable coils L1 and L2. This double tank circuit provides for rejecting unwanted spurious signals. The output of the filter is routed through capacitor C49 to two separate functions; a three stage IF output amplifier and an emitter follower amplifier Q6. The first stage, Q7, of the IF amplifier is an emitter follower which provides buffering between the incoming signal and the second stage Q8. This second stage amplifies the signal and connects it to still another emitter follower stage Q9 for buffering to the IF OUT connector J2, located on the rear panel.
(2) The emitter follower amplifier Q6 acts as a buffer in the same manner as Q7 above but its output is routed to three separate functions; the AGC detector circuit, the product detector and the FM detector through the RF switch. The signal routed to the AGC detector is described in paragraph c under AGC control circuits. The IF signal routed to the FM detector U18 through the RF switch CR22-CR25 is operated by receiver control. Receiver control directs the switch to connect the IF signal to the detector in the AM and FM modes only. In all other modes of operation, the RF switch connects the BFO signal to the FM detector U18. The IF signal routed to the product detector is connected to its signal port. All signals applied to the FM detector; AM, FM or BFO are connected to a limiting amplifier which removes modulation from the AM carrier and passes it or BFO through the output carrier of the FM detector to the carrier input of the product detector. In the FM mode the signal is detected, its carrier rejected, and an audio signal, from the detector audio output, is connected to the detector select switch. This switch, an integrated circuit transistor gate U19A will select the detected FM audio, only in the FM mode, as directed by receiver control. In all modes except FM, a carrier frequency (AM or BFO) is applied to the carrier input of the product detector. In all modes of operation the signal selected, through receiver control, appears on the signal input of the detector. The detector removes the carrier and routes the audio, through its output, to the detector select switch U19A described above. Receiver control directs this switch to select that audio in all modes except FM.


Figure 5-5. A4 IF Circuits, Functional Block Diagram
b. AF Circuits. Figure 5-6 shows a functional block diagram of the audio circuits contained on circuit card assembly A4. This circuit consists of an audio lowpass filter stage, a crosspoint switch, two attenuators and two output amplifiers. The audio signal from detector select U19A is connected, through capacitor C85, to a lowpass filter and amplifier U28. The filter rejects any unwanted frequencies above the audio frequency that might have passed through the detector. The amplifier U28 operates in two different modes. In the AM and FM modes transistor switch U19B disconnects capacitor C113 from the circuit while in all other modes the capacitor is connected across R128 effectively shunting this resistor; thus reducing the signal level in these modes. The output of amplifier U28 is connected to an audio crosspoint switch U25. This switch, through Receiver control, controls audio switching from A4 circuits described above and from the optional A5 circuit card when installed. In non ISB modes the switch routes the A4 signal to the AF GAIN input and to variable attenuator U30. The signal through the AF GAIN control is coupled through capacitor C96 to an audio output amplifier U26. The output of this amplifier is coupled through C108 to AF OUT connector J3 on the rear panel and to the PHONES jack on the front panel. The signal through variable attenuator U30 is routed to connector J8 for output to the ISB circuit card and is also routed back to the crosspoint switch. The switch, in this non ISB mode, connects the attenuated signal through capacitor C95 to a second audio output amplifier on integrated circuit U26. This amplifier drives transformer T1 through C107 and is coupled to the Monitor Line output on connector J3 on the rear panel. The variable attenuator is controlled through screwdriver adjust MAINLINE LEVEL located on the front panel and provides level control of the main (A4) signal to the Monitor Line output in the non ISB mode.
(1) In the ISB mode either main (USB) or ISB (LSB) is selected and routed to the same circuits described in non ISB mode. In addition, the crosspoint switch couples both the USB and LSB through variable attenuators to their respective amplifiers on the ISB circuit card. The USB is routed through attenuator U30 and controls the signal as described in non ISB mode. The LSB signal is routed through attenuator U31 to J8 and back to the crosspoint switch in the same manner as USB. Attenuator U31 is controlled through screwdriver adjust I-LSB LINE LEVEL located on the front panel and provides level control of the LSB (A5) audio signal. The LSB and USB are routed through connector J8 to their respective amplifiers on circuit card A5 and returns through circuit card A4 to Line 1 Output (USB) and Line 2 output (LSB) on AF OUT Connector J3 on the rear panel.
(2) Two audio signals are routed to the AF metering circuit contained on circuit card A4. One signal is tapped from the AF GAIN control input and the second signal from the monitor output amplifier. These two signals are connected to transistor gate U19C which selects between the two signals on direction from receiver control. In all modes except BITE the signal from the monitor output amplifier is selected and routed to the AF metering circuit. This circuit is described under AGC circuits.


Figure 5-6. A4 AF Circuits Functional Block Diagram
c. AGC Control Circuits. Figure 5-7illustrates a functional block diagram of the AGC circuitry contained on circuit card assembly A4. The circuits consist of an AGC detector, AGC decay, peak signal detector, decay time constants, an integrator, filter, a gain control distribution amplifier, a digital to analog converter and various electronic switches controlled from the receiver control circuits. The description also includes the AF/RF meter comparator circuit.
(1) The AGC circuitry is designed to provide three modes or techniques for controlling the gain of the Receiver; Manual, Automatic and Automatic with a selectable threshold. In the automatic mode the level of the IF amplifier U8 is controlled automatically with three selectable decay times; SHORT, MEDIUM and LONG. In the manual mode the IF GAIN control is used to control the level of the AGC signal applied to the IF amplifier U8. The IF GAIN control is used to select the threshold in the automatic with selectable threshold mode. The same decay times as in automatic are selectable in this mode.
(2) An IF signal taken from IF emitter follower Q6 is coupled through capacitor C31 to U1OA for detection. The three transistor array U10 acts as a detector to the IF signal with U1OC connected as an emitter follower for buffering the DC signal to two circuits; AGC decay and peak signal amplifier. Peak signal amplifier U7C couples the signal, across a decay time select circuit, to integrator amplifier U14A. The signal routed to the hang circuit which consists of amplifier U7A and U7B is time controlled through capacitor C42, resistors R45 and R146 and transistors Q2 and Q10. When short time decay is selected, Receiver control turns on transistor Q2 and transistor gate U12A. Capacitor C42 is shorted to ground through transistor Q2 which turns on transistor U1 OD and a short delay is asserted using combination resistors R52 and R55. When medium time decay is selected transistor Q10 and transistor gate U11A is asserted. Capacitor C42 discharges through the parallel resistance of R45 and R146 providing a short hang time, after which U10 is turned on and a medium delay is asserted through R52 and R53. When long time decay is selected capacitor C42 discharges through R45 providing a long hang time, after which U10 is turned on, decay time is through R52.
(3) The AGC applied to integrator amplifier U14A is mixed with signals from diversity AGC through amplifier U14B and gain control or threshold from amplifier U14C when AGC mode dictates. In the manual mode both transistor gates U11C and U12B are enabled through receiver control and the gain control voltage is asserted directly to the input of U14A. In the manual with automatic threshold mode U12B is turned off and the voltage from the IF GAIN control asserts itself through diode CR20 only when that level is higher than the AGC signal at the input of U14A. The digital to analog converter is coupled through U11D and is used to insert threshold level from a remote location through receiver control. Diversity AGC applied through transistor gate U11C to the input of U14C and to amplifier U14B influences the AGC signal only when its level is higher. When AGC dump is enabled (during certain BITE modes and local/remote operations that require dumping of AGC), receiver control enables flip flop U9A which turns on transistor U10E. This rapidly discharges capacitor C52 thereby preventing U14A from acting as an integrator.
(4) The integrator amplifier is coupled to AGC filtering; consisting of capacitor C59, resistors R76, R77, R81 and R83, diode CR21 and amplifier U17A. If AGC dump is asserted (in certain BITE modes) transistor gate U12D is turned on providing a much faster charge path for C59 through resistor R78. The output of the filter amplifier U17A is coupled to amplifier U17B.
This amplifier provides the AGC signal to IF amplifier U8. At the same time U17B provides one input to A3 AGC drive amplifier U17D through diode CR26. If ISB is installed and enabled a second AGC signal from that circuit card is coupled to U17D through diode CR27. The two diodes bias the strongest of the two signals to the input of U17D. The output of this amplifier is coupled through J 2 to AGC circuits on circuit card assembly A3.
(5) An AF/RF meter comparator circuit is contained on circuit card assembly A4. This circuit monitors the main RF, the ISB RF and the AF that may be input from either the main or the ISB signal. The circuits consist mainly of comparator amplifiers U24A (AF), U24B (ISB-RF) and U24C (main-RF). All three amplifiers operate in the same manner with their negative inputs accepting the AF or RF reference while the positive input is referenced from the digital-to-analog converter U21. The output of each amplifier is output through connector J 2 to receiver control. From this information, the microprocessor adjusts the input to U21 which in turn adjusts the converter signal to the AGC in all modes except manual. The output of the three comparators are also processed to the front panel meter readout where the RF or AF signal level can be monitored.
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Figure 5-7. A4 AGC Circuits Functional Block Diagram

5-14. INDEPENDENT SIDEBAND (ISB), A5. The Independent Sideband (ISB) circuit card assembly A5 contains IF circuits, AF circuits and AGC circuits. The description of these basic circuits are divided into those three basic functions and shown in two separate functional block diagrams ir Figures 5-7 and 5-8. Input signals to the circuit card include; IF signal, BFO signal, AGC, and audio signals from A4. Output signals include; AGC and audio to A4, and AGC and audio to AF OUT-J3 on the rear panel, but through circuit card assembly A4. The ISB circuit card assembly schematic diagram is shown in Figure 8-6
a. IF Circuits. Figure 5-8 shows a functional block diagram of the IF circuitry as it functions on circuit card assembly A5 along with the AF function. This circuitry consists of a four stage IF amplifier, a BFO amplifier, a product detector, an audio amplifier, two audio line drive amplifiers, and AGC circuits. The IF signal is routed from bandpass filter FL1 by the filter selection switch, located on circuit card assembly A4 to the IF amplifier on A5. This filter is selected in the ISB mode of operation, so that an IF signal is routed to A5 only in that mode. The IF amplifier is identical to the one located on A4 and is described in Paragraph 5-13. . The IF signal from the AGC controlled IF amplifier is routed both to the AGC circuits and to the signal input port of a product detector U11. A BFO signal from A4 is applied to the base of transistor Q7 through capacitor C30, amplified and applied to the carrier input port of the same detector. The detector removes the carrier and applies, through its audio output port, the audio signal, through capacitor C43, to an audio amplifier.
b. AF Circuits. Figure 5-8 shows a functional block diagram of the AF circuitry, along with the IF circuits, as they function on circuit card assembly A5. This circuitry consists of an audio amplifier and two audio line driver amplifiers. The audio signal, as received from the product detector, is connected, through capacitor C43 and resistors R78 and R79, to the base of emitter follower amplifier Q8. The output of this amplifier is then routed to circuit card assembly A4. Refer to Paragraph 5-13b for a description of the ISB audio on A4. The attenuated audio signal is routed back to A5 and applied to the input of one line driver amplifier integrated in U12. The output of this amplifier drives transformer T2 with a center tapped 600 ohm output. This output is routed to AF OUT-J3 on the rear panel. The main audio signal from the A4 audio is connected to the second amplifier U12 and processed in the same manner as the ISB audio except through transformer T1.
c. AGC Circuits. Figure 5-7 shows a functional block diagram of the AGC circuits for circuit card assembly A4. These circuits are identical to ISB AGC circuits and are described in Paragraph 5-13.

5-15. FIRST LOCAL OSCILLATOR SYNTHESIZER, A7. The first local oscillator synthesizer circuit card assembly A7 contains circuits to produce the first oscillator frequency of 40.955 to 70.454999 MHz for the first mixer which in turn produces the first IF signal of 40.455 MHz . The description of this circuitry is divided into three basic groups; operation of phase lock loops, the digital control circuitry and the oscillator control circuitry. Figures 5-9 through 5-11 respectively, show simplified block diagrams of the three basic circuit divisions. Input signals to the circuit card assembly include receiver control and a 1 MHz reference frequency. The only output signal from the assembly is the 40.955 to 70.454999 MHz oscillator signal.


Figure 5-8. ISB Functional Block Diagram
a. Operation of Phase Lock Loops A basic phase lock loop consists of essentially four main blocks. These are shown in Figure 5-9 and are a Voltage Controlled Oscillator (VCO), a divider capable of dividing the output of the VCO by an integer number $(\div \mathrm{N})$, a phase detector (0), digital-to-analog converter and al Loop Filter Amplifier (LFA). a phase lock loop configured in this manner is capable of locking to the incoming reference frequency ( F ref). This is derived from the fundamental formula for this type of loop which is Fo $=\mathrm{F}$ ref $\times \mathrm{N}$; therefore, to vary the main VCO Frequency (Fo) either Freq or $N$ would have to be changed. Most loops perform frequency change by modification of $N$, the integer divide ratio. It is noted, however, that this type of simple single loop can only vary in frequency steps as small as the reference frequency (Fref). However, it is assumed that a system is possible whereby the main oscillator (Fo) frequency can effectively divide a fractional number, then it is possible to achieve a much finer resolution given the same higher frequency reference. Assume that Fo is 50.123467 MHz and the reference frequency is 100 kHz , then using the above formula the result is with $n$ as the ratio, a non-integer number: $50.123467=0.1 \times \mathrm{n}$; therefore, $\mathrm{n}=501.23467$. If we split this number into its integer part and its decimal part, the result is a three decade integer and a five decade decimal number. Generating the non-integer part as an actual frequency is done by considering a portion of the frequency spectrum of interest between 50.1 MHz and 50.2 MHz where this finally generated frequency will occur. Thus, it is possible to generate any signal between these two frequencies by an averaging technique, that is to say (see Figure 511) if the signal at 50.2 MHz is sampled, 23,467 times and the signal at $50.1 \mathrm{MHz}, 76,533(100,000-23,467)$ times then the average or apparent signal produced by this sampling would occur at the frequency of interest at 50.123467 MHz . This type of sampling produces a large number of sampling sidebands on the main output frequency. These can be removed, however, by producing a signal equal and opposite to these predictable sidebands and adding this to the oscillator control signal and effectively nullifying the production of these sidebands.
(1) In the synthesizer, the circuits can be split in two; for the operating analysis, those circuits involved in the generation of the digital signals to control the generation of the 5-decade decimal part of the divide ratio number ( n ) which in turn controls the sampling technique and the signal to sum with the oscillator control signal, and those circuits including an oscillator, 3-decade integer divider, phase detector, summing amplifier and lowpass filter making up the components of a simple phase lock loop. These two parts in further discussions will be referred to as the digital control and the oscillator control circuitry. A separate section is included for auxiliary circuitry which is provided to produce large frequency step control and outof4-lock indications for the receiver.
b. Digital Control. Figure 5-10 illustrates a simplified block diagram of the digital control circuitry. The circuitry associated with the time control, the incoming 1 MHz reference signal from A 7 J 2 , is used as the clock for accumulator and registers through the NOR gate U4D which drives U20, U21 and U22, and is routed directly to U3, U5 and U18. The Hex D flip-flops U3 and US with U1A and U1B provide a 10-level, ring counter. This counter is used to provide timed pulses to clock the accumulator from first accumulation to second accumulation and sequentially clock out the data in the latches U8 through U12 to the full adder U15. Flip-flop U3 provides a pulse 1 clock pulse wide but delayed 5 pulses from Do the input, to Q1 the output. The output at U3Q4 is connected to the input Do of U5. U5 also provides 1 clock pulse wide pulses but each output Q0 through Q4 is used to drive the incoming data latches. U4A and U4B convert the narrow pulses from U3 Q0 and U5 Q0 into a $50 \%$ duty cycle square wave with a period of 10 clock pulses (each half cycle 5 clock pulses long). The 1800 out-of-phase outputs at U4A pin 1 and U4B pin 4 provide control to U2A, U22, U19A and U19B to ensure that these devices are enabled during the correct half cycle. The 100 kHz reference for the reference side of the phase comparator is taken from U5 output Q4. U5 output Q0 is provided to U4C via U7C to reclock the CARRY IN to U15 and also to U6 to provide the clock for alignment of signals out of the HEX D flip-flop U6.


NOTE:
FOR NORMAL OPERATION
$\mathrm{fo}=\mathrm{fr} \times \mathrm{N}$
WHERE:
$f_{0}=$ OSCILLATOR FREQUENCY
$\mathrm{fr}=$ REFERENCE FREQUENCY
N = DIVIDE BY NUMBER

Figure 5-9. Typical Phase Lock Loop Functional Block Diagram.
(1) Serial-to-Parallel Conversion. The incoming serial data stream from the A9 receiver control assembly consists of DATA, CLOCK and STROBE signals. The strobe is routed to U6 input D1 where its output is reclocked. This output at U6 Q1 is fed back to D2 and its output Q2 provides a strobe input to U13 and U14 one clock pulse delayed. The incoming CLOCK is fed to U14 through U8 in parallel. The serial DATA is fed first into U14 which from its output on U 14 pin 10 to U 12 and U13 shift registers. The output from U12 at pin 10 is fed to U 10 and U 11 , and so on to U 9 and U8 to complete the data load and forming the serial-to-parallel conversion of synthesizer data into the data registers. The data registers U8 through U14 hold the data for the synthesizer frequency, the U8 register holding the 4-bit BCD data for the 1 Hz digit and each register the next decade so the U9 register holds the 10 Hz data and so on to the U14 register which holds the 1 MHz and 10 MHz data.
(a) If the front panel RF frequency is set to 10.426800 MHz then the actual loaded data is 39.255 MHz above this frequency which is 49.680800 . The first IF frequency is 40.455 MHz so we can see that a further offset of 1.2 MHz less than the main LO frequency of 50.881800 is introduced by the microcomputer into the serial data stream sent to A7. This is accounted for in the actual mathematical process in the first and second accumulator circuitry, which replaces this offset before generating the final VCO control voltage to the Local Oscillator.
(2) Accumulator operation. The data loaded into the registers U8 through U12s fed in 4 parallel boards under control of the ring counter U5 during the first half cycle of the tuning as discussed in the tuning section to the Full NBCD Adder U15. As the accumulation proceeds, the accumulating sum is passed from the sum outputs of U15 to the 4-bit wide latch U18. The Carry Out signal from U15 is stored in U18 and is clocked out to U7A and then under control of U7C from the tuning circuits through U4C back into the Carry-In port of U15 deriving the first accumulation the outputs from U18 will be propagated through into the 4 stage shift registers contained in U20 and U21. U20 and U21 are 18 stage registers divided each into 2 four stage registers and 2 five stage registers. The four outputs from U18 are fed into the 4stage registers in each half of each of U20 and U21 and then the output of these 4 -stage registers is fed back to the 5stage register in each half of U20 and U21. At the end of the first accumulation the data at the output of the 4 -stage register appears at the input to the tri-state 4-bit buffer U2A.
(a) As the second accumulation begins U2A is enabled, under control of U4A, and the data at its inputs is transferred to the B inputs of U15 the NBCD Adder. During the five clock periods of the second accumulation the data in U20 and U21 is shifted back to the B inputs of U15. During this period, the data in U8 through U12 is held as the tristate output enable of these registers is not enabled ensuring the results of the first accumulation is added again in the second accumulation. At the end of the second accumulation the results of the first accumulation will be propagated through the 5 -stage register in each half of U20 and U21 and will appear at the inputs Do through D3 of the $4 \times 4$ multiport register U22, and at the Al through A8 inputs to U15. If no new data is loaded into U8 through U12 from the serial input data stream, then the two cycles of accumulation will continue; first accumulating the contents of U8 through U12 on one half cycle with the data at Al through A8 in U15, and on the second half cycle adding the results from the first accumulation back into the B inputs of U15. Temporary storage for the results of each accumulation is provided by U18, U20 and U21.
(b) If the result of adding numbers in U15 is a terminal count, the adder will produce a Carry-Out pulse at U15 pin 6 and reset to zero and start counting again. In a real situation this process is on going and the adder is continually providing Carry-Out pulses. (See Fig. 5-12, line B.) This Carry-Out pulse is fed to U6 to be reclocked. The reclocked output at U6 pin 7 is routed through U1C to a further adder U17. The Carry-Out is also clocked by U19A so that if it occurs on one edge of the accumulating half cycle controlled from U4B, it will appear at the Q output of U19A and after reclocking in U6 through the fifth latch it is applied to the Carry input of the 4-bit Full Adder U17.

## 5-29



UK and US Patents have been obtained covering the synthesizer circuits described nthis page as follows: US Patent No. 4,204.174 Phase Locked Loop Variabl Frequency Generator and US Patent No 3,555,446 Frequency Synthesizer.

Figure 5-10. First Local Oscillator Digital Circuits Functional Block Diagram


Figure 5-11. First Local Oscillator Analog Circuits Functional Block Diagram
(See Fig. 5-13 line C.) The Adder U17 continually updates, by addition, based on the Carry Out information from U15, the 100 kHz frequency information, provided by the input storage register U13. The addition in this adder is continuous so that the outputs at U17 pins 10 through 13 are constantly changing to provide the averaging action previously discussed.
(3) DAC Control. U22 the $4 \times 4$ multiport register provides storage for the results of the constant accumulations and provides the information to the digital-to-analog converter U23. U19B divides the accumulator control signal by two so that all four registers in U22 can be loaded. The read cycles to these registers are controlled by the R0A, R0B, R1A and R1B inputs of U22. The R0A and R0B inputs are fixed and the R1A and R1B inputs are controlled by the output of U19B so that during two accumulations R1 is loaded, each register R0 and R1 consisting of two 4-bit data storage areas. The data stored is that which appears at the data inputs of U22, Do through D3. This stored information is transferred to the register A outputs and register B outputs when W Enable is high and either W0 is high transferring R0A and R0B contents or W0 is high transferring the contents of R1A and R1B.
(a) The digital-to-analog converter provides an output based on the changing data at its inputs as a voltage ramp whose amplitude and DC offset is modified by the adjustment of R5. The D/A also provides a reference source for the pulse to voltage converter U33B. NOTE: The pulses demonstrated in Fig. 5-12 can be reproduced in circuit if the RF front panel frequency is set to 1.046000 MHz . This ensures that the accumulations start from a zero condition. B will then be at TP3, C at TP4 and E at U23, pin 2 (the D/A output).
(4) Oscillator Control. Figure 5-11shows a simplified block diagram of the oscillator control circuits.
(a) Main Division. The BCD data outputs for $100 \mathrm{kHz}, 1 \mathrm{MHz}$ and 10 MHz provided by U 17 and U 14 are applied to a 2-modulus, 3-decade divider consisting of U27, U29, U30 and U31. This form of division ensures that by using a 2 -modulus high speed control device U27 that can divide by 10 or 11 under control of its M1 and M2 input control can divide a high frequency input by an integer value. The terminal count from U29 is applied inverted by Q5 to this control input of U27. The resultant divided signal at pin 9 of U29, U30 and U31 is applied to a TTL to ECL converter network consisting of R30, R31 and CR7 to a flip-flop U28A. This flip-flop reclocks the divided output under control of the clock signal on U28A pin 6 from the ECL output U27 pin 8, and then applied to one side of the phase comparator from its quadrature outputs on pins 2 and 3.
(b) Phase Comparator and Pulse-to-Voltage Converter. The phase comparator reference is derived from the 100 kHz signal from U5 pin 12, reclocked against the 1 MHz reference in U26A. The reference output at U26A is applied to the other side of the phase comparator consisting of U28B, U26B and U32. The ECL comparator provides phase comparator outputs at TP7 and TP8. The variable input from U28A pin 3 is applied to a pulse-width detector consisting of CR9, CR10, Q6 and U33B. As the pulse width changes as the frequency varies from 40.455 MHz to 70.454999 MHz the voltage at the emitter of Q6 varies continuously and linearly over a range of approximately I volt. The DC offset of this voltage is determined by the D/A ref from U23. U33A and Q8 form one-half of a current source to CR12 and CR14 and Q7 and Q9 from the bottom half of this current drive through CR13 and CR15. Phase compared outputs at TP7 and TP8 are fed into the diode network formed by CR12, CR13, CR14 and CR15 and an output from this pulse-to-current converter is fed to R60.
(c) VCO and Analog Control Circuitry. The current output of the phase comparator is combined with the voltage ramp from the D/A through C80. This combined signal is then applied to an integrating amplifier U35.


UK and US Patents have been obtained covering the synthesizer circuits described on this page as follows: US Patent No. 4,204,174 Phase Locked Loop Variable Frequency Generator and US Patent No. 3,555,446 Frequency Synthesizer.

Figure 5-12. Detailed Timing Diagram

In normal operation, the output of U35 is sent to a signal linearizing/inverter circuit U37A and to the out-of-lock window detector comprising U34C and U34D. This is described in more detail in the speed-up and out-of-lock circuit operation. The output of U37A at TP10 is a DC voltage that can vary from a high voltage up to 18 volts and a low voltage equal to 1 volt, it will be high voltage when the selected frequency is at 29.999 MHz and low when the system requires 0.5 MHz . This DC voltage is then passed through U37B which along with its associated resistors and capacitors forms a low-pass filter. This output is then buffered from the VCO by 120K resistor R88 between TP11 and TP12. A further lead-lag network is then in the VCO control line between TP12 and ground formed by R92, R93, R94 and C98. This voltage is then applied to the VCO control varactors CR3 and CR4 through R85 and L4. A voltage applied to CR3 and CR4 will vary the capacitance across the main VCO coil L5 and thus vary the frequency generated. Q1 is the main LO active device and an output from its drain is capacitively coupled to a buffer amplifier of the cascade type formed by Q12 and Q13. The output of this feeds the 2-modulus divider controller U27. A further output from the oscillator coil is tapped off and provides the main LO output through Q2 and Q3 with step down transformer T1.
(d) Speed-Up and Out-of-Lock Operation. When a large step of frequency is introduced on the front panel or from remote the window detector U34C and U34D comparators, compares the inputs on pins 9 and 10 from the VCO control circuitry with fixed high and low references on pins 11 and 8 . If the voltage goes higher or lower (frequency step up or down) than these references a pulse will appear at the comparator outputs on pins 13 and 14. This pulse is applied through an RC network to a voltage converter consisting of U34B and CR19 and CR20. This voltage offset pulse is then used to drive three switches U36B, U36C and U36D. These switches by-pass the lowpass filter U37B and increase the integrating bandwidth of U35 and one switch, provides a feed forward from TP10 to the positive input of U40. U40 provides an integrated drive to push-pull drivers Q10 and Q11, the action of these drivers is to high-speed charge or discharge C98 through R91, C97 and R94. When the control voltage is at approximately the correct voltage for the frequency selected this circuit becomes operative. U34A provides the Out-of-Lock signal for feeding to the A9 receiver control board and then to A6A2 for processing. If a pulse or a constant low level is applied to pin 6 of U34A then its output will go low indicating OOL.

5-16. SECOND LO AND BFO GENERATOR A8. The second local oscillator and beat frequency oscillator circuit card assembly contains the circuitry for these two oscillators. An internal/external frequency reference circuit is also contained on this circuit card. The second local oscillator develops the fixed 40 MHz signals for the second mixer that in turn provides the 455 kHz second IF signal. The BFO is a variable oscillator that provides both the 455 kHz second IF signal and the basic 455 kHz beat frequency for sideband and CW modes of operation. The oscillator, through receiver control, may be either set at 455 kHz or varied plus or minus 8 kHz either side of its basic frequency for CW operation. The internal/external frequency reference circuit provides a reference frequency for both oscillators phase lock loops as well as the first LO contained on A7. In addition, the circuit includes an internal temperature controlled crystal oscillator which supplies a selectable $1 \mathrm{MHz}, 5 \mathrm{MHz}$ or 10 MHz reference frequency output at the rear panel. An external reference frequency can be used in place of the internal reference. The circuit description for the internal/external reference, the second LO and BFO are described under their respective headings with functional block diagrams shown in Figures 5-13 and 5-15. Schematic diagrams of these circuits are shown in Figure 8-11.
a. Internal/External Reference Frequency. The A8 circuit card contains circuitry that permits either an internal or external reference frequency. This reference frequency is required for the operation of all three oscillator synthesizers. A reference in/out connector and switch on the rear panel in addition to linkage on the A8 circuit card provide for the selection of either internal or external frequency and for selecting the proper divide by N frequency for the phase comparator.
(1) Internal Mode. A 5 MHz crystal oscillator Y 1 , located on A 8 is used as the internal reference frequency. With the rear panel REF INT/EXT switch in the INT position, the base of Q1 is grounded through R7 which turns voltage regulator U1 on. The voltage from this regulator enables the temperature controlled crystal oscillator. Approximately 30 minutes are required for maximum stability. The oscillator output is coupled through capacitor CS to the base of transistor switch Q5. With the ground applied through the INT/EXT switch, the base of Q5 is held high through inverter U2A and diode CR3 while the base of transistor switch Q4 is held low through diode CR2. Transistor Q5 conducts, transferring the 5 MHz signal through a TTL square wave shaper Q 6 to one clock input of the phase comparator. See paragraph $b$ (3). The 20 MHz voltage controlled oscillator, described in paragraph (b), is stabilized through the use of this reference frequency. The 20 MHz output of the oscillator is divided by three dividers (two $\div 2$, U7A and U7B and a $\therefore 10$, U6). The two dividers $(\div 2)$ are contained in a single dual D flip-flop. The clock signal ( 20 MHz ) is applied to the clock input of U7A. With the Q output connected to the $D$ input, the $Q$ output provides the 10 MHz reference. The 10 MHz is also connected to the clock input of the second flip-flop and to the divide by 10 circuit. The second flip-flops Q output provides the 5 MHz reference signal. The divide by 10 circuit is a two stage divider ( $\div 2$ and $\div 5$ ). The 10 MHz drives the clock input for the divide by 5 at $B$ input pin 1 . The output of this divider (QD pin 11) is connected back to the clock of the second divider ( $\div 2$ ) at A input pin 14 . Dividing by 5 first then by 2 provides a more symmetrical 1 MHz reference. The 1 MHz signal is output from the divider $(\div 2)$ at $Q A$ output pin 12 . The $1 \mathrm{MHz}, 5 \mathrm{MHz}$ and 10 MHz frequencies derived from these dividers are available for reference through data select switches U4 and U5. The 1 MHz reference is routed directly to the BFO synthesizer and to the first local oscillator synthesizer through NAND gate U11C and connector J2. One of the three frequencies will be selected by data select U5 and routed to rear panel connector J7. Either one of the three frequencies may be selected by proper connection of links LK1 and LK2. In this internal mode, data select $C$ of both U4 and U5 is held low through the INT/EXT switch. Linking LK1 makes data select A low which outputs DO input of both U4 and U5 to their respective Y outputs. For U 4 this is 5 MHz , for U 5 it is 10 MHz . When only LK2 is linked data select A will be high and B low which connects both D1 inputs to their respective Y outputs (U4-5 MHz, $\mathrm{U} 5-5 \mathrm{MHz}$ ). When both LK1 and LK2 is linked both data select $A$ and $B$ are held low which outputs DO of both U 4 and $\mathrm{U} 5(\mathrm{U} 4-5 \mathrm{MHz}, \mathrm{U} 5-1 \mathrm{MHz})$. As noted in this internal mode, U 4 always selects the 5 MHz . It is then routed to the phase comparator as the oscillator reference signal. The output frequency selected by U5 is routed through resistor R13 to buffer amplifier stages Q3 and Q2. These stages provide for output into 50 ohms through a high pass filter, L1, C6 and C 7 , and connector J7 on the rear panel. The high pass filter also provides filtering for reference frequencies applied externally through $J 7$ while resistors $R 8$ and $R 9$ provide a 50 ohm impedance to the incoming reference frequency.
(2) External Mode. In this mode of operation the INT/EXT switch is set to EXT and this line goes high from the +5 volts through resistor R78. This causes transistor Q1 to turn voltage regulator U1 off which in turn turns off the internal crystal oscillator Y1. When an external oscillator is connected to connector J7 on the rear panel, the input is routed through the high pass filter, and capacitor C8 to the base of transistor switch Q4. The input of NOR gate U2A is now high which in turn keeps the base of Q5 low through diode CR3. Transistor Q4, whose base is no longer low, conducts which applies the external reference through the TTL shaper to the same clock input of the phase comparator that the 5 MHz reference was applied in the internal mode. The appropriate reference frequency for application to the second clock input to the phase comparator can be selected through LK1 and LK2 as in the internal mode; however, data select $C$ is now high. Linking LKI selects $D 6(10 \mathrm{MHz})$, LK2 selects $\mathrm{D} 5(5 \mathrm{MHz})$ or both LK1 and LK2 select D4 (1 MHz). The D0, D1, and D2 inputs to U5 cannot be selected when data select C is high (external mode) and no output appears on the Y output of U5.

## Change 1 5-35

b. Second Local Oscillator. Figure 5-13 shows a simplified functional block diagram of the second local oscillator. The circuit consists mainly of a crystal referenced, voltage controlled oscillator, a frequency doubler output circuit and a phase lock loop. The phase lock loop includes amplifiers, an ECL to TTL buffer, three frequency dividers (two -2 and a . 10), reference frequency select circuit, a phase comparator and a digital to analog converter. The three dividers are used to provide a choice of reference frequencies for internal or external reference. These circuits are described in Paragraph 5-16a.
(1) Phase Lock Loop. The 20 MHz oscillator frequency is kept on frequency through a phase locked loop (See Figure 5-9). The oscillator output is routed through a divide by N circuit that provides an oscillator reference frequency. This divided oscillator frequency is coupled to the second clock input of a phase comparator. The phase comparator, described in Paragraph (3), detects any phase shift between the oscillator frequency and a reference frequency also connected to the phase comparator. The phase comparator then changes the digital to analog converter output voltage which is connected to the oscillator varactor and crystal. This then causes the oscillator frequency to change. This loop action will continue until the oscillator frequency is brought into phase (same frequency) with the reference frequency.
(2) Voltage Controlled Oscillator. The 20 MHz oscillator consists of an ECL OR gate U22D, 20 MHz crystal Y2, varactor CR4, resistors R47, R48, R49 and R50 and capacitors C34 through C38. Oscillation frequency is derived from the parallel combination of crystal Y2 in series with C37 and with varactor CR4 in series with C34. The dc voltage applied at the junction of CR4 and Y2 controls the reactance of the parallel circuit, mainly through CR4. This dc voltage, controlled through the phase lock loop described in Paragraph (1) compensates for any frequency shift of the oscillator. The oscillator output is coupled to two buffer amplifiers U22A and U22B. Buffer U22A is used to drive a TTL shaper buffer amplifier, Q10, which shapes the oscillator output into a square wave for input to the divider. These circuits are described in Paragraph a. Buffer U22B drives a frequency doubler circuit, Q11, and associated components. This tuned circuit selects the 40 MHz component of the signal and outputs it through J 3 to circuit card A 3 as the second oscillator frequency.
(3) Phase Comparator and Digital to Analog Converter. The phase comparator is used to detect the phase shift between the reference frequency and the oscillator frequency and to apply control to the digital to analog converter in relation to that phase shift. The comparator consists of dual flip-flop U3A and U3B and NAND gate U2B. Flip-flop U3A is clocked from the reference frequency while U3B is clocked from the divided oscillator frequency. The D inputs to both flip-flops are held high through the +5 volts. When both $Q$ outputs are high the output of NAND gate U2B goes low, resetting both flip-flops with a delay through R32 and C20. Thus when the positive edge of a clock signal clocks a flipflop, $Q$ goes high while $\boldsymbol{Q}$ goes low. If the two clock signals to the flip-flops are in phase, the $\boldsymbol{Q}$ outputs then will go high from the reset action through NAND gate U2B, initiated from the leading edge of both clock pulses. TheQ outputs will remain high through the on-off period of that particular clock pulse. Except for a small delay time introduced by the flipflops the $\boldsymbol{Q}$ outputs will be high most of the time when the two clock signals are in phase. (See Figure 5-14.) When the oscillator frequency leads the reference signal (frequency high), the $\boldsymbol{Q}$ output of U3B remains on less than the $\boldsymbol{Q}$ output of U3A because the leading pulse sets back to $Q$ on U3B before the lagging pulse of the reference signal sets back to $Q$ on U3A. The reverse of this is true when the oscillator frequency lags the reference signal (frequency low). The twoQ outputs of the phase comparator are coupled to a digital to analog converter which consists of transistors Q7, Q8 and Q9, resistors R33 through R39 and capacitors C24 through C26. TheQ output of reference flip-flop U3A is connected to the emitter of Q7 through R33 while the Q output of the oscillator flip-flop U3B is connected to the emitter of Q9 through R34. When the oscillator and reference signal are in phase, the two outputs are the same and transistors Q7 and Q9


UK and US Patents have been obtained covering the synthesizer circuits described on this page as follows: US Patent No. 4,204,174 Phase Locked Loop Variable Freque 555 Generator and C Pate

Figure 5-13. Second Local Oscillator Functional Block Diagram


Figure 5-14. Waveform Diagram: Phase Comparator
conduct at a rate dependent on the amplitude and time period of the pulse. With the amplitude always constant, the amount of conduction then depends only on the time period. When the pulse goes high, the emitters of Q7 and Q9 go more positive causing them to conduct less and when the pulse goes low, they conduct more. The voltage output at the common collectors of Q8 and Q9 would tend to follow this rise and fall in the pulse; however, the inverted signal at the base of Q8 also causes it to conduct less when the pulse is high and more when the pulse goes low. This action converts the pulse signals into a dc level sawtooth waveform at the Q8-Q9 common collectors. The low pass filter, C25, C26 and R39 smooths this waveform and dampens sudden changes caused from changes in the phase comparator pulse rates and in turn stabilizes the phase lock loop. When oscillator frequency increases, the pulse rate increases at Q7 and decreases at Q9, causing a reduction of the dc level output. When oscillator frequency decreases the reverse action takes place. This analog dc output is coupled through R47 to the oscillator for frequency control.
(4) Out of Lock Detector. The out of lock detector consists of NAND gate U2C, resistors R43 and R44 and capacitors C29 and C31. The NAND gate output is held low (phase loop in lock) by the Q outputs of the phase comparator. The resistor capacitor combinations R43-C29 and R44-C31 integrate the square wave signal to provide a constant high on the two NAND gate inputs. If either or both of the two $Q$ outputs from the phase comparator remains low the output of the NAND gate will go high. The output from this circuit is routed to A9 front panel control for processing.
c. Beat Frequency Oscillator (BFO). Figure 5-15 s a simplified block diagram of the beat frequency oscillator. This circuit provides the variable 455 kHz BFO for Receiver CW and sideband operation. The BFO is varied plus or minus 8 kHz through Receiver control. The oscillator operates in a phase locked loop which consists of a voltage controlled oscillator (VCO), a buffer amplifier, a programmed divider, a phase comparator and a digital to analog converter. A divide by 2000 circuit is included to provide a 500 Hz reference signal for the phase comparator. The VCO operates at a center frequency of 22.75 MHz which is 50 times the BFO center frequency of 455 kHz . The oscillator output is routed through a buffer amplifier, TTL shaper, divide by 50 circuit and filter to provide the 455 kHz BFO to the A4 circuit card. An out of lock circuit is also included to detect any out of lock condition of the phase lock loop.
(1) Phase Lock Loop. The phase lock loop for the BFOoscillator functions in the same way as the circuit described for the second local oscillator except the divide by N is made variable through BFO input data to the divide by N circuit.
(2) Voltage Controlled Oscillator. The VCO consists of field effect transistor Q18, opto isolator U21, coils L4 and L5, resistors R66 and R67, capacitors C52 through C56 and C59 and varactors CR6 and CR7. The capacitive reactance of the two varactors in conjunctions with L4 determines the frequency at which the circuit will oscillate. Since varactors change capacitance in relation to the level of the dc voltage applied, the frequency of the oscillator is controlled from the output of the digital to analog converter that is applied to varactors CR6 and CR7 through coil L3. Opto isolator U21, connected to the source of Q18 through resistor R67, provides for on-off control of the oscillator by isolated control of the oscillator source bias. When BFO is enabled in the CW and sideband modes, U21 is enabled through pin 2 which in turn completes the bias path for Q18 through R67. The output of the oscillator is coupled through capacitor C57 to the gate of field effect transistor Q19 which acts as a buffer amplifier between the oscillator and two output circuits. One output of the buffer amplifier provides the oscillator reference frequency through capacitor C72 to the programmed dividers. This circuit is described in Paragraph (3). The second output is coupled through capacitor C60 to a shaper circuit, Q20, R72, R73, R74 and C61. This circuit shapes the waveform into a square-wave for the TTL logic of the divide by circuit U20. The dual decade counter U20 is externally strapped to provide a division of 50 on its QD output,


Figure 5-15. BFO Functional Block Diagram
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pin 9. The variable 22.35 to 23.15 MHz oscillator signal is reduced in frequency by the division of 50 which provides the 447 to 463 kHz BFO. This output is filtered through the filter network consisting of C63, C64, C65, L6, R75, R76 and R77. This filter shapes the digital waveform from U20 into an approximate sine wave signal before being routed to circuit card A4 through connector J4.
(2.1) In the R-2174A version the enable signal is routed directly through R67 to the source of Q18 rather than being routed through opto-isolator U21. Opto-isolator U21 is not used in the R-2174A.
(3) Programmed Dividers. The programmed dividers determine the divide by N number by which the oscillator frequency will be divided for a variable reference to the phase comparator. The program dividers consist of presettable BCD decade counters U14 through U18, divide by 10 or 112 -modules controller U19, NOR gates U12A through U12D and AND gates U13A through U13C. The program divider has two reference inputs; the oscillator frequency coupled through C72 to the V reference and clock inputs of U19 at pins 15 and 16 and the BFO data control inputs to U 15 through U 18 . It is the BFO data inputs in conjunction with U 19 that sets the divide by N number for dividing the VCO frequency. To divide the 22.75 MHz to 500 Hz , for the second clock input to the phase comparator, would require a division of 45500 . The five decade counters are externally strapped to count down from a maximum count of $100000(99999+1)$. The actual data then that would be set on the BFO inputs would be $10000045500=$ 54500 ;however due to gating restrictions between the decade counters the actual number set at the BCD inputs is 54509 at center frequency ( 455 kHz ). The BFO is adjustable plus or minus 8 kHz so that the swing in the BCD inputs must be from $53709(54509-800)$ to $55309(54509+800)$ with 54509 as center. At center frequency counter U18 receives the 10 Hz BCD digits on its parallel inputs P0 through P3 with P0 and P3 $\left(2+2^{4}=9\right)$ high. Counter U17 receives the 100 Hz digits with all inputs low (0), U16 receives the 1 kHz digits with P 0 and $\mathrm{P} 2\left(2+2^{2}=5\right)$ high, and U 15 receives the 10 kHz digits with only P2 $\left(2^{2}=4\right)$ high. Counter U 14 is strapped P 0 and $\mathrm{P} 2\left(2+2^{2}=5\right)$ to the +5 volts (high) and P1 and P3 to ground (low). This supplies the 100 kHz digit which is always 5 . Counter U15 receives only the $2,2^{1}$ and $2^{2}$ BCD digits since the 10 kHz swing is never greater than 7 . The counter is a two modulus divide by 10 or 11 controller and will divide by 10 or 11 for different periods in the clock signal. The TC output pin 15 of U18 coupled to the M I-M2 inputs of UI9 determine the periods at which it will divide with either 10 or 11 . NOR gate U12C ensures that the count enable input of U18 goes low when TC goes high. This assures that the TC output will be retained long enough for action on the M1-M2 inputs of U19. The QTTL output (pin 11) is used to clock the decade counters U14 through U18 and as a reclocking source for NOR gate U12A. Four control inputs, parallel enable (PE), count enable parallel (CEP), count enable trickle (CET) and reset ( R ) select the counters mode of operation. The R and CET of all counters is held high through the +5 volts connected to all counter CET and R inputs. The CEP input of counter $U 17(100 \mathrm{~Hz})$ is also held high from the same +5 volts since this counter does not receive a carry out from previous counters. The CEP of counter U16 is high only when the TC output of U17 is high, the CEP of U15 is high only when the TC output of both C16 and C17 through AND gate U13A are high and the CEP input of U14 is high only when the TC output of both U15 and U16 through AND gate U13B are high. This AND gating of the TC outputs help prevent extra pulses from occurring that are caused from delays in the counters. The PE of all counters are alternately low and high as the TC output to inverter U12B is alternately high and low. With the R and CET of all counters held high (counter resets when R is low) the count mode is enabled when CEP and PE goes high. When PE goes low the counters will synchronously load the data from the BFO inputs into the counters with the count occurring each 500 Hz . The counters output on TC only when PE is held high; however when CEP is held low the TC output will be retained until the next clock pulse. The TC output of the programmed dividers is connected to one input of a two input NOR gate U12A. The clock signal is connected to the second input so that any unwanted pulses, created by delay in the counters and not in sequence with the clock pulse will be rejected. The output of $U 12 \mathrm{~A}$ is routed through inverter U 12 D to one clock input of the phase comparator as the oscillator reference frequency.
(4) Reference Frequency. The 1 MHz reference frequency supplied from internal/ external reference circuits is divided by 2000 to provide a 500 Hz reference frequency to the phase comparator. This division of 2000 is accomplished with two dual decade counters U8 and U9. Counter U8 provides a division by 100 while counter U9 provides division by 20. Each counter has two divide by 2 circuits and two divide by 5 circuits and are externally strapped to provide the divisions by 100 and 20. The division by 100 is accomplished by using all four dividers in the order shown $(\div 2=500 \mathrm{kHz}, \div 5=100 \mathrm{kHz}, \div 5=20 \mathrm{kHz}$ and $\div 2=10 \mathrm{kHz}$ ). The divide by 20 is accomplished in the same manner except its input is the 10 kHz output of the $\div 100$ and the second $\div 5$ is bypassed $(\div 2=5 \mathrm{kHz}, \div 5=1$ kHz , and $\div 2=500 \mathrm{~Hz}$ ). The resultant 500 Hz output is coupled to the clock input of D flip-flop U10A. This flip-flop is contained in a dual flip-flop package which together with NOR gate U11A make up the phase comparator.
(5) Phase Comparator and Digital to Analog Converter. As described in Paragraph (4), the 500 Hz reference frequency is connected to flip-flop UIOA. The second flip-flop UIOB receives its clock signal from the programmed dividers. The D inputs of both flip-flops are tied to the +5 volts (logic 1 ) while both $Q$ outputs are connected through 2 input AND gate U11A and resistor R83 to the reset of both flip-flops. Both flip-flops will reset each time that both Q's go high, causing a logic 0 at the resets of both flip-flops. The clock input signal to U10A (reference) consists of positive going pulses while the signal from the programmed divider (oscillator reference) also contains positive going pulses and is connected to the clock input of U10B. Each flip-flop triggers $Q$ on (high) the positive going pulse of its respective clock signal and at the same time triggers $\boldsymbol{Q}$ to zero (low). Previously as described, when both Q outputs are high the output from AND gate $U 11 \mathrm{~A}$ is low clearing both flip-flops through R83. This resets the Q outputs to low and the $\mathbf{Q}$ outputs to high. Refer to Figure 5-14. The two $Q$ outputs are connected to the digital to analog converter which consists of transistors Q15, Q16 and Q17 and their associated components. This circuit operates in the same manner as the digital to analog converter described in Paragraph b (3) except that adjustments in the BFO frequency provide a different setting of the dc control voltage. This causes-the VCO to change frequency in relation to the BFO setting.
(6) Out of Lock Detector. The out of lock detector consists of NAND gate U11D, resistors R57, R58 and R84 and capacitors C83 and C86. This circuit operates in the same manner as the second local oscillator out of lock circuit described in Paragraph b (4).

5-17. FRONT PANEL AND RECEIVER CONTROL A9. The front panel receiver control circuit card A9 provides for the in and out flow of receiver control data, both from local (front panel) and remote locations. The front panel contains controls and indicators for local operation of the Receiver which includes: two sixteen keypad switch sets, a tuning knob, two Liquid Crystal Displays (LCD), a fault indicator, an IF and AF GAIN control, two audio line level controls, an audio phones jack and a POWER-ON switch. All of these controls, except POWER-ON switch S1, are interfaced to receiver circuits through the A9 circuit card. The IF and AF GAIN controls, two audio level controls and phones jack are routed directly through the A9 circuit card to their respective receiver functions. The keypads, tuning encoder, LCD's and fault indicator connect through various interface circuits on the A9 circuit card to the microprocessor (A6A2). The data is processed by the microcomputer and routed back through interface circuits (A9) to various receiver functions (A4, A5, A7 and A8). Receiver control data entered from a remote location is routed through interface circuits (A6A1) to the microprocessor (A6A2). This data is processed and routed through A9 in the same way as front panel data. Figure 5-16 shows a functional block diagram while the schematic diagram is shown on three sheets in Figure 8-12.
a. Front Panel Switches. Two sets of switch panels, each containing 16 pushbutton switches, are used to enter Receiver control data from the front panel. Switch panel 1 is used for receiver tuning, BFO tuning and for selecting remote or local control. Switch panel 2 is used to select receiver mode, AGC mode, bandwidth and RF/AF meter indication. Each pushbutton switch is a single pole, single throw switch with normally open momentary contacts.

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One contact of each switch is commonly connected to ground within each keypad set. The other contact of each switch is connected to an input of six hex buffers U7, U8, U9, U23, U24 and U25 (two inputs of U8 and U24 are unused) and to +5 volts through a 10 K ohm resistor supplied through resistor array U1, U2, U5 and U6. When any particular switch is open (not pushed), its input to the buffer is held high through the 10 K ohm resistor to the +5 volts. When the switch is closed (pushed), the input line to the buffer goes low. The high or low condition of any switch is passed through the respective buffers to the data bus when that particular buffer is enabled from the address multiplexer U29. The 32 outputs (four are unused) of the six hex buffers are strapped to the 8 -line data bus in groups of eight with each group controlled from a separate address enable signal from U29. This assures that the status of two or more switches will not be transferred through the buffers to the same data line at the same time. The enable signals from U29 are controlled from in/out read and address control signals generated by the microprocessor in order to regularly scan the switch pads for new data. Data on these lines are transferred to the microprocessor as described in Paragraph e.
b. Tuning Encoder. The tuning encoder, operated from the front panel tuning knob, is used to adjust, in fine frequency increments, either Receiver frequency or BFO frequency. Pushbuttons, on the right keypad, select the mode of the tuning encoder and the rate at which frequency data may be entered. The TUNE RATE pushbutton selects one of three rates of change for main Receiver frequency that the tuning encoder will respond; fine ( 1 Hz increments), slow ( 30 Hz increments) and fast ( 1 kHz increments). The BFO pushbutton selects BFO frequency tuning in one rate only ( 10 Hz increments). The LOCK pushbutton disables the encoder while the TUNE RATE or BFO pushbutton enables the encoder from a locked condition. The tuning encoder consists of a 25 segmented disk operating in conjunction with two offset reflective object sensors U59 and U60 with the sensors being mounted on the A9 circuit card. The light from the sensor is reflected from the alternately reflective and non-reflective segments of the disk to the detectors as the disk is rotated. This produces two pulse waveforms (se Figure 5-17). The two detectors are physically positioned, in relation to the disk segments, so that the one waveform either leads or lags the other by 90 degrees. Clockwise rotation of the tuning knob causes the output of U59 to lead that of U60 causing frequency entered to increase. Counter-clockwise rotation causes the waveform of U60 to lead that of U59 causing frequency to decrease. The outputs of the object sensors are coupled, through resistors R16 and R17, to a comparator U58A and U58B. The two inputs to this circuit are compared to a fixed dc voltage input to the comparator from resistor combinations (R8 and R10 for U58A and R9 and R11 for U58B) which control the comparator switching point. The square wave outputs of the comparator are routed to buffer U42B. This data buffer is enabled, from multiplexer U29, in program sequence to transfer the encoder data to data lines FP0 and FP1 for processing as described in Paragraph e.
c. Audio and IF. Audio and IF controls and an audio phones jack are contained on the front panel for control and monitoring of these functions. Control and audio to and from these components are routed through the A9 circuit card from the A4 circuit card with no connection to the in-out digital control hardware on A9. A complete description of their function can be found in Paragraph 5-13
d. Liquid Crystal Displays. Two Liquid Crystal Displays (LCD) are contained on the front panel for displaying Receiver status. Receiver frequency and BFO frequency are displayed on LCD U3 while Receiver mode, AGC mode, tuning mode, remote/local mode, bandwidth and AF/RF metering are displayed on LCD U4. The 8 -line data bus is routed directly to ten 4 -line LCD drivers (U19-U22, U36-U40 and U49) and to thirteen BCD to seven segment decoder/drivers (U10-U18 and U31-U34). The ten line drivers provide decoding and drive for all annunciator displays while the thirteen

CLOCKWISE: (INCREASE FREQUENCY)


COUNTER-CLOCKWISE: (DECREASE FREQUENCY)


Figure 5-17. Encoder Output Signals

7-segment drivers provide decoding and drive for all numerical displays. Multivibrator U26, timed by R1 and C8, provides a 100 Hz signal to the DF (display frequency) inputs of the line drivers, the BP (backplane) inputs of the 7segment drivers and the two LCD's. This signal is used to drive the LCD's display. (NOTE: LCD displays operate from low frequency square wave pulse trains. The contrast of the display being a function of the frequency.) Timing to the line drivers is provided by strobe signals from BCD to decimal decoder U50 while the 7segment drivers receive timing from 4-bit latch and decoder U35. These two decoders receive their input from binary counter U51. The clock pulse (CP) for U51 is derived from three strobes S7, S11 and S15 combined by NOR gate U48A. Decoder U35 is controlled by inversion of strobe S7 through U48B while U50 is timed by strobe S15 through inverter U56. When a strobe signal in program time enables a line driver or 7 -segment driver, the data at its input port is latched into the driver. The 100 Hz signal on the DF or BP input of the drivers is applied to the display segments and is inverted or not inverted through control of the data input to each driver. This in turn controls whether the segment is turned on or off. Any segment of an LCD is turned on by a 180 degree-difference of phase between segment input and the main backplane drive to the LCD (see Figure 5-18). To turn the display segment off the phase difference between these two signals must be zero degrees; that is, the two signals must be in phase (see Figure 5-18).
e. Circuit Card A9. Receiver Control Circuit Card A9 routes data, entered at the front panel, to the microprocessor bus and routes processed data from A6A2 to various Receiver functions on A4, A5, A7 and A8 circuit cards. This data flow in the data bus IOD0 through IOD7 is controlled by read, write and address controls (IOC0 through IOC7) signals from the microprocessor. All data entered from the front panel except audio and IF functions described in Paragraph c, is transferred to the microprocessor for processing, then routed back through A9 to various Receiver functions. Data entered from a remote location is routed to the microprocessor, then the processed data is routed to various Receiver functions through A9 in the same manner as processed front panel data.
(1) Data $\mathrm{In} /$ Out Control. Data flow between the A9 circuit card and the microprocessor is via their respective 8 bi-directional data lines. The flow of data from A9 to A6A2 and vice versa is buffered and controlled by U27, U28 and U46. The bi-directional switches U27 and U28 are configured to transfer data from A9 to A6A2 when enabled in a READ cycle via inverter U56. The tri-state, 8-bit latch U46 is used to transfer data from A6A2 to A9 during a WRITE cycle under the control of the WSTB signal. Control signals (IOCO through IOC3) from the microprocessor are used to generate strobe signals through 4 -bit latch to 16 -line decoder, U47. The strobe signals are timed through the WSTB signal to the strobe input of U47 and from multivibrator U55A and U55B. The multivibrator, controlled from the WSTB signal and inverted IOC5 signal through NOR gate U48, provides timing to the inhibit input of U47. The strobe signals (U47 outputs) are then latched in proper timed sequence and are used to time latches, decoders, gates, etc., that in turn control data flow to various Receiver control functions or displays. The data flow through these components and the strobe signal functions are described in the various Receiver functions that follow.
(2) Receiver Frequency. Processed Receiver frequency data from the microprocessor is routed through data line FPO to one input of two input AND gate U57A. The second input to the AND gate comes from the Q0 output of decoder U64 which receives its input from pre-settable binary counter U51. Strobes S5 and S7, S11 and S15 through NOR gate U48 time U51, strobe S11 through inverter U56 time U64. The frequency data in proper timed sequence is serially routed through AND gate U57A to the A7 circuit card. Strobe signal S12 and the Q0 output of U64 are connected to two input AND gate U57B whose output is routed to A7 as a clock signal for circuits on that card. The Q0 output of U64 is also routed to A7 through inverter U56 as a strobe signal to those same circuits. These circuits on the A7 circuit card further process the serial frequency data in conjunction with the clock and strobe signals, and adjust the oscillator frequency as first directed from the front panel or remote location.


Figure 5-18. Liquid Crystal Display Control Signals.
(3) BFO Frequency. BFO frequency data after being processed by the microprocessor (A6A2) is routed through data lines FP0, FP1 and FP3 to the same binary counter (U51) as the Receiver frequency data. Outputs (Q1, Q2 and Q3) of the binary counter are connected to the three binary inputs of four 8-bit addressable latches U61 through U64. These latches supply the BCD inputs to a programmed divider circuit on the A8 circuit card for BFO frequency control. Timing for the latches is supplied by strobe S11 through inverter U56. The Q4 output of latch U61 provides a signal through transistor Q2 that enables the BFO synthesizer on A8 in CW and SSB/ISB modes.
(4) Circuit Card A4 Functions. All eight data lines along with six strobe signals are routed to control circuits on the A4 circuit card for control of various Receiver functions operating on this circuit card. The control circuits (located on A4) are a direct function of the A9 interface circuits; therefore, a description of their operation is included in this paragraph. All references to components in this paragraph (unless otherwise noted) refer to components on the A4 circuit card. Refer to Figure 8-5 for the schematic diagram of this circuit card. The eight data lines of A9 are routed to 2level translators (U3 and US) on the A4 circuit card. Six strobe signals (STB10 through STB15) are also routed to A4 with strobes STB12 through STB15 connected to the input of level translator U16. These level translators provide a new voltage level to the incoming data for operation of circuits on the A4 card. The strobe outputs from translator U16 are used to clock data latches U2, U4, U13, U15 and U23, the digital-to-analog converter U21, and the crosspoint switch U25. The data bus output from translators U3 and U5 is routed to these components and applied to the various A4 functions as directed by the strobe signals. Data latch U2 is used to select the bandpass filters, but through BCD to 1 of 10 decoder U1. This decoder has only one output high at a time as directed by the BCD input. This allows selection of filter slots FL2 through FL7 in accordance with the three digit BCD on the data line as clocked by strobe 12 on the clock input of U2. The Q4 output of U2 is routed directly to the diode switch of FL1 which is used to select that filter slot at the appropriate program time. Data through data latch U4 is used to control the RF switch that selects between BFO and the IF signal for input to the limiting amplifier and FM detector and D flip-flop U9A. This flip-flop controls the AGC dump line. Data latch U4 is also timed by strobe 12. Data through data latches U13 and U15 control switches in the AGC circuits for various AGC modes. These latches are timed through strobe 13. The output of data latch U23, timed by strobe 15, controls the detector select switch U19A, an audio filter level control switch U19B and an AF meter audio select switch U19C. All eight data lines are connected to the digital-to-analog converter for digital control of its analog output. This unit is timed through strobe 14 and is used to provide analog-to-digital conversion of the DIV AGC line and audio line by peak detector U22A using a successive approximation technique. The digital information gathered by the microprocessor is used for front panel metering. Five data lines are routed to crosspoint switch U25 which is timed by strobe S15. The audio function of the crosspoint switch along with other A4 functions controlled through A9 are described under the A4 circuit card in Paragraph 5-13
(5) Circuit Card A5 Functions. Two data latches on the A5 circuit card are used to control AGC circuit functions on this circuit card. The two latches U7 and U8 have A5 reference designators and are shown on the A5 schematic diagram in Figure 8-6. The 8 -line data bus from translators (described in Paragraph 4) on the A4 circuit card are routed to the data latches with strobe 10 timing both latches through transistor Q6, also located on A5. The data output from the latches is used to control the AGC circuits described under the A5 circuit card in Paragraph 5-14.
(6) Out of Lock (OOL) Functions. Three out of lock circuits, that monitor the condition of the phase lock loops
of the three oscillators, drive OOL indicators on the A9 circuit card and supply their output data to the microprocessor. This data is routed through circuits on the A9 card that drive a fault indicator on the front panel and also a fault indicating circuit that provides a TTL level related to the FAULT condition to connector J3 on the rear panel via A4 circuit card. The three OOL circuit outputs from the three oscillator phase lock loops are connected to the S inputs of dual D flip-flops U45 and U54. These flip-flops are clocked from address multiplexer U29. With the data and reset inputs tied to ground (low) and the $S$ inputs low (phase lock loop in lock) the Q outputs of the flip-flops will be low. If an $S$ input goes high (out of lock) its Q output will go high causing a low through its respective inverter U44A, U44B or U53A. This in turn will enable the applicable LED indicator DS1, DS2 or DS3. These LED indicators are located on the A9 circuit card for accurate determination of the OOL circuit. The Q outputs of the flip-flops are also connected to the inputs of buffer U43 which is also timed by multiplexer U29. The buffer, in program sequence, outputs the status of the OOL circuits on the data bus which is then routed to the microprocessor. The processed FAULT data is routed through binary counter U51 and decoders U63 and U64. The Q5 (FAULT) output of decoder U63 is routed through inverter U52B to the rear panel while the Q5 (FAULT) output of U64 drives the front panel fault indicator through inverter U52A.
(7) AF/RF Meter Comparator Functions. The main RF, ISB RF, and AF comparator circuits along with an ISB fitted circuit from the A4 circuit card are connected to four inputs of non-inverting buffer U42A. The inputs are strapped to +5 volts through 22 K ohm resistors in resistor array U41. When the optional A5 circuit card is installed, input I1 is strapped low through a grounding circuit on that card. The three comparator input (I2, I3 and I4) levels depend on the comparator outputs. Buffer U42A is timed from multiplexer U29 and in program sequence transfers the comparator data as used by A6A2 in a successive approximation technique to generate a bar graph for front panel display to data lines FP0, FP1 and FP2 and the ISB fitted status to data line FP4.
(8) Power Supply Distribution. The A9 circuit card provides a distribution path for dc power, generated from the A10 power supply module, directly to circuit cards A4, A6A2, A7 and A8. This dc power through A9 is further distributed through the A4 circuit card to the optional A5 circuit card and through A6A2 circuit card to optional A6A1 circuit card. The +15 volts dc is also routed to the front panel for operation of the LCD back lighting. The distribution of this dc power is as shown below.

| DC | From A10 To A9J4 | A4 from A9J5 | A6A2 from A9W1 | Output To A7 from A9J6 | A8 from A9J7 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Power | Pin Numbers | Pin Numbers | Pin Numbers | Pin Numbers | Pin Numbers |
| +5 Volts | 3, 4, 7 and 8 | 1 | 23 and 24 | 20 | 26 |
| +5 Volts (unregulated) | 9 and10 |  | 15 and 16 | 15 and 16 |  |
| +15 Volts | $\begin{aligned} & 13,15,19 \\ & 21 \text { and } 22 \end{aligned}$ | 2 and 4 | 19 and 20 | 17 and 18 | 24 |
| +15 Volts (unregulated) | 5 and 6 | 8 |  |  |  |
| -15 Volts | 18 and 20 | 6 | 21 and 22 | 14 | 22 |
| +20 Volts | 1 and 2 |  |  |  |  |

5-18. MICROCOMPUTER BOARD A6A2. A functional description of the microcomputer board A6A2 is given in Paragraph 5-6 with a functional block diagram shown in Figure 5-20
a. As shown on the electrical schematic diagram. Figure 8-9, the microcomputer board A6A2 makes use of a type 3850 CPU (U1). The 3850 CPU is from the Fairchild F8 microcomputer component family. A detailed description of the microcomputer family operation, including timing and instruction set, is given in the F8 Users Guide. This guide is available from the Fairchild Camera and Instrument Corporation, Mountain View, California.
b. The CPU (U1) controls Receiver operation by reading the operating program and routing data and control signals throughout the Receiver based on the procedures and algorithms of this program. The CPU 8 -bit data word bi-directional ports (DBO-DB7) connect to an 8 -line data bus. This bus is common to major Receiver control circuitry (directly or through buffers) and is the primary means of communication between the CPU and other parts of the Receiver.
c. The CPU clock is provided by the 2.0 MHz crystal Y 1 which is connected across pins 38 and 39 of the CPU. The $\phi$ and Write pulse outputs (shown ir Figure 5-19) are clock outputs which provide timing drive for all microcomputer circuitry. The ROMC-0-4 outputs from the CPU connect to other circuits in the Receiver and identify operations which these circuits must perform during any instruction cycle. Interrupt requests are received through the /INT REQ port while the acknowledgment that the CPU will respond to the interrupt requests is routed through the /ICB port. The Input/Output (I/O) ports 00-07 and 10-17 are ports through which the CPU communicates with logic external to the microprocessor. Here, seven of these ports are used and are connected to the serial asynchronous interface module A6A1 (when it is installed in the Receiver).
d. Figure 5-20 is a functional block diagram of the CPU. The inputs and outputs, described previously, are shown as well as the basic functions performed by the CPU. The CPU performs instructions which are obtained from the control program contained in the program memory (EPROMs, U5, U6 and U14). These instructions are routed to the instruction register of the CPU and then carried out during instruction cycles. The control unit logic of the CPU sends out the ROMC signals to other parts of the Receiver during each instruction cycle. These specify the functions to be performed for each instruction. There are four or six clock periods (Figure 5-19) in an instruction cycle (dependent upon the number of 8-bit instruction words required), which are determined by the ROMC signals generated by the CPU. The CPU performs computations, when required by the program, in its ALU (Arithmetic Logic Unit) making use of the accumulator, status register and scratch pad memory registers. The ISAR (Information Storage Address Register) is the address register for the scratch pad memory.
e. The CPU receives and follows the program (sequence of instructions) which is stored in the erasable/programmable read only memory (EPROMs) contained in U5, U6 and U14 (Figure 8-9) These are type 2716 EPROMs. Each of these units contains 2 K 8 -bit words for a total of 6 K 8 -bit words of programmable memory. The CPU also uses the temporary Random Access Memory (RAMs) contained in U7 and U8. These are type 5101L RAMs and provide 2568 -bit words of working memory (temporary data storage). U7 provides 4 bits of each 8 -bit word and U8 provides the remaining 4 bits of each word.
f. The memories are addressed by the CPU through U2, the Static Memory Interface (SMI) type 3853 (Figure 8-9). The CPU sends the ROMC-0-4, Write and $\phi$ signals to the SMI. The SMI, in timed sequence, recognizes the ROMC-0-4 code for a memory access operation, and addresses the appropriate memory, program (EPROM) or working (RAM), over the address bus A0-A10. The SMI (U2) address outputs A11 and A12, through decoders U3A and U3B are also used in


Figure 5-19. Central Processor Unit Timing Signal Diagram
addressing by enabling the appropriate memory chip U5-U8 and U14. Directed by the ROMC-0-4 code and write signals from the CPU, the SMI also directs reading or writing through its CPU READ and /MEM W outputs. Thus, the memory units (when addressed and directed place data on the CPU data bus or accept data from the CPU data bus (DBO-DB7). The EPROMs place 8 -bit instruction words on the bus and the RAMs supply data to or read data from the bus when addressed and directed by the CPU under program control.
g. Figure 5-21 is a functional block diagram of the System Memory Interface (SMI). The inputs and outputs, described previously, are shown. The SMI contains the program counter which contains the program memory address (at which instructions are located) and the data counters which contain the working memory (RAM) addresses. The program counter is either incremented as each instruction in the program is executed or new addresses are inserted by the program or by the interrupt address vector. The working memory addresses are generated by the program. The SMI also contains a timer which is used to generate internal interrupts for initiating program sub-routines at required times during the program cycle.
h. As described previously, the CPU (under program control) writes and reads data to and from the front panel controls and displays and the Receiver circuitry. This data consists of front panel control settings, display readouts and Receiver circuit control and status signals in the form of 8 -bit digital words. The CPU data bus (D00-D07) connects to these units via the tristate bi-directional switch U4 (Figure 8-9) and then through the bi-directional data bus (IOD0-IOD7). The direction of data flow and timing of switch openings are controlled through pins 1 and 19 of U4. Signals to these pins are generated by the strobe logic gates U10-U12 and the Q7 output from the tri-state latched switch U13 at the appropriate times in the program. The strobe logic inputs consist of the ROMC-0-4 and write signals. Thus, at appropriate program times, the CPU data bus connects through U4 and through pins 1-8 of connector J 2 to the front panel and Receiver control circuitry. Also, outputs /IO READ and /WSTB from pins 25 and 27 of J1 go to the front panel and Receiver control circuitry to direct reading or writing data from or to the bus at the appropriate times in the program. The U13 latched outputs (IOC0-IOC7) provide the programmed codes for the selection (or addressing) of the various elements in the front panel and Receiver control circuitry (to accept or supply data from or to the common data bus) at appropriate times in the program. These addresses (IOCO-IOC7) are latched into the $U 13$ outputs from the CPU data bus (D00-D07) at the proper times in the program through the U13 OE (output enable) and CP clock inputs. The OE and CP inputs are generated by the ROMC-04 and write outputs from the CPU through gates U10 and U12.
i. As described previously, the CPU also receives and sends data from and to the Serial Asynchronous Interface modules (A6A1) via the data bus. This data consists of commands and Receiver settings to the CPU and Receiver status to the remote controller in the form of 8 -bit digital words. Connections between these units are made through connector J2. The data bus to the Serial Asynchronous Interface module is labeled PB0-PB7. The microcomputer receives the interrupt request (IINT REQ) signal from the Serial Asynchronous Interface module. The CPU also sends the WRITE, $\phi$ timing signal, strobe logic outputs /IO READ and /WSTB, the tri-state latched outputs IOC0-IOC7, the ROMC-0-4 outputs, the CPU 00-07 and 10-17 outputs and the interrupt acknowledge /ICB signal. These signals are used to direct operations of the interface module and to synchronize operations between the CPU and the remote controller, as will be described later.
j. The reset and memory retention system is composed of stages Q1 through Q7, the 2.0 volt internal power source BT1, R9, CR3 and C16, and the battery charging circuitry, CR6, R9. At Power On, the +5 volt unregulated line (from pins 15 and 16 of J 1 ) switches Q1 and Q2, bringing the /RESET line high. Also at Power On /RESET turns Q4 and Q5 on, C15 is charged, the gate of Q6 is taken negative and CE2 of the RAMs (U7 and U8) is asserted.


Figure 5-20. Central Processor Unit Functional Block Diagram

This activates the RAMs in their normal operating, higher power drain, mode. Also, the gate of Q7 goes negative taking the low off the CPU RESET. The CPU now starts the initialization process. During the turn on cycle the +12 V through R8 quickly turns on Q3, charging C16, which supplies Vcc to the RAMs. CR2 limits the voltage to Vcc while R11 limits the current. CR3 is back biased and the +5 volt supply through CR6 and R9 charges the internal battery BT1.
k. The Schmidt trigger, Q1 and Q2, detects voltage variations in the +5 volt unregulated line. The collector of Q2 will go to ground when the power supply +5 volt unregulated line decreases (such as at power turn off or failure) past the trigger levels set by CR1 and R2-R5-R6. When this occurs, the /RESET line goes low. Also, the voltage at D of Q7 going low, instructs the CPU to come to a stop at the end of the next execution cycle. At power down, Q3 will stop conducting. Initially, the voltage across C 16 still supplies sufficient VDD voltage to the RAMs for the CPU to complete its cycle. When the voltage across C16 falls to approximately 2.4 volts the internal battery BT1, through R9 and CR3 supplies the memory retention power to the RAMs. The voltage out at D of Q6, now low, sets the RAMs at their lower power drain memory retention mode. The battery leakage is greater than the retention memory current drawn by the RAMs; therefore, several months of memory retention is available.
I. As indicated in Figure 8-9, the CPU may receive an interrupt request (/INT REQ) into its pin 23 from the SMI (internal interrupt) or from the Remote Controller Serial Asynchronous Interface (when used) through J2. During an interrupt routine, the CPU will assert its /ICB output (from pin 22), thus not accepting any further interrupts until it has completed the routine. When a Remote Controller Serial Asynchronous Interface is not used, with no J2 connection and with LINK 1 installed (see Figure 8-9), the /ICB output from the CPU connects directly to the /PRI IN input of the SMI. This prevents an internal interrupt from being generated by the SMI until any previously accepted interrupt has been serviced by the CPU. When a Remote Controller Serial Asynchronous Interface is used with the interface connected to J2 and LINK 1 removed, the /ICB output goes directly to the Serial Asynchronous Interface circuitry. As will be described in more detail later, this prevents the interface from generating an external interrupt until any previously accepted interrupt has been serviced by the CPU. In addition, the PRI input (into pin 24 of J2) from the Serial Asynchronous Interface goes to the /PRI IN of the SMI. The Serial Asynchronous Interface is waiting to request an interrupt. Thus, the SMI cannot generate an internal interrupt while either of these conditions exist. It will be noted that the DR and TBRE outputs from the UART (in the Serial Asynchronous Interface, see Figure 8-8 are sent out through connector P1, pins 37 and 36 , to the CPU I/O terminals 11 and 10 . These are used by the microcomputer to determine whether the interrupt request was for a receive or transmit routine.
m . The error outputs from the UART; PE (parity error), OE (overload error) and FE (framing error) are sent to the CPU I/O ports 14, 13 and 12 through connector P1 pins 5, 43 and 42.

5-19. SERIAL ASYNCHRONOUS INTERFACE MODULE A6A1. A functional description of the Serial Asynchronous Interface Module is given in Paragraph 5-6 in conjunction with the functional block diagrams, Figure 5-16.
a. Figure 8-8 is the electrical schematic diagram of the Serial Asynchronous Inteface Module A6A1. This module interfaces an external remote controller (when used) through a serial data line to the microcomputer through an 8-bit data bus. The Universal Asynchronous Receiver transmitter, UART (U9), provides the interface between the serial data lines to and from the remote controller and the microcomputer parallel data bus (PB0-PB7).


Figure 5-21. System Memory Interface Functional Block Diagram
b. The UART contains a transmit and a receive station. A block diagram for each of these sections is shown in Figures 5-22 and 5-23. The receive section converts the incoming serial stream (from the remote controller into the pin 20 RR1 input) to 8 -bit parallel words and places them on the microcomputer bus through output pins 5 through 12. The transmit section, when directed by the microcomputer (through pin 23, data strobe TBRL) takes the 8-bit parallel words from the microcomputer data bus (through pins 26-33) and puts them in a serial format for serial transmission to the remote controller (through pin 25). The serial format, as described previously in Paragraph 4-6. is an 11-bit word containing one start bit, the 7-bit ASCII Code data word, a parity bit followed by two stop bits. Coding for the data word is described in Paragraph 4-6. The UART parallel inputs and outputs are tied together and onto the common microcomputer data bus (Figure 8-8). The UART Receiver inputs and outputs are tri-state, with only inputs or outputs activated at one time.
c. The UART is initialized by the /RESET line from the microcomputer, through pin 2 of connector P1 and inverter U6A, into its MR (pin 21) input. Transmit and receive clock signals, TRC and RRC (pins 40 and 17), which determine the rate of data transfer, are supplied by the programmable Baud Rate generator U4. This generator divides the frequency of crystal, Y 1 , down to the programmed transmit and receive clock rates. Both the receive and transmit clock rates are set at 16 times the baud rate. The frequency generated by $U 4$ is obtained from the data bus, upon initialization by the microcomputer program from information supplied by the configuration set on the rear panel cover into its input RA-RD and TA-TD. This data is strobed into U4 by the strobing signal from output VO (pin 14) of decoder U19. This decoder is driven by microcomputer address and strobing signals IOC0-IOC1, IO READ and /WSTB.
d. As indicated by Table III on the A6A1 schematic diagram the baud rate is set by connecting the appropriate A6A1W1J1 external connector terminals $\mathrm{W}, \mathrm{X}, \mathrm{Y}$ and Z to common. Terminals not connected to common will have +5 V connected through U1. These terminals connect to inputs of tri-state buffer U12. The inputs are strobed onto the data bus (to go to the baud generator inputs, at the appropriate times), by the strobe signal out of X1 (pin 4) of decoder U19. The baud rate is set to match that of the Remote controller. Table 2-7 gives information for setting the baud rate.
e. Also strobed, upon initialization, from output Y1 of decoder U19 is the CRL (pin 34) input to the UART. This loads the UART control register with the EPE, CLS1, CLS2, SBS and P1 inputs at that time. This sets whether parity is used (P1) odd or even parity (EPE), number of bits in a data word (CL1, CL2) and number of stop bits (SBS) used. As described previously, the word length is fixed at 8 and the number of stop bits is fixed at 2 . Table I on the schematic shows that the parity (ON or OFF) and even or odd parity are set through terminals $U$ and $V$ of external connector A6A 1 W 1 J 1 . These are also set by connecting the appropriate terminals to common ground. The unconnected terminals will have +5 V connected through U1. These terminals also connect to tri-state buffer U12 and are strobed on to the data bus (and to P1 and EPE) at the appropriate time by the strobe signal out of X2 (pin 4) of decoder U19.
f. The Receiver number for remote addressing is selected by connecting the appropriate terminals $K, L, M, N, P, R$, S, T of connector A6A1W1J1 (See Table I on schematic) to common. This setting through tri-state buffers U11 and U13B and Pull-up resistors U10 and U1 is strobed onto the data bus, for readout by the microprocessor, by the strobe from X1 (pin 3) of decoder U19.
g. Receiving Data - Figure 5-23 is a block diagram of the receive station of the UART. When data is. received from the remote controller, it comes through the external connector A6A1W1J1 to Receiver U2. Receiver U2 yields signals that are compatible to the receiver

## Change 1 5-57



Figure 5-22. UART Transmitter Functional Block Diagram


Figure 5-23. UART Receiver Functional Block Diagram
circuits of the UART (U9) through its RR1 input (pin 20). The UART Receiver section converts the incoming serial data stream from the remote controller to 8-bit parallel words and places them on the microcomputer bus (PBO through PB7) through outputs RBR1 through RBR8. The serial format, as described previously in Paragraph 4-6, is an 11-bit word containing a start bit, a 7 -bit ASCII code, a parity bit followed by two stop bits. Instruction word codes are also described in Chapter 4. When data is available from the remote controller, an interrupt request is made by asserting the DR output, indicating data is available to the microcomputer. The DR output will go through gate U8A, if not inhibited by computer outputs 00 and 01 (microcomputer not accepting these interrupts), and then through gate U8B, at the correct time (ROMC inputs through U 17A, U6C, U16B-D, and U17B) to drive flip-flop U16A. The outputs of this flip-flop drive one of the two inputs to each of gates U15B and U15C. The other input to these gates is the inverted /ICB signal from the microcomputer. If the CPU is currently blocking, thus ignoring interrupt requests, /ICB will be high. Under this condition U15B and C gates are inhibited. When /ICB is not high (microcomputer accepting interrupts), gates U15B and C are enabled. The output from U15B, through switch U14B, sends the /INT REQ to the microcomputer (through pin 27 of P1) to initiate an interrupt. The microprocessor will orderly stop its normal Receiver monitoring functions as directed by the control program (from EPROM) and start the interrupt routine to accept data from the remote controller.
h. During the interrupt routine the microcomputer asserts its /ICB output, thus not accepting any further interrupts until it has completed the routine. It will be noted that the DR output from the UART are routed through connector P1, pin 37 to the microprocessor I/O terminal 11. This input enables the microcomputer to determine that the interrupt request was for a receive routine.
i. Transmitting data Figure 5-22 is a block diagram of the transmit section of the UART. When the remote controller has requested data (status) and the transmit buffers of the UART are empty, so that it can accept data from the microcomputer for transmission to the remote controller, the TBRE output is asserted. The TBRE signal will go through gate U8A, if it is not inhibited by microcomputer outputs 00 and 01 (microcomputer not accepting interrupts), and then through gate U8B, at the correct time (ROMC inputs through U17A, U6C, U16B-D and U17B) to drive flip-flop U16A. The outputs of these flip-flops drive one of the two inputs to each of gates U15B and U 15C. The other input to these gates is the inverted /ICB signal from the microcomputer. When the CPU in the microcomputer is busy and ignoring interrupt requests, /ICB will be high. Under this condition U15B and C gates are inhibited. When /ICB is not high these gates (U15B and C) are enabled and the /PRI output from U15C is output through pin 24 of P1. The output from U15B, through switch U14B, sends the /INT REQ to the microcomputer (through pin 27 of P1) to initiate an interrupt. The microprocessor will orderly stop its present program and start the interrupt routine to send Receiver status to the remote controller. During the transmit interrupt routine the microcomputer asserts its /ICB output, and will not accept any further interrupts until it has sent the status data to the remote controller. The TBRE output from the UART is routed through connector P1 pin 36 to the microcomputer I/O terminal 10 to identify the interrupt as a transmit routine. The serial output from the UART is output TRO (pin 25) and is routed through line driver U3 to the remote controller through external connector A6A1W1J1.
j. Connections between the serial asynchronous interface module and the remote controller may be selected to accommodate signals compatible with MIL-188C/RS-232C, RS-423 and RS422. The selection is made by connecting the remote controller to the interface external connector A6A1W1J1 in accordance with Table II (shown on the schematic) and using the links (on the Serial Asynchronous Interface module) in accordance with Table IV (shown on the schematic).
k. The interrupt circuitry on the A6A1 module not only generates the interrupt request,
but it also provides the microcomputer with the interrupt vector. This is the program memory address at which the interrupt routine starts. A specific sequence of ROMC signals (put out when the CPU expects the interrupt vector address) is detected by gates U15D, U15A and flip-flops U16C, U16D and U16A. When these signals are received, gate U15A enables switches U13A and U14A to output the interrupt vector on the data bus to be read by the microprocessor.

5-20. POWER SUPPLY, A10. The Receiver contains a power supply module that provides the power required for operation of the Receiver. Refer to schematic diagram, Figure 8-13 The module operates from an ac line input, steps down the voltage, rectifies the ac, filters and regulates the various divided voltages. The unit contains a circuit card switch which provides for switching the transformer input for $100,120,220$ or 240 volts $\pm 10 \%$ operation from the input power line. This line frequency can be between 43 and 420 Hz and is controlled through the POWER-ON toggle switch located on the front panel. The input power is also fused through F1, located on the rear panel for easy access. The 100 or 120 volt input must be fused differently than the 220 or 240 volt input. The alternate fuse is contained in a fuse holder located inside the power supply.
a. DC Power Output. The secondary of transformer T1 contains three separate windings that provide the six different dc outputs for Receiver operation. These six dc outputs, along with their tolerances are listed below:

$$
\begin{aligned}
& +20 \pm 1 \text { volt } \\
& +15 \pm 0.5 \text { volt } \\
& +15 \text { volts unregulated (nominally }+22 \text { volts) } \\
& -15 \pm 0.5 \text { volt } \\
& +5+0.5-0.2 \text { volts } \\
& +5 \text { volts unregulated (nominally }+10 \text { volts) }
\end{aligned}
$$

(1) Conventional bridge rectifiers CR1, CR2 and CR3 provide ac to dc rectification while capacitors C 1, C 4 , C7 and C 10 provide filtering and to smooth the pulsating dc. Capacitors C1 through C9 are connected adjacent to the three voltage regulators to suppress possible oscillations. The rectified and filtered dc from one winding 12 to 13 of the transformer is coupled to dc regulator A10A2, which provides the regulated +20 volts to pins 1 and 14 of A10J3. Winding 6 to 8 provides the +15 volts unregulated to pins 3 and 16 , the +15 volts regulated through regulator U2 to pins $7,8,10$, 11 and 24 and the -15 volts regulated through regulator U3 to pins 22 and 23 . Winding 9 to 11 provides the +5 volts unregulated to pins 5 and 18 and the +5 volts regulated through regulator U 1 to pins $2,4,15$ and 17 of A 10 J 3 . The six dc outputs from A10J3 are routed to various applications throughout the receiver circuitry.

## SECTION III. FUNCTIONAL OPERATION OF MECHANICAL ASSEMBLIES

5-21. TUNING ENCODER. This section contains a description of the mechanical operation of the tuning encoder. The encoder provides a means of selecting either the BFO or reception frequency through the tuning knob located on the front panel. The tuning knob controls the operation of the optical encoder while keypad switches select the mode (BFO or reception) and the tune rate (fast, slow, and fine).
a. The key components of the encoder are a four inch mirror segmented disk and two offset reflective object sensors. One side of the four inch diameter disk is coated with a nonreflective black finish with mirrored surfaces 7.2 degrees wide and placed each $7.2^{\circ}$ around the outer radius of the disk. This provides 25 equally spaced mirrored and black segments approximately $1 / 2$ inch long that are placed between 1.4 and 1.9 inches on the outer radius of the disk.

The disk is mounted on a shaft that holds the disk to the rear of circuit card A9. The shaft extends through a bushing to the front of the front panel with the tuning knob attached to that end. This mechanical arrangement provides for a one to one turning ratio between the knob and disk and is so designed to provide easy spin operation.
b. The two object sensors, U59 and U60, are mounted on circuit card A9, and are placed so that they are adjacent to the segmented surfaces of the disk. When the disk is rotated, the nonreflective black segments of the disk absorb the light emitted by the sensors while the mirrored segments reflect the light back to the sensors. This produces two pulse waveforms whose frequency depends on the rate at which the disk is rotated. The two object sensors are also positioned in relation to the disk segments and to each other so that the waveforms are either leading or lagging each other by 90 degrees (refer to Figure 5-17). The direction of rotation determines which waveform leads the other. A complete description of the electronic operation of the Encoder is described in Paragraph 5-17b while maintenance procedures are presented in Chapter 6.

## CHAPTER 6 MAINTENANCE

6-1. INTRODUCTION. This chapter contains the organizational and intermediate level maintenance procedures for the R-2174(P)/URR Radio Receiver. The information is intended to aid the technician in maintaining the overall performance and general appearance of the unit. Maintenance information is presented in three main sections: Section I contains the organizational and intermediate maintenance which includes preventive maintenance, operational checks and performance test tables. Section II, Special Maintenance, presents the instructions for maintaining the encoder and other items on the front panel assembly. Section III includes the intermediate level performance tests.

6-2. MAINTENANCE REQUIREMENTS. Organizational level maintenance includes inspections, operational checks, faulty circuit card isolation using BITE (Built-In Test Equipment) and minor repairs to the overall Receiver. Intermediate level maintenance includes additional faulty circuit card isolation using the test equipment listed in Table 6-1 the performance test procedures Table 6-3, and special maintenance instructions. When discrepancies or malfunctions are encountered that require extensive dismantling, or isolating failures to individual components, the unit should be returned to the Technical Repair Center. Circuit Cards should also be sent to the Technical Repair Center for repair and alignment. Maintenance of the Receiver should be performed only by technicians with proper training in electronic theory and maintenance including digital and RF circuitry. In addition, the technician should be familiar with the information included in Chapters 1 through 5 of this manual and with any options installed in the particular Receiver to be serviced.

## SECTION I. ORGANIZATIONAL AND INTERMEDIATE MAINTENANCE

6-3. GENERAL. Section I provides the information required to periodically check the general appearance, operational checks of the Receiver and for fault isolation when malfunctions occur. The preventive maintenance procedures call for inspections, cleaning and repairing damaged or deteriorated components and for periodic operational checks. BITE and performance tests check the overall unit performance and for isolating malfunctions to a particular module or circuit card. The procedures should be performed periodically as called for and a maintenance log kept of each performance accomplished, discrepancy found and correction made. Paragraph 6-> covers Organizational Level Maintenance Paragraphs 6-8, 6-9, and 6-10 cover Intermediate Level Maintenance.

## WARNING

6-4. SAFETY PRECAUTIONS. Observe all safety regulations. Do not make connections or replace modules or circuit cards with the ac line cord plugged in. Dangerous potentials may exist with all power removed due to charges retained by capacitors.

6-5. MAINTENANCE SUPPORT EQUIPMENT. Table 6-1 lists the equipment required for intermediate level maintenance procedures. This table includes the manufacturer's model or type number and operating characteristics of the equipment. Equivalent units may be substituted when necessary.

## Change 1 6-1

TABLE 6-1. MAINTENANCE SUPPORT EQUIPMENT

| Item | Equipment Identification (or equiv.) | Characteristics |
| :---: | :---: | :---: |
| 1. | Digital Voltmeter, Fluke 8040A-01 | Range: 0 to 150 Vac and dc 0 to 1 A ac and dc <br> Display: $31 / 2$ digits <br> Accuracy: $\pm 2$ L.S. digit |
| 2. | Oscilloscope, Dual Trace, Portable AN/USM-425(v)1, Tektronix 465M | Sensitivity: 5mV/div. <br> Frequency: dc to 100 MHz |
| 3. | Spectrum Analyzer, Tektronix 4960РT3T | Frequency Range: 0 kHz to 110 MHz <br> Frequency Bandwidth: $10 \mathrm{~Hz}-300 \mathrm{~Hz}$ |
| 4. | HI-Impedance Probe, Spectrum Analyzer Hewlett Packard HP 1121 A | $\begin{aligned} & \text { Frequency Response: } \pm 0.5 \mathrm{~dB} \text { from } 0.1 \\ & \text { to } 110 \mathrm{MHz} \\ & \pm 3 \mathrm{~dB} \text { from } 1 \mathrm{KHz} \\ & \text { to }>500 \mathrm{MHz} \\ & \text { Input Impedance: } 100 \mathrm{~K} \text { ohms shunt } \\ & \text { capacity of } 3 \mathrm{pF} @ 100 \mathrm{MHz} \text { with } 10.1 \\ & \text { divider, } 1 \mathrm{M} \text { ohm with } 1 \mathrm{pF@} @ 100 \mathrm{MHz} \end{aligned}$ |
| 5. | SG-1093/1J AM/FM Signal Generator, Hewlett Packard 8640B -003 | Frequency Range: 500 kHz to 100 MHz <br> Accuracy: $\pm 0.5 \%$ of dial setting <br> Stability: 10 parts in $10^{6}$ <br> Output Level Range: -140 dBm to +20 dBm <br> Modulation: AM -0 to 100\% <br> FM $\pm 150 \mathrm{kHz}$ @ 30 MHz <br> Output Impedance: 50 ohms |
| 6. | X10 Oscilloscope Probe, Tektronix $010-6105-03$ | Tip Impedance: 10 Megohms, 13.5 pF |

6-6. PREVENTIVE MAINTENANCE. Preventive maintenance for the receiver consists of routine maintenance tasks and checks necessary to detect potential malfunction or failure of components. In addition, preventive maintenance defines the necessary cleaning, operational checks and minor calibration required to maintain operational performance standards.
a. Cleaning. The Receiver should be inspected each 168 days, or prior to operation if the unit has been inoperative for an extended period, for cleanliness and good general appearance.

## CAUTION

Use only approved solvents throughout the cleaning procedures. Do not use abrasive or other chemical cleaning agents containing benzene, toluene, zylene, acetone or other chemicals that can damage paint, circuit cards, electrical components and plastics used in the Receiver.
(1) Remove dust, dirt and other foreign matter from the front and rear panels with a soft brush or cloth. Make sure heat sinks are free of dirt and foreign matter.
(2) Inspect the top and bottom covers. Ventilation holes must be kept free of dust and dirt. Refer to Paragraph 6-12d and remove the covers, clean with a soft brush or compressed air.

## WARNING

Compressed air used for cleaning can create airborne particles that may enter the eyes. Pressure shall not exceed 30 PSIG and wearing of goggles is required.

## CAUTION

Do not use bristle brushes or cloths to clean circuit cards. This material may create static electricity which can damage CMOS integrated circuits.
(3) Inspect the interior for dust and dirt collection and for corrosion of solder joints, connections, circuit cards, etc. Use low pressure compressed air to remove dust and other loose matter. Use an approved corrosion solvent to remove corrosion and crystallization.

## WARNING

The filter capacitors used in the power supply will retain an electrical charge after power is removed. The capacitors should be discharged slowly by shorting the terminals through a protected resistive device.
(4) Remove the power supply cover and after discharging filter capacitors inspect and clean in the same manner as described in step (3).
(5) Inspect the encoder disk for scratches, dirt, fingerprints and corrosion. The mirrored surfaces should be especially clean. Clean with a soft cloth and cleaner-solvent.
b. Damage Inspection. Inspect the Receiver every 168 days, or prior to operation if the unit has been out of operation for an extended period, for internal and external damage.
(1) Inspect knobs, heat sinks, switches and controls for damage, tightness, freedom to operate, etc. Replace damaged items as required.
(2) Inspect front and rear panels for missing screws, paint scratches and general appearance. Replace missing screws and touch up paint with matching finish as required.
(3) Inspect the LCD edged-lighting lamps, keypads and FAULT indicator. Refer to Section II if replacement of these components are required.
(4) To inspect the interior of the receiver, remove the top and bottom covers as described in Paragraph 6-12d. Inspect internal circuit boards and components for signs of excessive heat, corrosion, damaged circuits, loose connections or other signs of damage.
(5) Inspect the encoder assembly for freedom of operation and general appearance. Encoder should operate and spin freely when operated with encoder knob on the front panel. Encoder disk should be tight on the shaft and free from wobble as it is rotated. Refer to Section II for encoder repair.
(6) Inspect all connectors for bent or broken pins, damaged shells and corrosion. Clean corrosion with an approved solvent. Replace damaged connectors as required.
(7) Inspect the fuse and circuit card switch in the power supply located on the rear panel. Make sure the fuse is not blown or the circuit card damaged. Replace as necessary.
(8) Inspect cables for frayed insulation, damaged conductors or connectors. Repair oreplace as required.
(9) Inspect the nickel cadmium battery BT1 located on circuit card A6A2. The battery should be replaced if there is any sign of damage, corrosion or leakage of electrolyte.
c. Lubrication. No lubrication is required for the Receiver.
d. Operational Checks. The operational check as outlined in Paragraph 6-7 should be performed monthly to determine that the overall performance of the Receiver is satisfactory. Any sub-standard conditions found through the performance tests should be corrected before placing the unit in normal operation. If discrepancies or malfunctions cannot be corrected, then the fault isolation procedures in paragraph 6-8 should be performed to isolate the malfunction to a circuit card.

6-7. OPERATIONAL CHECK AND FAULT ISOLATION. Operational checkout of the Receiver must include a symptom/diagnostic analysis to augment the results of BITE test procedures. Since BITE is not a panacea for all electromechanical problems, a degree of interpretation by maintenance personnel is necessary. The analysis must include the conditions that preceded the use of BITE as well as those during and following its use. Interpretations will, with a probability of greater than $90 \%$, verify the Receiver operation or isolate a Receiver malfunction to a specific board. Both verification and fault isolation depend upon careful observation of all symptoms starting with the initial step of energizing the Receiver. Further, for fault isolation, certain assumptions must be made in order to facilitate an intelligent assessment. These assumptions are: (1) the previous configuration of the Receiver is correct, i.e., filter complement installed correctly, system interfaces properly connected; (2) the Receiver was properly installed in an operational position (station); (3) the Receiver was functioning correctly prior to the occurrence of the fault; and (4) all connections, connectors, cables and components were checked for correct placement, continuity and tightness.
a. Initial Check. The following procedures detail verification and fault isolation.
(1) Verify that the printed circuit switch in A10J1 on Receiver rear panel matches available line voltage (100, 120,220 or 240 Volts $\mathrm{ac} \pm 10 \%$ at 43 to 420 Hz ) and that the proper fuse is inserted.
(2) Make sure that the REF IN/OUT switch, S2 on rear panel, is set to the INT position.
(3) Energize receiver by turning POWER ON switch to "ON" position. Check FAULT indicator, if on, refer to paragraph 6-8 and correct fault before proceeding.
(4) Observe edge lighting and Liquid Crystal Displays (LCD's). When edge lighting is present and LCD displays contain data, the Power Supply (A10) is working properly.
(a) If edge lighting is not present, the +15 volts from the A10 power supply is malfunctioning. Check this voltage through to A10 and correct the malfunction.
(b) If edge lighting is present and frequency display contains mostly zeros, and no mode indication is present, either the wafer does not match available line voltage, the A10 module is faulty or the A6A2 Microcomputer Assembly is faulty and should be replaced.
(5) Depress the front panel LOCAL/REMOTE push-button switch. The LCD display should indicate a change in Receiver control from no display to REMOTE or vice versa. Repeated pressing of the LOCAL/REMOTE push-button switch should alternate the display between no display and REMOTE indications.
(a) If the Receiver indicates that it is in the REMOTE mode, and depressing the LOCAL/REMOTE pushbutton does not change the display, either the A6A2 Microcomputer Assembly or the A9 Front Panel Assembly is faulty and must be replaced.
(b) If the Receiver display does not indicate that it is in REMOTE, further isolation is possible. Depress the METER RF/AF push-button switch. The meter display should change correspondingly. If it does change, the REMOTE/LOCAL push-button switch is probably bad and the A9 Front Panel Assembly should be replaced. If the meter display does not change, initialize the Receiver by depressing the LOCK and AM push-buttons in that order, then release in the opposite order that they were pressed. Allow the Receiver to initialize (approximately one minute). Once again attempt to change from LOCAL to REMOTE and back; and from RF to AF meter indications. If the display still will not change, the fault is probably on the A6A2 Microcomputer Assembly which should be replaced.

## NOTE

Once steps (1) through (4) have been successfully accomplished, it can be assumed with $60 \%$ confidence that both the A6A2 Microcomputer and the A9 Front Panel Assemblies are functional. To increase the confidence factor, perform step (5).
(6) Ensure that the Receiver is under LOCAL control (indicated by absence of REMOTE in LCD display). Depress the ENTER push-button switch momentarily, followed by numerals 12.345678. These numerals should appear as 12.345678 on the frequency LCD. If this display is correct, then a confidence factor of $99 \%$ can be assumed regarding the total reliability of the A6A2 Assembly and a confidence factor of $60 \%$ for the A9 Assembly.
(7) Ensure that the Receiver is under LOCAL control. If MAN is not displayed in LCD, momentarily depress the MAN push-button on the front panel to achieve Manual IF gain control. Once MAN is visible on the LCD, ensure that SHORT, MED, or LONG are not present in the display. If any of these three indications are present, momentarily depress the corresponding SHORT, MED, or LONG push-button on the front panel to remove the indication from the display.
(8) When set up correctly, only the MAN display should be present in the LCD. Also, ensure that the meter indication is set for RF level. Once this condition has been attained, slowly rotate the IF GAIN potentiometer on the front panel from the full clockwise to full counter-clockwise positions and back again while observing the RF level meter display in the LCD. The RF level display should range from zero (no indication) to full scale and back again to zero. If this range is not attainable, then either the A4 Main IF Assembly (60\% probability) or the A9 Front Panel Assembly (40\% probability) is at fault and should be replaced.

## NOTE

Failure to attain the desired range definitely indicates that proceeding with the BITE checks is useless until the problem is corrected. On the other hand, full range indication demonstrates that the A4 Main IF measurement system used by BITE is functional and BITE checks will yield useful results.
b. BITE Check. Steps (1) through (8) in Paragraph a must be successful before BITE testing can be effectively applied.
(1) Press the LOCK push-button and while still holding, press the AM push-button then release them in the opposite order they were pressed. The Receiver will enter its BITE check mode as indicated by (a) the appearance of REMOTE in the LCD display; (b) the rapidly changing frequencies, modes, and LCD displays on the front panel, and (c) the presence of rapidly changing AF tones and sounds (if AF is being monitored). BITE performs checkout procedures on the following assemblies:

| A2 - First Mixer | A6A2 - (Memory portion) Microcomputer |
| :--- | :--- |
| A3 - Second Mixer | A7 - First and Third LO Synthesizers |
| A4 - Main IF/AF | A8 - Second LO Synthesizer |
| A5 - (If installed) ISB | IF Bandwidth Filters |
|  |  |
|  | NOTE |

By observing the frequency display LCD, it is possible to isolate a fault to the board level by noting BITE error code(s) which appear in that display.
(2) If an error code does appear in the frequency LCD (a two-digit number) refer to Table 6-2. BITE 区ror Code Identification. If more than one board is identified as a suspected fault, depress and hold the LOCK then the CW pushbuttons, then release in the opposite order. Observe the display. A subsequent number will appear which, when considered along with the number which first appeared should identify the faulty board.

## NOTE

When boards A4 and A7 are indicated to be defective, they should be replaced only at the intermediate maintenance level.
(3) Replace the faulty board and repeat steps 1 through3 until no further faults are indicated by BITE. (This will be indicated by the exit from REMOTE and the return of the Receiver to its pre-BITE status.
(4) Successful completion of Paragraphs a and b verify the operational readiness status of all boards except: A1 Lowpass Filter, A6A1 Remote Interface, and the audio portions of A4 Main IF/AF and A5 ISB (if installed).

TABLE 6-2. BITE ERROR CODE IDENTIFICATION

| Displayed Error | Description | Probable Fault |
| :---: | :---: | :---: |
| 01 | First Local Oscillator synthesizer not locked after 100 millisecond delay from 500 kHz step change. | A7 |
| 02 | Second Local Oscillator (Reference) synthesizer not locked. | A8 |
| 03 | First Local Oscillator synthesizer does not break lock to enter fast sampling mode on 500 kHz step change. | A7 |
| 04 | Third Local Oscillator synthesizer not locked after 100 millisecond delay from 500 kHz step change. | A8 |
| 05 | Third Local Oscillator synthesizer not locked after 100 millisecond delay from 500 kHz and 500 Hz step change, respectively. 03 and 05 indicates A7 faults; $04 \& 05$ by itself indicates A8 fault. | A8, A7 |
| 06 | Filter slot one contains a symmetrical filter, but there is/are SSB filter(s) also in the system. | Wrong or bad filter. |
| 07 | Filter slot one contains an upper sideband filter. ISB operation, if installed, will be impaired. | Wrong or bad filter. |
| 08 | No USB filter has been found in the system, and filter slot one does not contain a symmetrical filter. | Wrong or bad filter. |
| 09 | Too many symmetrical filters installed in the system. installed. | Check filters |
| 10 | Not used. |  |
| 11 | No LSB filter has been found in the system and filter slot one does not contain a symmetrical filter. | Wrong or bad filter. |
| 12 | No symmetrical filters have been found in the system. | Wrong or bad filter. |
| 13 | Filter slot one does not contain a lower sideband filter, but ISB is installed. If ISB is installed and no 13 error, A5 is functional. | Wrong or bad filter. |
| 14 | Random access memory test failure: Data written to memory different from data read back. | A6A2 |
| 15 | Either no filters are installed in the system, or the synthesizer signal strength is out of range prescribed for BITE. NOTE: If no 15 error, A2, A3, and A4 are fully functional. | A2 |
| 16 | Filter slot one contains no filter. | Wrong or bad filter. |

TABLE 6-2. BITE ERROR CODE IDENTIFICATION (Cont.)

| Displayed Error | Description | Probable Fault |
| :---: | :---: | :---: |
| 17 | Two or more LSB filters have been found in the system. | Wrong or bad filter. |
| 18 | Two or more USB filters have been found in the system. | Wrong or bad filters. |
| 19 | Although a lower sideband filter has been found in this system, it is not installed in filter slot one. ISB operation, if installed, will be impaired. | Wrong or bad filters. |
| 20 | Not used. |  |
| 21 | Filter in filter slot one is skewed from the IF center frequency. | *Wrong or bad filter. |
| 22 | Filter in filter slot two is skewed from the IF center frequency. | *Wrong or bad filter. |
| 23 | Filter in filter slot three is skewed from the IF center frequency. | *Wrong or bad filter. |
| 24 | Filter in filter slot four is skewed from the IF center frequency. | *Wrong or bad filter. |
| 25 | Filter in filter slot five is skewed from the IF center frequency. | *Wrong or bad filter. |
| 26 | Filter in filter slot six is skewed from the IF center frequency. | *Wrong or bad filter. |
| 27 | Filter in filter slot seven is skewed from the IF center frequency. | *Wrong or bad filter. |
| 28 | Not used. |  |
| 29 | Not used. |  |
| 30 | Not used. |  |
|  | NOTE: 31 through 37; 80\% probability of bad filter; $30 \%$ probability A4 board. |  |
| 31 | BITE frequency sweep underflowed while attempting to measure bandwidth of filter installed in filter slot one. | Bad filter, A4 |
| 32 | BITE frequency sweep underflowed while attempting to measure bandwidth of filter installed in filter slot two. | Bad filter, A4 |

TABLE 6-2. BITE ERROR CODE IDENTIFICATION (Cont.)

| Display <br> Error | Description | Probable Fault |
| :--- | :--- | :--- |
| 33 | BITE frequency sweep underflowed while attempting to <br> measure bandwidth of filter installed in filter slot three. | Bad filter, A4 |
| 34 | BITE frequency sweep underflowed while attempting to <br> measure bandwidth of filter installed in filter slot four. | Bad filter, A4 |
| 35 | BITE frequency sweep underflowed while attempting to <br> measure bandwidth of filter installed in filter slot five. | Bad filter, A4 |
| 36 | BITE frequency sweep underflowed while attempting to <br> measure bandwidth of filter installed in filter slot six. | Bad filter, A4 |
| BITE frequency sweep underflowed while attempting to <br> measure bandwidth of filter installed in filter slot seven. | Bad filter, A4 |  |

*Synthesizers off frequency or wrong or bad filters.
c. A1 Low Pass Filter Check
(1) Connect a coaxial cable (i.e. RG 223/U) between Reference IN/OUT connector J7 and RF IN connector J1. Ensure that S 2 is in the INTERNAL position. Set the receiver controls as follows:

| MODE: | CW |
| :--- | :--- |
| AGC: | Short |
| BFO: | +1.00 kHz |
| FILTER: | Nearest 3.2 kHz |

(2) Using the Frequency Keypad, enter the following frequencies in order, noting that as each frequency is selected, a 1 kHz tone is present at the audio output jack; 01.000000, $05.000000,10.000000$ and 25.000000 MHz .
(3) Should the audio tone not be heard at any of the above frequencies, the A1 Filter Assembly is possibly defective, and should be replaced.
d. A4 Main IF/AF (AF Section) Check.
(1) Connect J 7 to J 1 as described in paragraph c. Using push-button frequency entry, set receiver to 01.000000 MHz . Select CW, BFO +1.00 kHz , and 3.2 kHz (or nearest) bandpass filter. Ensure that the AF GAIN Control has been advanced at least one third clockwise from the fully counter-clockwise position. If the A4 board audio circuitry is operative, a 1 kHz tone will be heard at the phone jack on the front panel. Place the AF/RF meter in the AF mode. Using a screwdriver rotate the MAIN LINE LEVEL control, and note that the AF meter varies from -10 dBm to greater than 0 dBm . This last step will check the operation of the A4 line amplifier.
e. A5 ISB (AF Section) Check (If Installed).
(1) Connect J7 to J1 as described in paragraph c. Using push-button freqency entry, set receiver to 01.001000 MHz . Select ISB/L from the Mode panel. Ensure that the AF

ADJUST AS DIRECTED
UNDER SIGNAL GENERATOR
SETTINGS FIGURE 6-2.


CONNECT AS DIRECTED IN FIGURE 6-2.

ADJUST FOR SIGNAL
BEING MONITORED AS
SHOWN IN FIGURE 6-2.

ADJUST AS DIRECTED UNDER RECEIVER SETTINGS FIGURE 6-2.


ALL MEASUREMENTS MADE REFERENCED TO RECEIVER GROUND UNLESS OTHERWISE NOTED.

Figure 6-1. Fault Isolation, Equipment Functions

GAIN control has been advanced at least one-third clockwise rotation from the fully off position. If the A5 board is operative, a 1 kHz tone will be heard at the phone jack on the front panel. Place the RF/AF meter in the AF mode. Using a screwdriver rotate the I-LSB LINE LEVEL control and note that the AF meter varies -10 dBm to greater than 0 dBm . This last step will check the operation of the ISB line amplifier.

6-8. BOARD LEVEL FAULT ISOLATION. (Intermediate Level Only) Figure 6-1 shows equipment connection methods while Figure 6-2 shows signal flow between individual circuit cards, and the jacks and/or test points for measuring signal values. This information, along with the BITE check detailed in Paragraph 6-7 and the Performance Test Tables detailed in Paragraph 6-9 should be used in isolating malfunctions to the board level. Receiver settings, signal generator settings, and connector pin location diagrams are shown as part of Figure 6-2. The following procedures should be used to perform signal level tests:
a. Set signal generator outputs as specified in Figure 6-2
b. Set receiver front panel controls as specified ir Figure 6-2
c. Measure values indicated at each jack or test point using thetest equipment shown in Table 6-3.
d. Tolerances are: $\pm 3 \mathrm{dBm} ; \pm 0.5$ volts.
e. Once a fault has been isolated, the faulty circuit card should be replaced in accordance with the Receiver assembly and disassembly procedures detailed ih Paragraph 6-12; and Receiver proper operation verified. Verification is accomplished by performing BITE check (paragraph 6-7) and Performance Tests in Section III.

6-9. PERFORMANCE TEST TABLES. (Intermediate Level Only) Table 6-3 lists the signal values present at the test points and significant points of measurement in the Receiver. Unless otherwise stated, the measurements were made with the Receiver set as described in subparagraph b and c. For convenience the circuit cards are grouped in the Table as they are in the Receiver.
a. Test Equipment Required. The test equipment required for the performance test includes the signal generator, item 7; spectrum analyzer, item 4; oscilloscope, item 2; digital multimeter, item 1, of Table 6-1 which lists the recommended test equipment and its performance specifications.
b. Maximum Gain Test Set-Up. The Receiver to be tested should be placed on a convenient work bench. The top and bottom covers of the Receiver and the individual module covers should be removed to gain access to the test points. Set the signal generator to produce a 5.5 MHz signal at an output level of -50 dBm . Connect the signal generator output to the RF INPUT on the Receiver (rear panel connector J1). Tune the Receiver frequency to 5.5 MHz . Select the CW operating mode and adjust the BFO to provide $\mathrm{a}+1.00 \mathrm{kHz}$ offset. Select an IF filter as close as possible to 3.2 kHz bandwidth. Set the AGC to the MAN operating mode and IF GAIN control to maximum clockwise position. Ensure Receiver is operating from own internal reference. Set switch S2, on rear panel to INT position.
c. AGC Test Set-Up. All procedures for AGC test set-up are the same as paragraph b. except set AGC to SHORT operating mode.

TABLE 6-3. PERFORMANCE TEST TABLE

| Step | Connection of Test Equipment |  | Point of Test | Control Settings | Performance Standards |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | FIRST MIXER, CIRCUIT CARD ASSEMBLY A2 |  |  |  |  |
| 1 | Spectrum Analyzer with High Z Probe |  | A2TP1 | As described in | $+6 \mathrm{dBm}, \pm 3 \mathrm{~dB} 45.955 \mathrm{MHz}$ |
| 2 |  |  | A2TP2 | Paragraph 6-9b | $+5 \mathrm{dBm}, \pm 3 \mathrm{~dB} 45.955 \mathrm{MHz}$ |
| 3 | Center Frequency: | 45.955 MHz | A2TP3 |  | + $8 \mathrm{dBm}, 45.955 \mathrm{MHz}$ |
| 4 | Output Level: | $\pm 10 \mathrm{dBm}$ | A2E1 |  | $-51 \mathrm{dBm}, \pm 3 \mathrm{~dB} 5.5 \mathrm{MHz}$ |
| 5 | Input Attenuator: | 20 dB | A2E3 |  | $-51 \mathrm{dBm}, \pm 3 \mathrm{~dB} 5.5 . \mathrm{MHz}$ |
| 6 | Log Reference Level: Bandwidth: Scanwidth: | $\begin{aligned} & 10 \mathrm{dBm}, 10 \mathrm{~dB} / \mathrm{log} \\ & 300 \mathrm{kHz} \\ & 1 \mathrm{MHz} / \text { Div } \end{aligned}$ | A2E8 |  | $-51 \mathrm{dBm}, \pm 3 \mathrm{~dB} 40.455 \mathrm{MHz}$ |
| 78 | Digital Multimeter |  | A2TP4 | As described in | $0.57 \pm 0.05$ Volts dc |
|  |  |  | A2TP5 | Paragraph 6-9b | $+11.6 \pm 0.15$ Volts dc |
|  | SECOND MIXER, CIRCUIT CARD ASSEMBLY A3 |  |  |  |  |
| 9 | Spectrum Analyzer |  | A3E1 | As described in | $-55 \mathrm{dBm}, \pm 3 \mathrm{~dB} 40.455 \mathrm{MHz}$ |
| 10 | with High Z Probe |  | A3E3 | Paragraph 6-9b | $-50 \mathrm{dBm}, \pm 3 \mathrm{~dB} 40.455 \mathrm{MHz}$ |
| 11 | Center Frequency: | 40.455 MHz | A3TP1 |  | $-34 \mathrm{dBm}, \pm 3 \mathrm{~dB} 40.455 \mathrm{MHz}$ |
| 12 | Output Level: | $-50 \mathrm{dBm}$ | A3TP2 |  | $-25 \mathrm{dBm}, \pm 3 \mathrm{~dB} 455 \mathrm{kHz}$ |
| 14 | Input Attenuator: | 10 dB | A3TP3 |  | $-18 \mathrm{dBm}, \pm 3 \mathrm{~dB} 455 \mathrm{kHz}$ |
| 15 | Log Reference Level: Bandwidth: | $0 \mathrm{~dB}, 10 \mathrm{~dB} / \mathrm{log}$ 300 kHz |  |  | $0 \mathrm{dBm}, \pm 3 \mathrm{~dB} 455 \mathrm{kHz}$ |
|  | Scanwidth: | $1 \mathrm{MHz} /$ Div |  |  |  |
|  | Scan Time: | $20 \mathrm{Msec} / \mathrm{Dlv}$ |  |  |  |
|  | Video Filter: | 10 kHz <br> FIRST LO SYNTHESIZER, CIRCUIT CARD ASSEMBLY A7 |  |  |  |
| 16 | Digital Multimeter |  | A7TP1 | As described in | + $5 \pm 0.2$ Volts, dc |
| 17 |  |  | A7TP5 | Paragraph 6-9b. | $+5.2 \pm 0.05$ Volts, dc |
| 18 |  |  | A7TP2 | Reset Receiver to CW | Ground |
| 19 | Oscilloscope |  | A7TP3 | mode (if necessary). | 4 Volt p-p random digital signal |
| 20 | with X10 Probe |  | A7TP3 |  | 4 Volt p-p random digital signal |
| 21 |  |  | A7TP4 |  | 4 Volt p-p random digital signal |
| 22 |  |  | A7TP6 |  | 5 Volt negative going pulses, kHz |
| 23 |  |  | A7TP7 |  | 1 Volt negative going pulse, pos, going |
| 25 | Digital Multimeter |  | A7TP8 |  | 1 Volt narrow pulse, negative going |
|  |  |  | A7TP9 |  | $+8.1 \pm 0.5$ Volts, dc |
| 26 |  |  | A7TP10 |  | $+5.2 \pm 0.5$ Volts, dc |
| 27 |  |  | A7TP11 |  | $+5.2+0.5$ Volts, dc |
| 28 |  |  | A7TP12 |  | $+5.2 \pm 0.5$ Volts, dc |
| 29 |  |  | A7TP13 |  | $+0.2 \pm 0.1$ Volts, dc |
|  |  | SECOND LO/BFO SYNTHESIZER, CIRCUIT CARD ASSEMBLY A8 |  |  |  |
| 30 | Oscilloscope with |  | A8TP10 | As described in | 4 Volts p-p, 5 MHz |
| 31 | X10 Probe |  | A8TP1 | Paragraph 6-9b | 5 Volts p-p, 5 MHz |
| 32 |  |  | A8TP2 |  | 4 Volts p-p, 5 MHz |
| 33 |  |  | A8TP3 |  | 4 Volt Negative going pulse, 5 MHz |
| 34 |  |  | A8TP4 |  | 4 Volt Negative going pulse, 5 MHz |
| 35 | Digital Multimeter |  | A8TP5 |  | 3.2 to 7.5 Volts dc |
| 36 |  |  | A8TP11 |  | 0.4 Volts dc maximum |
| 37 |  |  | A8TP12 |  | 0.4 Volts dc maximum |
| 38 | Oscilloscope with |  | A8TP6 |  | 5 Volt Negative going pulses, 500 Hz |
| 39 | X10 Probe |  | A8TP7 |  | 5 Volt Negative going pulses, 500 Hz |
| 40 | Digital Multimeter |  | A8TP8 |  | 3.2 to 7.5 Volts dc |
|  |  |  | Chang | 1 6-14 |  |

TABLE 6-3. PERFORMANCE TEST TABLE(Cont.)

| Step | Connection of Test Equipment |  | Point of Test | Control Settings | Performance Standards |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 41 | X10 Probe. |  | A8TP9 |  | 0.25 Volts p-p minimum, 22.70 MHz |
|  |  | MAIN IF/AF, CIRCUIT CARD ASSEMBLY A4 |  |  |  |
|  | Spectrum Analyzer | $\begin{aligned} & 455 \mathrm{kHz} \\ & -50 \mathrm{dBm} \\ & 10 \mathrm{~dB} \\ & 0 \mathrm{~dB}, 10 \mathrm{~dB} / \mathrm{log} \\ & 1200 \mathrm{~Hz} \\ & 0.5 \mathrm{kHz} / \mathrm{div} \\ & .5 \mathrm{Seconds} \\ & 100 \mathrm{~Hz} \end{aligned}$ | A4TP1 | As described in | $-38 \mathrm{dBm}, \pm 3 \mathrm{~dB}, 455 \mathrm{kHz}$ |
| 42 | with High Z Probe |  | A4TP2 | Paragraph 6-9c | $-38 \mathrm{dBm}, \pm 3 \mathrm{~dB}, 455 \mathrm{kHz}$ |
| 43 | Center Frequency: |  | A4TP6 |  | $0 \mathrm{dBm}, \pm 3 \mathrm{~dB}, 455 \mathrm{kHz}$ |
| 44 | Output Level: |  | A4TP7 |  | $0 \mathrm{dBm}, \pm 3 \mathrm{~dB}, 455 \mathrm{kHz}$ |
| 45 | Input Attenuator: |  |  |  |  |
|  | Log Reference Level: |  |  |  |  |
|  | Bandwidth: |  |  |  |  |
|  | Scanwidth: |  |  |  |  |
|  | Scan Time: |  |  |  |  |
|  | Video Filter: |  |  |  |  |
| 46 | Digital Multimeter |  | A4TP3 |  | $+0,9 \pm 0.05$ Volts dc |
| 47 |  |  | A4TP4 |  | $+12.7 \pm 0.3$ Volts dc |
| 48 |  |  | A4TP5 |  | $+0.9 \pm 0.05$ Volts dc |
| 49 |  |  | A4TP8 |  | $+6.8 \pm 0.3$ Volts dc |
| 50 |  |  | A4TP9 |  | $+6.8 \pm 0.3$ Volts dc |
| 51 |  |  | A4TP10 |  | $+0.6 \pm 0.04$ Volts dc |
| 52 |  |  | A4TP11 |  | $+5.7 \pm 0.8$ Volts dc |
| 53 |  |  | A4TP14 |  | $+6.5 \pm 0.8$ Volts dc |
| 54 | Oscilloscope with |  | A4TP13 |  | 1.5 Volts P-P, 454 kHz |
| 55 | X10 Probe |  | A4TP15 |  | 1.4 Volts P-P, 1 kHz (with high freq. |
| 56 |  |  | A4TP16 | LINE and AF | 1.4 Volts P-P, 1 kHz component) |
| 57 |  |  | A4TP17 | GAIN controls max. clockwise | 1.2 Volts P-P, 1 kHz sinewave |
|  |  | ISB IF/AF, CIRCUIT | CARD ASSEMBLY A5 (If Installed - Optional) |  |  |
| 58 | Spectrum Analyzer |  | A5TP1 | As described in | $-38 \mathrm{dBm}, 456.8 \mathrm{kHz}$ |
| 59 | with High Z Probe |  | A5TP2 | Paragraph 6-94; | -38 dBm, 455 kHz |
| 60 | Center Frequency: | 455 kHz | A5TP6 | except that the | $0 \mathrm{dBm}, 455 \mathrm{kHz}$ |
| 61 | Output Level: | -54 dBm | A5TP7 | Receiver must be | $0 \mathrm{dBm}, 455 \mathrm{kHz}$ |
|  | Input Attenuator: | 10 dB |  | set to the ISB mode |  |
|  | Log Reference Level: | $0 \mathrm{~dB}, 10 \mathrm{~dB} / \mathrm{log}$ |  | and generator set to |  |
|  | Bandwidth: | $3 \mathrm{kHz} / \mathrm{Div}$ |  | 5.4982 MHz. |  |
|  | Scanwidth: | 20kHz/Div |  |  |  |
|  | Scan Time: | . 5 Seconds |  |  |  |
|  | Video Filter: | 100 Hz |  |  |  |
| 62 | Digital Multimeter |  | A5TP4 |  | +0.9 Volts dc $\pm 0.05$ |
| 63 |  |  | A5TP5 |  | +12.7 Volts dc $\pm 0.3$ |
| 64 |  |  | A5TP8 |  | +6.8 Volts dc $\pm 0.3$ |
| 65 |  |  | A5TP9 |  | +6.8 Volts dc $\pm 0.3$ |
| 66 | Oscilloscope |  | A5TP10 |  | 1.5 Volts p-p. 454 kHz |
| 67 | with X10 Probe |  | A4TP13 |  | 1.4 Volts p-p, 1 kHz |
| 68 |  |  | A5TP11 |  | No level present |
| 69 |  |  | A5TP12 | I-LSB LINE control max. clockwise | 1.4 Volts p-p, 1 kHz |

6-10. FAULT CORRECTION. When a fault or malfunction is traced to a particular module or circuit card it should be replaced with a known functional module or circuit card. To replace the faulty unit refer to paragraph 6-11.

6-11. RECEIVER DISASSEMBLY, INSPECTION AND REASSEMBLY. The Receiver is constructed around a full width, rigid die-cast chassis. The two side gussets are attached directly to this chassis with the front and rear panel assemblies attached to the side gussets. This mainframe assembly houses the receiver circuitry that is accessible through top and bottom covers that attach to the mainframe with quick-lock fasteners. The receiver circuitry is constructed on eleven circuit cards or modules which are attached to the mainframe. Refer tc Figures 6-3 and 6-4.
a. Disassembly. The disassembly procedures are designed to allow the technician to remove any module or circuit card in the Receiver without performing the whole procedure. To remove a particular circuit card perform the preliminary procedures, remove the applicable top or bottom cover then proceed to the paragraph for that circuit card. Figure 6-3 shows a top view of the receiver with cover removed, while Figure 6-4 shows the bottom view with bottom cover and module covers removed.
b. Inspection. After a module or circuit card has been removed it should be carefully inspected for the following:
(1) Check components for excessive heat, loose connections, corroded leads and other damage.
(2) Check connectors for damaged or bent pins, loose connections,corrosion and other damage.
(3) Check the circuit cards for damaged tracks, excessive heat, corrosion or other signs of deterioration.
c. Reassembly. Before reassembly, make sure that the circuit card or module is functional. Repaired circuit cards should be tested before being installed in the Receiver. Refer to the same paragraph and illustration as used for disassembly. Reassembly instructions follow disassembly for each module or circuit card.
d. Preliminary Procedure. The only tools required for disassembly or reassembly are a flat blade screwdriver and screw starter.
(1) Refer th Chapter 2 and remove the Receiver from its mounting rack and place on a suitable work space.
(2) Refer t Figures 6-3 and 6-4 to determine the cover to be removed for access to the module or circuit card to be removed.
(3) Module A1 and front panel assembly A9 require both covers to be removed, circuit cards A2, A3, A7 and A8 require only the bottom covers removal while A4, A5, A6A1, A6A2 and A10 require only the top cover to be removed.
(4) Position the receiver horizontally so that the cover to be removed is up.
(5) Turn the six quarter-turn fasteners, that secure the cover, counter-clockwise one fourth turn to loosen.
(6) Raise the cover up from the back and at the same time slide toward the back of the Receiver until the cover is free from the groove in the back of the front panel.
(7) Inspect the cover for damage, inspect the quarter-turn fasteners in the covers and their receptacks in the side gussets and rear panel. Make sure that ventilation holes in the covers are free of all foreign matter.
(8) After reassembly of module or circuit card the cover may be reinstalled by aligning the cover on the receiver and sliding into the front panel groove. Tighten the quarter-turn fasteners by turning clockwise while applying downward pressure.
e. Module A1 Removal. Refer to Figures 6-3, 6-4 and 6-5.
(1) Perform the Preliminary Procedure as directed in Paragraph d.
(2) Disconnect SMB connector A1W1P1 from J5 on the Receiver frame.
(3) Disconnect BNC connector A1W1J1 from J1 (the RF IN jack) on the rear panel.
(4) Stand the Receiver on its side.
(5) Working from the bottom of the Receiver, loosen and remove the two screws and hardware from the ends of the A1 module.

## CAUTION

Do not allow the A1 module to drop when the screws are removed. Damage to the A1 module or to the Receiver could result from dropping.
(6) Carefully lift out the A1 module from the top of the Receiver.
f. Module A1 Installation. Refer to Figures 6-3, 6-4 and 6-5.
(1) Position the A1 module with the end marked "IN," toward S2 (EXT-INT REF switch).
(2) Install mounting screw with lock washer and flat washer at the "IN" end, ensure that the ground lug on the A1W1J1 cable is placed inside the A1 module between the A1 cover and the threaded mount, and that the mounting screw goes through the lug. Do not tighten.
(3) Install the mounting screw with lock washer and flat washer at the "OUT" end and tighten. Tighte the screw in the "IN" end.
(4) Connect A1W1 J1 connector to J 1 then connect A1W1P1 to J5.
g. Circuit Card A2 Removal. Refer to Figures 6-4 and 6-6.
(1) Perform the Preliminary Procedure as directed in Paragraph d.
(2) Release the six (6) captive screws which hold down the A2 cover.
(3) Remove the A2 cover.


Figure 6-3. Receiver R-2174 (P) URR, Top View, Cover Removed


Figure 6-4. Receiver R-2174 (P) URR, Bottom View, Cover Removed


Figure 6-5A. RF Filter Module Assembly A1


Figure 6-5B. RF Filter A1A1 Printed Circuit
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NOTE: Refer to Parts List for difference in the R-2174A Receiver.


Figure 6-6A. First mixer Circuit Card Assembly A2


Figure 6-6B. First Mixer A2 Printed Circuit


Figure 6-7A. Second Mixer, Circuit Card Assembly A3


Figure 6-7B. Second Mixer A3 Printed Circuit
(4) Disconnect eight-pin connector W21P1 from A2J2.
(5) Disconnect SMB connector W2P1 from A2J1.
(6) Disconnect SMB connector W1P1 from J5 on the Receiver frame.
(7) Disconnect SMB connector A3W1P1 from A2J3.
(8) Loosen and remove the six (6) standoffs which hold down the A2 circuit card.
(9) Lift out the A2 circuit card, tipping it slightly to avoid bumping it against J5 and the four (4) feed-through capacitors located on the side of the receiver frame.

## CAUTION

Bumping J5 or the capacitors may cause damage to the A2 printed circuit track or components.
h. Circuit Card A2 Installation. Refer to Figures 6-4 and 6-6.
(1) Place the A2 circuit card in place by tipping it slightly to avoid bumping it against J 5 and the four feed thru capacitors located on the side of the Receiver frame.
(2) Secure the A2 circuit card with the six standoffs and tighten.
(3) Connect SMB connector A3W1P1 to A2J3.
(4) Connect SMB connector W1P1 to J5.
(5) Connect SMB connector W2P1 to A2J1.
(6) Connect eight pin connector W21P1 to A2J2.
(7) Install the A2 circuit card cover and secure with its six captive screws.
i. Circuit Card A3 Removal. Refer to Fiqures 6-4 and 6-7.
(1) Perform the preliminary procedure as directed in Paragraph d.
(2) Using the pull-tab on the A3 circuit card cover, carefully remove the A3 cover.
(3) Disconnect eight-pin connector W22P1 from A3J1.
(4) Disconnect SMB connector A3W2P1 from J6 on the receiver frame.
(5) Disconnect SMB connector W4P1 from A3J2.
(6) Loosen and remove the six (6) screws, lock washers and flat washers which hold down the A2 circuit card cover.
(7) Remove the A2 cover.
(8) Disconnect SMB connector A3W1P1 from A2J3 on the A2 circuit card.
(9) Loosen and remove the six (6) screws, lock washers and flat washers which hold down the A3 circuit card.
(10) Lift out the A3 circuit card tipping it slightly to avoid bumping the card against J6 and the five (5) feedthrough capacitors located on the side of the receiver frame.

## CAUTION

## Bumping J6 or the capacitors may cause damage to the A3 printed circuit track or components.

j. Circuit Card A3 Installation. Refer to Figures 6-4 and 6-7.
(1) Place the A3 circuit card in place by tipping it slightly to avoid bumping the card against J6 and the five feed-through capacitors on the receiver frame.
(2) Secure the A3 circuit card with six screws, lock washers and flat washers.
(3) Connect SMB connector A3W1P1 to A2J3 on the A2 circuit card. Make sure cable is routed through notch in receiver frame between the A2 and A3 module locations.
(4) Install the A2 circuit card cover with six screws, lock washers and flat washers.
(5) Connect SMB connector W4P1 to A3J2.
(6) Connect SMB connector A3W2P1 to J6 on the Receiver frame.
(7) Connect eight pin connector W22P1 to A3J1.
(8) Carefully install the A3 circuit card cover.
k. Circuit Card A4 Removal. Refer to Figures 6-3 and 6-8.
(1) Perform the preliminary procedures as directed in Paragraph d.
(2) Remove the bandwidth filter cover and bandwidth filters as described in Chapter 2
(3) Disconnect SMB connector W1P2 from A4J1.
(4) Disconnect SMB connector W11P1 from A4J3.
(5) Disconnect SMB connector W12P1 from A4J4.
(6) Disconnect SMB connector W6P2 from A4J5.
(7) Disconnect SMB connector W10P1 from A4J6.


Figure 6-8A. Main IF/AF Circuit Card Assembly A4
(8) Disconnect ribbon connector A5WIPI from A4J8.
(9) Disconnect ribbon connector WI 6PI from A4J7.
(10) Disconnect ribbon connector W15PI from A4J2.
(11) Loosen and remove the twelve (12) screws, lock washers and flat washers whichold down the A4 circuit card.

NOTE
The two (2) screws which hold down transformer T1 to the A4 circuit card do not need to be removed to remove the module nor does the screw through U27. Ten (10) hold down screws are located around the A4 module edges, and the two (2) additional screws are located in the middle area of the module.
(12) Lift out the A4 module.
I. Circuit Card A4 Installation. Refer to Figures 6-3 and 6-8.
(1) Make sure that cables and connectors are out of the way, then place theA4 circuit card in place and secure with twelve (12) screws, lock washers and flat washers.
(2) Connect ribbon connector W15PI to A4J2.
(3) Connect ribbon connector W16P1 to A4J7.
(4) Connect ribbon connector ASWIP1 to A4J8.
(5) Connect SMB connector WIOP1 to A4J6.
(6) Connect SMB connector W6P2 to A4J5.
(7) Connect SMB connector W12PI to A4J4.
(8) Connect SMB connector W11PI to A4J3.
(9) Connect SMB connector WIP2 to A4J1.
(10) Install the bandwidth filters and filter cover as directed in Chapter2
m. Circuit Card A5 Removal. Circuit card A5 is optional and may or may not be installed in the Receiver. Refer to Figures 6-3 and 6-9.
(1) Perform the preliminary procedure as directed in Paragraph d.
(2) Disconnect ribbon connector A5WIP1 from A4J8.

NOTE
The connector disconnects from the A4 module, not the A5 module, and the ribbon cable is part of the A5 module.


Figure 6-9A. ISB Circuit Card Assembly A5


Figure 6-9B. ISB A5 Printed Circuit
(3) Disconnect SMB connector W10P2 from A5J3.
(4) Disconnect SMB connector W11P2 from A5J1.
(5) Loosen and remove the four (4) mounting screws, lock washers and flat washers from the four corners of the A5 module.
(6) Lift out the AS module.
n. Circuit Card A5 Installation. If optional circuit card A5 is being installed for the first time, refer to Chapter 2 for installation of all associated hardware. Refer to Figures 6-3 and 6-9.
(1) Place the A5 module with the ribbon cable toward the A4 module and secure with four screws, lock washers and flat washers.
(2) Connect SMB connector W11P2 to A5J1.
(3) Connect SMB connector W10P2 to A5J3.
(4) Connect ribbon connector ASW1P1 to A4J8.
o. Module A6 Removal. Module A6 consists of circuit card A6A2 and optional circuit card A6Al which may or may not be installed in the Receiver. If it is installed the two circuit cards must be removed together. Refer to Figures 6-3, 610 and 6-11.
(1) Perform the Preliminary Procedure in Paragraph d.
(2) Disconnect ribbon connector A9W1P2 from A6J I on circuit card A6A2.
(3) If circuit card A6A1 is not installed, proceed to step (6). If it is installed, proceed with step (4).
(4) Remove two (2) screws, lock washers, and flat washers, from outside the rear panel, that secure A6A1W1J1 to the rear panel.

NOTE
Some Receivers are supplied with a nylon washer between the A6A1 and its middle-top mounting post. If this nylon washer is installed, it must be saved and reinstalled with the circuit card.
(5) Remove three screws, lock washers and flat washers securing the top of A6A1 to the Receiver. Remove the nylon washer, (if present) secured by the center screw, from between the circuit card and mounting post.
(6) Remove three screws, lock washers and flat washers that secure the top of A6A2 to the Receier.
(7) Carefully lift out both (if A6A1 is installed) circuit cards or A6A2 circuit card. Circuit card A6AI (if installed) may be separated from A6A2 by pulling the two apart which separates the 50 pin connectors A6P1 from A6J 1.
p. Module A6 Installation. If circuit card A6A1 is to be installed in the Receiver it must first be joined with A6A2 and the two then installed together. Refer to Figures 6-3 6-10 and 6-11.
(1) If A6A1 is to be installed with A6A2,perform steps (2) through (8). If A6A2 is installed alone, perform steps (3), (7) and (8).
(2) Carefully align and mate the 50 pin connectors of A6A1 and A6A2. Apply just enough pressure to bring them snugly together.
(3) Install either the two circuit cards together or the A6A2 circuit card alone by inserting the bottom of the card(s) into the nylon track on the receiver frame. Be sure that screw holes in the circuit card(s) align with screw holes in the receiver standoffs.

## CAUTION

If a nylon washer is supplied as part of the hardware kit for A6A1 it must be installed between the Receiver standoff and the center mounting hole of A6A1 to prevent electrical damage from occurring.
(4) Place nylon washer (if supplied) between center mounting hole of circuit card A6A1 and receiver standoff. Insert screw, lock washer and flat washer through circuit card and nylon washer into standoff. Do not tighten screw.
(5) Install the other two (2) screws, lock washers and flat washers into circuit card A6A1. Do not tighten screws.
(6) Install two (2) screws, lock washers and flat washers from rear of rear panel into connector A6A1W1J1 and tighten. Tighten the three (3) screws holding A6A1 circuit card.
(7) Install three (3) screws, lock washers and flat washers through A6A2 mounting holes into the receiver standoffs and tighten screws.
(8) Connect ribbon connector A9W1P2 to A6J1 on circuit card A6A2.
q. Circuit Card A7 Removal. Refer to Figures 6-4 and 6-12.
(1) Perform the preliminary procedure as directed in Paragraph d.
(2) Loosen and remove the single screw and hardware that holds down the A7 cover.
(3) Using pull-tab remove the A7 cover.
(4) Disconnect ribbon connector WI3P1 from A7J1.
(5) Disconnect SMB connector W3PI from A7J2.
(6) Disconnect SMB connector W2P2 from A7J4.
(7) Loosen and remove the eleven (11) screws, lock washers and flat washers and nylon mounting post (from which the A7 cover screw was removed in Step (2) above) which holds down the A7 circuit card.


Figure 6-10A. Serial Asynchronous Interface Circuit Card A6A1 Assembly


Figure 6-10B. Serial Asynchronous Interface A6A1 Printed Circuit


Figure 6-11A. Microcomputer Circuit Card A6A2 Assembly


Figure 6-11B. Microcomputer A6A2 Printed Circuit

NOTE
Refer to Parts List for differences in the R-2174A Receiver.


Figure 6-12A. First L.O. Circuit Card Assembly A7
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## NOTE

The screw through A7U39 is not a card holddown screw. The screw through regulator A7Q4 is a module holddown screw. When removing the screw through A7Q4, take care not to lose the mica insulator which fits between A7Q4 and the A7 circuit card.
(8) Lift out the A7 circuit card.
r. Circuit Card A7 Installation. Refer to Figures 6-4 and 6-12.
(1) Place the A7 circuit card in place, with ribbon connector A7J 1 located next to chassis cut out, and secure with ten screws, lock washers and flat washers and nylon mounting post.

## CAUTION

A mica insulator must be in place between A7Q4 and the circuit card. Do not overtighten the screw that secures A7Q4 or damage may result.
(2) Make sure mica insulator is between A7Q4 and circuit card and secure A7Q4 with screw, lock washer and flat washer. Do not over tighten.
(3) Connect SMB connector W2P2 to A7J4.
(4) Connect SMB connector W3PI to A7J2.
(5) Connect ribbon connector WI2PI to A7J1.
(6) Carefully install the A7 cover and secure with screw, lock washers and flat washer.
s. Circuit Card A8 Removal. Refer to Figures 6-4 and 6-13.
(1) Perform the Preliminary Procedure as directed in Paragraph d.
(2) Using the pull-tab, remove the A8 cover.
(3) Disconnect ribbon connector W14P1 from A8J5.
(4) Disconnect SMB connector W7P1 from A8J1.
(5) Disconnect SMB connector W6P1 from A8J4.
(6) Disconnect SMB connector W3P2 from A8J2.
(7) Disconnect SMB connector W4P2 from A8J3.
(8) Remove the eight (8) screws, lock washers and flat washers which hold down the A8 circuit card.
(9) Lift out the A8 circuit card.

NOTE
Refer to Parts List for differences in the R-2174A Receiver.


Figure 6-13A. Second LO/BFO Circuit Card Assembly A8

NOTE
Refer to Parts List for differences in the R-2174A Receiver.


Figure 6-13B. Second LO/BFO Synthesizer A8 Printed Circuit


Figure 6-14A. Receiver Control Circuit Card A9 Assembly
t. Circuit Card A8 Installation. Refer to Figures 6-4 and 6-13.
(1) Place the A8 circuit card so that ribbon connector J5 is located next to the cutout in the receiver chassis and secure with eight (8) screws, lock washers and flat washers.
(2) Connect SMB connector W4P2 b A8J3.
(3) Connect SMB connector W3P2 to A8J2.
(4) Connect SMB connector W6P1 to A8J4.
(5) Connect SMB connector W7P1 to A8J1.
(6) Connect ribbon connector W14P1 to A8J5.
(7) Make sure cables entering the A8 compartment are routed through respective cutouts in the receiver chassis and install the A8 cover.
u. Circuit Card A9 Removal, Inspection and Installation. Circuit card A9 is installed on the front panel assembly which includes the front panel, encoder assembly, keypads, edgelighting, LCD's, and other controls and switches. The disassembly, inspection and reassembly of the front panel assembly must be accomplished in order to remove the A9 circuit card and requires intricate procedures. These procedures are included in Section II of this chapter under special maintenance.

## WARNING

The filter capacitors used in the power supply will retain an electrical charge after power is removed. The capacitors should be discharged slowly by shorting the terminals through a protected resistive device.
v. Module A10 Removal. Refer to Figures 6-3 and 6-15.
(1) Perform the preliminary procedure as directed in Paragraph d.
(2) Loosen (counter-clockwise) the four (4) $1 / 4$ turn fasteners on the A10 module cover.
(3) Carefully lift out the A10 module cover and discharge filter capacitors.
(4) Loosen three captive screws from the bandwidth filter cover on A4 and remove the cover.
(5) Disengage the locking clips (2 each) on connectors AI 0J2 and AI 0 J 3 located on the outside of the A10 module.
(6) Carefully unplug the connectors from A10J2 and A10J3.
(7) Loosen and remove the five (5) screws, lock washers and flat washers (located on the outside of the rear panel) which hold the back panel of the AI 0 module to the rear panel of the Receiver. Loosen the four captive screws at the base of the A0O module.
(8) Grasping the sides of the AIO module, slide the module away from the rear panel while lifting the end of the A10 module nearest the bandwidth filters and lift out the A10 module.


Figure 6-15A. Power Supply Module A10, Top View


Figure 6-15B. Power Supply Module A1 0, Rear View

CAUTION
Take care not to crush the A10J2 or A10J3 connectors, the cooling fins on A10U1 or A10U2, or to snag the WP2 wire which connects to A4J1.

## WARNING

The filter capacitors used in the power supply will retain an electrical charge after power is removed. The capacitors should be discharged slowly by shorting the terminals through a protected resistive device.
w. Module A10 Installation. Refer to Figures 6-3 and 6-15.
(1) Grasping the sides of the AIO module, keep the end toward the filters tilted upward and slide toward the rear panel.

## CAUTION

Take care not to crush the A10J2 or A10J3 connectors, the cooling fins of AIOUI or A 10U2 or to snag the WP2 wire which connects to A4J1.
(2) Install the five screws, lock washers and flat washers through the outside of the rear panel into the A 10 module. Do not tighten.
(3) Insert the 4 captive screws in the bottom of the AIOO module and tighten. Tighten the five screws in the rear panel.
(4) Connect W20PI to A10J2 and W19P3 to A10J3 and secure with locking clips.
(5) Install the power supply cover and secure with the four one quarter turn fasteners.
(6) Install the bandpass filter cover on circuit card A4 and æcure with its three (3) captive screws.

## SECTION II. SPECIAL MAINTENANCE

6-12. GENERAL. Special maintenance instructions are included here to facilitate maintenance at the intermediate level so that the unit does not have to be sent to the Technology Repair Center for these repairs. Included are the disassembly, repair/replacement and reassembly of the front panel assembly.

6-13. FRONT PANEL ASSEMBLY. The front panel assembly contains the A9 circuit card, encoder assembly, keypad switch assemblies S3 and S4 and other mechanically operated controls that can deteriorate during normal use. Refer to Figures 6-16 and 6-17 for illustrations of the encoder and front panel assembly. Since these sub-assemblies are entwined around the A9 circuit card (part of front panel assembly) a procedure for disassembly and reassembly of the complete front panel assembly is presented in step-by-step procedures. These procedures permit the removal of any sub-assembly desired by completing the step-by-step procedures to the point that removes that complete sub-assembly. Inspection, repair and replacement procedures are then presented to properly restore the affected parts. Reassembly instructions provided may then be used to reassemble the front panel assembly.

## a. Dismantling and Disassembly.

(1) Remove top and bottom cover plates from the Receiver by loosening six quarter turn fasteners and disengaging each cover from the slot in the front panel by sliding to the rear.
(2) Remove connectors A9W1P1 from A6A2 and W13P2, W15P2, W19P1 and W14P2 from the A9 circuit card.
(3) Remove four 10-32 screws and nylon washers securing the front panel assembly to the gussets of the Receiver and carefully fold the assembly face down.
(4) Remove cable W20PI connector from the A10 module connector A10J2.
(5) Remove the ribbon cables from A9P2 and A9P3 connectors that connect the key-board switch assemblies S3 and S4 to the A9 circuit card. Do not attempt to remove the connectors.
(6) Remove the 6-32 screw, lock washer and flat washer from encoder disk, remove the disk, then remove spring washer and flat washer under disk.
(7) Encoder knob, shaft, retainer ring, shim washer and flat washer can now be removed together from front of front panel. Shim washer(s) and flat washer are loose and may be removed from disk end of shaft. Retainer ring may be removed from groove of shaft with retainer ring tool or other suitable device. Encoder knob may be removed from shaft by loosening two allen head screws, accessible through holes in circumference of knob.
(8) Remove knob from each IF and AF gain control by loosening allen head screw in each knob.
(9) Remove $3 / 8 \times 32$ hex nut, and internal tooth washer from each IF and AF GAIN control and from PHONES jack.
(10) Remove $/ 2 \times 20$ hex nut, spring washer, flat washer and nylon washer from encoder spacer-bushing, located on component side of A9 circuit card.
(11) Remove ten 440 screws, lock washers and flat washers securing the A9 circuit card and carefully lift the circuit card away from the front panel, making sure the IF and AF GAIN controls and PHONES jack separates from the front panel. If these three items are to be removed from the circuit card their leads must be unsoldered.
(12) Disconnect W1P1 from A9J8 on the A8 circuit card.
(13) From the front of the front panel remove the $1 / 2 \times 20$ hex nut, spring washer and flat washer securing the encoder spacer-bushing to the front panel and remove the bushing.
(14) From the front of the front panel remove the $1 / 4 \times 24$ hex nut, and internal tooth washer securing the POWER-ON switch S 1 to the front panel and remove the switch, solder lug and cable assembly W20. NOTE: Cable W20 includes switch S1.
(15) Remove three (3) 440 hex nuts, lock washers and flat washers from the edge lighting assembly and remove the assembly.
(16) Remove five 440 hex nuts, lock washers and flat washers from light diffuser plate and lift the plate away from front panel, then lift switch assemblies S3 and S4 away from front panel.

## b. Inspection, Repair and or Replacement

(1) Encoder Assembly.
(a) Inspect the encoder spacer-bushing and encoder shaft for excessive wear or damage. Replace both of these items if excessive wear or damage is present.
(b) Inspect the flat washers, shim washers, spring washer and nylon washer. Spring washer must be able to take up end play in encoder shaft. Replace broken, damaged or worn items.
(c) Inspect the encoder disk for scratches, mars, dents or dirt on the mirrored orblack surfaces. Also lay the disk on a flat surface and check for warpage. Replace the disk if it is warped or the mirrored or black surfaces are in any way impaired. Scratches or dark spots on the mirrored surfaces or light spots or scratches on the black areas will affect the encoder operation.
(d) Temporarily assemble the retainer ring, flat washer, shim washer, spacer bushing (shoulder toward disk end of shaft), flat washer, spring washer, disk, flat washer, lock washer and 6-32 screw on the encoder shaft in the order given. Check the end play of the shaft within the spacer-bushing by holding the spacer-bushing and pushing on each end of the shaft. No end play should be detected when pushing on knob end of shaft, but approximately 0.030 inch of end play will be present when the shaft is pushed from the disk end; however, the spring washer should cause the shaft to return to its original position when released. If end play is too great shaft will float in the bushing and either additional or thicker shim washer must be added. If end play is insufficient shaft will not spin freely in bushing, and less shim is required. Disassemble the unit when correct end play is obtained.
(e) Inspect the encoder knob for damage, replace if necessary.
(2) Controls, POWER-ON Switch and PHONES Jack.
(a) Check the IF and AF GAIN controls for freedom of operation, signs of excessive heat or other damage. Using an ohmmeter check the resistance of each control. The IF GAIN control (RI) should be 50 K ohms $\pm 10 \%$. The AF GAIN control (R2) should be 25 K ohms $\pm 10 \%$. Replace either or both controls if they are found to be substandard.
(b) Inspect the AF and IF GAIN control knobs for any signs of damage, replace as required.
(c) Inspect the PHONES Jack for excessive wear or damage, replace if necessary.
(d) Inspect the POWER-ON switch for freedom of operation, signs of arcing or other damage, replace if necessary.
(3) Keyboard Switch Assemblies S3 and S4.
(a) Visually inspect the switch assemblies for wear, chipping, los of nomenclature or other physical damage. Switch assemblies are non-repairable and must be replaced if damage has occurred.
(b) Using an ohmmeter with one lead connected to pin I (common) of the switch connector and the other lead alternately connected to pins 2 through 17, check the resistance, both with the appropriate key depressed and open. Resistance should be essentially 0 ohms when depressed and infinite ohms when open. If any keys do not show continuity when pressed and open when released, the entire assembly must be replaced.


Figure 6-16. Front Panel (Assembly, Partially Disassembled.)


Figure 6-17. Encoder Assembly, Exploded View.
(4) Edge Lighting Assembly.
(a) Visually inspect the assembly for any signs of damage. Also insect the lead wires and connector W1P1. The assembly is nonrepairable and must be replaced if damage has occurred.
(b) Connect a 15 volt supply through connector W1P1 to the edge lighting. All lamps should illuminate. If any lamps do not illuminate, replace the lamp(s).
(5) Circuit Card Assembly A9.
(a) Check components for excessive heat, loose connections, corroded leads, and other damage.
(b) Check LCDs for looseness or damage to the units.
(c) Check connectors for damaged or bent pins, loose ©nnections, corrosion or other damage.
(d) Check the circuit card for damaged tracks, excessive heat, corrosion or other signs of deterioration.
c. Reassembly.
(1) Install keyboard switch assembly S3 in the opening on the right side (facing front of panel) with keys toward the front and with the ribbon cable extending toward the top of the panel. Place keyboard switch assembly S4 in the opening on the left side (facing the front panel) with keys toward the front and ribbon cable extending toward the top of the panel.
(2) Install light diffuser over studs in front panel so that windows match the openings in front panel. Install five (5) 4-40 hex nuts, lock washers and flat washers.
(3) Install the edge lighting assembly over the studs so that its windows fit the front panel openings. Install three (3) 4-40 hex nuts, lock washers and flat washers. Connect WIPI to A9J8 on circuit card A9.
(4) Cable W20 has been removed from front panel. Replace as follows: Install solder lug over shank of switch, then install shank through $1 / 4$ inch round hole in left side of front panel. Be sure switch is placed so that ON is toward top of panel. Secure with internal tooth washer and $\mathrm{V} 1 / 4 \times 24$ hex nut.
(5) Install the shoulderless end (shoulder toward circuit card A9) of the encoder spacer-bushing through the I/2 inch round hole located near the center of the front panel. Secure with flat washer, spring washer and $1 / 2 \times 20$ hex nut.
(6) If AF or IF GAIN control or PHONES jack leads have been unsoldered they must be esoldered. Refer to schematic diagram in Chapter \& for proper lead identification. Insert the shaft of these three items into their respective holes in the front panel and place circuit card A9 on the standoffs extending from the back of the front panel. Be sure encoder spacer-bushing is fitted through hole in circuit card. Secure the circuit card with ten flat washers, lock washers and 440 screws.

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(7) Place the $/ 2$ inch nylon washer, flat washer and spring wæher over the threads of the spacer-bushing in the order given and secure them with the $1 / 2 \times 20$ hex nut. Tighten only finger tight.
(8) Secure the IF and AF GAIN controls and PHONES jack each with an internal tooth washer and $3 / 8 \times 32$ hex nut.
(9) Install the IF and AF GAIN control knobs and secure them by tightening the alien head screw in each knob.
(10) Assemble the retainer ring, flat washer and shim washer on the encoder shaft, in the order given. Install the encoder shaft, from the front of the front panel, through the spacer-bushing. Install flat washer, spring washer, disk (mirror surfaces toward A9 circuit card), flat washer and lock washer in the order given. Secure with 6-32 screw.
(11) Install the encoder knob and secure by tightening two allen head screws in the knob. Inspect the encoder assembly for freedom to spin and for wobble in the disk.
(12) Install the ribbon cables into their respective connectors P2 and P3 located on the A9 circuit card.
(13) Connect connectors W14P2, W19P1,W15P2 and W13P2 to their respective connectors on the A9 circuit card. Connect connector A9W1 PI to its connector on the A6A2 circuit card, connect cable W20P1 connector to A10J2 on the A10 module.
(14) Install the front panel assembly to the gussets of the Receiver with four nylon washers and 10-32 screws.
(15) Install the top and bottom cover plates on the Receiver and secure each cover with six quarter turn fasteners.

## CHAPTER 7 ILLUSTRATED PARTS BREAKDOWN

7-1. INTRODUCTION. This illustrated parts breakdown (IPB) provides parts information on the R-2174(P)/URR Radio Receiver. The information is presented in four sections in accordance with Specification MIL-M-38807. The four sections include: Section I, Introduction; Section II, Maintenance Parts List (MPL); Section III, Numerical Index; and Section IV, Reference Designation Index. Section I, Introduction, describes the information contained in the IPB, the arrangement of that information and how to use it. Section II, Maintenance Parts List, lists all parts in disassembly sequence along with illustrations that show that sequence. Section III, Numerical Index, lists the same parts contained in Section II but in alpha-numerical sequence. Section IV lists all parts with reference designations in a reference designation sequence.

## Section I. INSTALLATION LOGISTICS

7-2. MODELS COVERED. The Illustrated Parts Breakdown covers only one model of the R-2174(P)/URR Radio Receiver.

7-3. PARTS LISTED. In general, the assemblies and parts installed at the time the end item was manufactured are listed and identified in the manual. When an assembly or part (including vendor items), which is different from the original was installed during the manufacture of later items, series, or blocks, all assemblies and parts are listed (and "Usable on Coded"). However, when the original assembly or part does not have continued application (no spares of replacement), only the preferred assembly or part is listed. Also, when an assembly or part was installed during modification, and the original does not have continued application, only the preferred item is listed. Interchangeable and substitute assemblies and parts, subsequently authorized by the government, are not listed in this manual; such items are identified by information available through the Interchangeable and Substitute (I\&S) Data Systems. Refer toTO 00-25-184. When a standard size part can be replaced with an oversize or undersize part, the latter parts, showing sizes, are also listed. Repair parts kits and quick change units are listed when they are available for replacement.

7-4. SIMILAR ASSEMBLIES. All assemblies in the R-2174(P)/URR Radio Receiver are sufficiently different to require separate listings for each assembly and are so listed in disassembly sequence.

7-5. QUICK CHANGE UNITS. Fifteen Bandpass Crystal Filters are available for plug-in to seven filter slots in the Receiver. All fifteen filters are listed as quick change units while the filters actually installed in the Receiver are left to the discretion of the operator.

7-6. SYMBOLS AND ABBREVIATIONS Table 7-1 lists symbols and abbreviations used throughout the illustrated parts breakdown. The table lists abbreviations, symbols, and reference designators in that order.

MANUFACTURERS CODES. Section II, Maintenance Parts List, lists along with each part number the manufacturer in a five digit code. This number listed under FSCM (Federal Supply Code for Manufacturers) is decoded in Table 7-2. This table lists the code in numerical sequence and defines the name and address of the manufacturer.

TABLE 7-1. SYMBOLS AND ABBREVIATIONS

| Symbol or Abbreviation | Description | Symbol or Abbreviation | Description |
| :---: | :---: | :---: | :---: |
| AC | Alternating Current | uh | Micro henry ( $10^{-6}$ ) |
| AF | Audio Frequency |  | Decal, logo |
| AM | Amplitude Modulation | \# | Number, size |
| AP | Attaching Part |  |  |
| BCD | Binary Coded Decimal | REFERENCE DESIGNATIONS |  |
| BFO | Beat Frequency Oscillator |  |  |
| Coax | Coaxial-used with Cable | A | Printed Circuit Boards |
| DC | Direct Current | BT | Battery |
| DPDT | Double Pole-Double Throw | C | Capacitor |
| EPROM | Erasable Programmable Read | CR | Diode |
|  | Only Memory | DS | Indicator |
| FM | Frequency Modulation | E | Terminal |
| Hex | Hexagonal | F | Fuse |
| IF | Intermediate Frequency | FL | Filter |
| ISB | Independent Side Band | J | Connector, Jack |
| K ohms | thousand ohms ( $10^{3}$ ) | L | Inductor, Choke, or Coil |
| LO | Local Oscillator | LK | Link |
| LPF | Low Pass Filter | P | Connector, Plug |
| meg ohms | million ohms ( $10^{6}$ ) | Q | Transistor |
| pf | Pico farrads ( $10^{-12}$ ) | R | Resistor |
| RAM | Random Access Memory | S | Switch |
| ROM | Read Only Memory | T | Transformer |
| RF | Radio Frequency | U | Integrated Circuit |
| SMI | System Memory Interface | W | Cable |
| UART | Universal Asynchronous | XF | Fuse Holder |
|  | Receiver Transmitter | XU | Socket, Integrated Circuit |
| WVDC uf | Working Volts, Direct Current Micro farrads $\left(10^{-6}\right)$ |  |  |

TABLE 7-2. FEDERAL SUPPLY CODES FOR MANUFACTURERS

| FSCM | Name and Address | FSCM | Name and Address |
| :---: | :---: | :---: | :---: |
| 01295 | Texas Instruments, Inc. Semiconductor Components Div. Dallas, TX 75222 | 07263 | Fairchild Semiconductor, A Division of Fairchild Camera and Instr. Corp. 464 Ellis St. <br> Mountain View, CA 94040 |
| 02735 | RCA Corp. <br> Solid State Division <br> Route 202 <br> Somerville, NJ 08876 | 07374 | Optron Corp. <br> 50 Fitch Street <br> New Haven, CT 06515 |
| 04222 | Aerovox Corp. <br> P.O. Box 867, Aerovox Road Myrtle Beach, SC 29577 | 07556 | Callabro Plastics, Inc. 8738 Westchester Pike Upper Darby, PA 19082 |
| 04713 | Motorola, Inc. <br> Semiconductor Products Div. <br> 5005 E.McDowell Road <br> Phoenix, AZ 85008 | 07623 | Eck and Kribs Scientific Laboratory Glass Apparatus Inc. <br> 2709 40th Ave. <br> Long Island City, NY 11101 |
| 04729 | Universal Components Corp. Rock Ave. \& Route 22 Green Brook, NJ 08812 | 08050 | Dexter Lock Div. of Kysor Industrial Corp. 1601 Madison Ave SE Grand Rapids, MI 49502 |
| 05245 | Components Corp. 2857 N.Halstead St. Chicago, IL 60657 | 08523 | Metal Craft Inc. 34 Burgess Wayne, NJ 07470 |
| 05820 | Wakefield Engineering Inc. Audubon Rd. <br> Wakefield, MA 01880 | 09922 | Burndy Corp. <br> Richards Ave. <br> Norwalk, CT 06852 |
| 05972 | Loctite Corp. 705 N.Mountain Road Newington, CT 06111 | 13103 | Thermoloy Co. 8717 Diplomacy Row Dallas, TX 75247 |
| 06540 | Amatom Electronic Hardware Division of Mite Corp. <br> New Rockville, NY | 13257 | Esna Ltd. <br> 10 Esna Park Drive Markham, Ontario, Canada |
| 06776 | Robinson Nugent, Inc. P.O. Box 470 800 E.8th St. <br> New Albany, IN 47150 | 14655 | Cornell Dubilier Electronics Div. of Federal Pacific Electric Co. Government Contracts Dept. 150 Ave L |
| 06915 | Richco Plastics Co. 5825 N. Tripp Ave. Chicago, IL 60646 | 16428 | Newark, NJ 07101 <br> Belden Corp. <br> P.O. Box 341 <br> Richmond, IN 47374 |

TABLE 7-2. FEDERAL SUPPLY CODES FOR MANUFACTURERS (Cont.)

| FSCM | Names and Addresses | FSCM | Names and Addresses |
| :---: | :---: | :---: | :---: |
| 17069 | Circuit Structures Lab. 3200 N. San Fernando Blvd. Burbank, CA 91504 | 28480 | Hewlett Packard Co. Corporate Hdqtrs. 1501 Page Mill Road Palo Alto, CA 94304 |
| 17896 | Jermyn Industries Mfg. Div. 712 Montgomery Street San Francisco, CA 94111 | 30983 | Electra/Midland Corp. A North American Phillips Co. 11468 Sorrento Valley Road San Diego, CA 92121 |
| 18324 | Signetics Corp. <br> 811 E. Arques <br> Sunnyvale, CA 94086 | 31148 | PMI Corp. 11335 Folsom Blvd. Rancho Cordova, CA 95670 |
| 22526 | Berg Electronics Inc. <br> Youk Expressway <br> New Cumberland, PA 17070 | 31433 | Union Carbide Corp. <br> Material Systems Div. Components Dept. <br> Highway 276 SE <br> Greenville, SC 29606 |
| 23386 | Racal Communications Inc. 5 Research Place Rockville, MD 20850 | 32293 | Intersil Inc. 10900 N. Tantau Ave. Cupertino, CA 95014 |
| 24355 | Analog Devices, Inc. P.O. Box 280, Route 1 Industrial Park Norwood, MA 02062 | 32997 | Bovins Inc. Trimpot Products Division 1200 Columbia Ave. Riverside, CA 92507 |
| 24446 | General Electric Co. <br> 1 River Road <br> Schenectady, NY 12305 | 34335 | Advanced Micro Devices 901 Thompson Place Sunnyvale, CA 94086 |
| 27014 | National Semiconductor Corp. 2950 San Ysidro Way Santa Clara, CA 95051 | 34649 | Intel Corp. 365 Middlefield Rd. Mountain View, CA 94040 |
| 27193 | Cutler Hammer Inc. Specialty Products Div. 420 N. 27th Street Milwaukee, WI 53216 | 49956 | Raytheon Corp. 141 Spring Street Lexington, MA 02173 |
| 27440 | Industrial Screw Products Co. 2238 Purdue Ave. <br> Los Angeles, CA 90064 | 50088 | Mostek Corp. 1400 Upperfield Drive Carrollton, TX 75006 |
| 28198 | Positronic Industries Inc. 1906 S. Stewart Springfield, MO 65804 | 52072 | Circuit Assembly Corp. 3169 Redhill Ave. Costa Mesa, CA 92626 |
|  |  | 52673 | KSW Electronics Corp. <br> South Bedford St. <br> Burlington, MA 01803 |

TABLE 7-2. FEDERAL SUPPLY CODES FOR MANUFACTURERS (Cont.)

| FSCM | Names and Addresses | FSCM | Names and Addresses |
| :---: | :---: | :---: | :---: |
| 52763 | Stettner Trush Inc. 67 Albany Street Cazenovia, NY 13035 | 81349 | Military Specifications <br> Promulgated by Military Departments/ Agenices under authority of Defense Standardization Manual 41203-M |
| 52783 | Gyrex Corp. <br> 436 E. Gutierrez | 83003 | Varo Inc. |
|  | Santa Barbara, CA 93101 |  | 800 W. Garland Ave. <br> Garland, TX 75040 |
| 53848 | Siliconix Inc. <br> 2201 Laurelwood Rd. <br> Santa Clara, CA 95054 | 83330 | Smith Herman H Inc. 812 Snediker Ave. Brooklyn, NY 11207 |
| 56289 | Sprague Electric Co. North Adams, MA 01247 | 86797 | Rogan Bros.Inc. 8031 N. Monticello |
| 71400 | Bussmann Mfg. Div.of McGraw Edison Co. 2536 W. University Street St. Louis, MO 63107 | 86928 | Snokie, IL 60076 <br> Seastrom Mfg. Co. Inc. <br> 701 Sonora Ave. <br> Glendale, CA 91201 |
| 71468 | ITT Cannon Electric 666 E. Dyer Road Santa Ana, CA 92702 | 91293 | Johanson Mfg. Co. <br> P.O. Box 329 <br> Boonton, NJ 07005 |
| 71590 | Centralab Electronics Division of Globe Union Inc. 5757 N. Green Bay Ave. Milwaukee, WI 53201 | 91637 | Dale Electronics Inc. P.O. Box 609 Columbus, NE 68601 |
| 73734 | Federal Screw Products Inc. 3917 N. Kenzie Ave. Chicago, IL 60618 | 91836 | Kings Electronics Co. Inc. 40 Marbledale Road Tuckahoe, MY 10707 |
| 73743 | Fischer Special Mfg. Co. 446 Morgan Street Cincinnati, OH 45206 | 95987 | Weckesser Co. Inc. 4444 West Irving Park Rd. Chicago, IL 60641 |
| 75037 | Minnesota Mining and Mfg. Co. <br> Electro Products Div. <br> 3M Center <br> St. Paul, MN 55101 | 96906 | Military Standards <br> Promulgated by Military Departments under authority of Defense Standardization Manual 41203-M |
| 75915 | Littlefuse Inc. 800 E. Northwest Hwy. Des Plaines, IL 60016 | 97244 | Mallory Metallurgical Co. 3029 E. Washington Street Indianapolis, IN 46206 |
| 78912 | Garlock Inc. Plastics Div. 602 N. 10th Street Camden, NJ 08101 | 98291 | Sealectro Corp. 225 Hoyt Mamaroneck, NY 10544 |

7-8. USABLE ON CODES. The Usable On Code column in the Section II parts list is included to show application of parts to different models, configurations, types, etc. of the Radio Receiver. The original R-2174 parts are shown with no code in this column. Differences in the R-2174A version are shown by an A in the Useable On Codes column.

7-9. SOURCE, MAINTENANCE, AND RECOVERABILITY (SMR) CODES. Definitions of applicable source, maintenance, and recoverability (SMR) codes are set forth in TO 00-25-195. These codes for all parts listed are included in the MPL in Section II under the SMR column.

7-10. FINDING PART NUMBER, ILLUSTRATION, DESCRIPTION. The IPB includes part numbers, illustrations of parts, description of parts, reference designations of electrical parts, units per assembly, quantity per end item, codes for manufacturers of each part listed and SMR codes. Figure 7-0 illustrates a technique to aid the technician in locating information in the manual. This illustration shows how a part number can be found through the table of contents and illustrations or how the illustration of a part can be located when a part number is known. When a part number is located in the MPL, its description, units per assembly, SMR and FSCM codes will be in the same line as the part number. If the quantity per end item is desired refer to the part number in the Numerical Index. If only a reference designation is known refer to that Section which lists the Figure and index number for each electrical part in alpha-numerical sequence of reference designations.

SECTION II MAINTENANCE
PARTS LIST (MPL)

## HOW TO USE THE ILLUSTRATED PARTS BREAKDOWN



Figure 7-0. Finding Part Number, Illustrated Description


Figure 7-1. Receiver Assembly, Radio, R-2174(P)/URR (Sheet 1 of 5)

FIGURE 7-1. RECEIVER ASSEMBLY, RADIO R-2174(P)/URR


FIGURE 7-1. RECEIVER ASSEMBLY RADIO R-2174 (P)/URR (Cont.)


FIGURE 7-1. RECEIVER ASSEMBLY RADIO R-2174 (P)/URR (Cont.)



Figure 7-1. Receiver Assembly, Radio, R-2174(P)/URR (Sheet 2 of 5)

FIGURE 7-1. RECEIVER ASSEMBLY RADIO R-2174 (P)/URR (Cont.)


FIGURE 7-1. RECEIVER ASSEMBLY RADIO R-2174 (P)/URR (Cont.)



Figure 7-1. Receiver Assembly, Radio, R-2174(P)/URR (Sheet 3 of 5)

FIGURE 7-1. RECEIVER ASSEMBLY RADIO R-2174 (P)/URR (Cont.)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ |  | USABLE ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-1- $\begin{array}{rr} \\ & / 3-20 \\ & / 3-44 \\ & / 3-51 \\ & / 3-54 \\ & / 3-56 \\ & \\ & \\ & \\ \\ & 3.57 \\ & \\ & 3-58\end{array}$ |  |  |  |  |  |  |
|  | 08563 | 23386 | - CABLE ASSEMBLY, W15, Control | 1 |  | MOOZZN |
|  | 08376 | 23386 | - CABLE ASSEMBLY,W13,Control ..... | 1 |  | MOOZZN |
|  | 08558 | 23386 | - CABLE ASSEMBLY, W14, Control.... | 1 |  | MOOZZN |
|  | 08569 | 23386 | - CABLE ASSEMBLY, W19, DC Power. | 1 |  | MOOZZN |
|  | MS35234-63 | 96906 | - SCREW, Machine, Pan Head ........... $10-32 \times 1 / 2 \text { (AP) }$ | 4 |  | PAOZZN |
|  | 5610-280-031 | 86928 | - WASHER, Nylon, $\qquad$ <br> 0.200 inches I.D (AP) | 4 |  | PAOZZN |
|  | 08557 | 23386 | - KNOB, Encoder, Modified ................ | 1 |  | PAOZZN |
|  | 08538 | 23386 | - DISK, Encoder | 1 |  | PAFZZN |
|  | MS35233-27 | 96906 | - SCREW, Machine, Pan Head $\qquad$ $6-32 \times 5 / 16 \text { (AP) }$ | 1 |  | PAOZZN |
|  | MS35338-79 | 81349 | - WASHER, Split Lock $\qquad$ No. 6 (AP) | 1 |  | PAOZZN |
|  | MS15795-305 | 81349 | - WASHER, Flat, No. 6 (AP) .............. | 1 |  | PAOZZN |
|  | CYL-1400-250 | 08523 | - WASHER, Spring (AP) ................... | 1 |  | PAOZZN |
|  | 5710-54-25 | 86928 | - WASHER, Flat (AP) ........................ | 1 |  | PAOZZN |
| /3-60 | 08553 | 23386 | - SHAFT, Encoder............................ | 1 |  | XBOZZN |
| /3-61 | 3100-25 | 27440 | - RING, Retainer, Shaft $\qquad$ <br> Encoder (AP) | 1 |  | PAOZZN |
|  | 5710-54-25 | 86928 | - WASHER, Flat (AP) ....................... | 1 |  | PAOZZN |
|  | 5720-21 | 86928 | - WASHER, Shim, 1/4 inch ................ | AR |  | PAFZZN |
| /3-62 | 08554 | 23386 | - BUSHING, Spacer, Encoder ............ | 1 |  | XBOZZN |
| /3-63 | 76050-NP | 73734 | - NUT, Hex $1 / 2 \times 20 \times 1 / 8$ $\qquad$ inch thick (AP) | 2 |  | PAOZZN |
|  | 5806-28-1 | 86928 | - WASHER, Spring (AP) ................... | 2 |  | PAOZZN |
|  | 5710-94-015 | 86928 | - WASHER, Flat (AP) | 2 |  | PAOZZN |
|  | 5610-84-032 | 86928 | - WASHER, Nylon (AP) | 1 |  | PAOZZN |
| /3-64 | 08570 | 23386 | - CABLE ASSEMBLY, W20 $\qquad$ <br> AC line switching | 1 |  | MOOZZN |
| /3-65 | JMT-223 | 55459 | -• SWITCH, S1, Miniature $\qquad$ DPDT | 1 |  | PAOZZN |
| /3-66 |  |  | -• NUT, Hex, 1/4 x 24 (AP) ................. |  |  |  |
| /3-67 |  |  | -• LUG, Solder, 1/4 inch I.D ................ |  |  |  |
| /3-68 | RB-67-15K-7 | $86797$ | - KNOB, Potentiometer, R ................. | $1$ |  | PAOZZN |
| /3-69 | 08552 | 23386 | - POTENTIOMETER, IF Gain ............. <br> RI, 50K ohms | 1 |  | PAFZZN |
| /3-70 | 9002-NP | 73734 | - NUT, $3 / 8 \times 32$ (AP). ........................ | 1 |  | PAOZZN |
|  | 30-350 | 73734 | - WASHER, Internal Tooth $\qquad$ 3/8 (AP) | 1 |  | PAOZZN |
| $\begin{aligned} & / 3-71 \\ & / 3-72 \end{aligned}$ | RB-67-15K-7 | 86797 | - KNOB, Potentiometer, R2 ................ | 1 |  | PAOZZN |
|  | 08551 | 23386 | - POTENTIOMETER, AF Gain $\qquad$ <br> R2, 25 K ohms | 1 |  | PAFZZN |
| /3-73 | $\begin{aligned} & 9002-N P \\ & 30-350 \end{aligned}$ | $\begin{aligned} & 73734 \\ & 73734 \end{aligned}$ | NUT, $3 / 8 \times 32$ (AP) $\qquad$ <br> - WASHER Internal Tooth | $1$ |  | $\begin{aligned} & \text { PAOZZN } \\ & \text { PAOZZN } \end{aligned}$ |
|  |  |  | $3 / 8 \text { (AP) }$ |  |  |  |

FIGURE 7-1. RECEIVER ASSEMBLY RADIO R-2174 (P)/URR (Cont.)



Figure 7-1. Receiver Assembly, Radio, R-2174(P)/URR (Sheet 4 of 5)

FIGURE 7-1. RECEIVER ASSEMBLY RADIO R-2174 (P)/URR (Cont.)



Figure 7-1. Receiver Assembly, Radio, R-2174(P)/URR (Sheet 5 of 5)

FIGURE 7-1. RECEIVER ASSEMBLY RADIO R-2174 (P)/URR (Cont.)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ |  | USABLE <br> ON CODE | $\begin{aligned} & \text { SMR } \\ & \text { CODE } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-1 |  |  |  |  |  |  |
| /5-38 | 08492 | 23386 | - PLATE, Blanking, for A6AI $\qquad$ Position | 1 |  | XBOZZN |
| /5-39 | MS35233-14 | 96906 | - SCREW, Machine, Pan Head, $440 \times 5 / 16(A P)$ | 2 |  | PAOZZN |
| /5-92 | MS35233-28 | 96906 | - SCREW, Machine, Pan Head $6-32 \times 3 / 8(\mathrm{AP})$ | 5 |  | PAOZZN |
|  | MS35338-79 | 81349 | - WASHER, Split Lock, No. 6 (AP) ....... | 5 |  | PAOZZN |
|  | MS 15795-305 | 81349 | - WASHER, Flat, No. 6 (AP) ............... | 5 |  | PAOZZN |
| /5-95 | 08445 | 23386 | - PANEL, Rear............................... | 1 |  | XBOZZN |
| /5-96 | MS35233-43 | 96906 | - SCREW, Machine, Pan Head $8-32 \times 3 / 8 \text { (AP) }$ | 4 |  | PAOZZN |
|  | MS35338-80 | 81349 | - WASHER, Split Lock, No. 8 (AP) ....... | 4 |  | PAOZZN |
|  | MS15795-307 | 81349 | - WASHER, Flat, No. 8 (AP) ............... | 4 |  | PAOZZN |
| /5-97 | KC19-110 | 91836 | - CONNECTOR, J7, BNC | $1$ |  | PAOZZN |
| $/ 5-98$ |  |  | -. NUT, (AP) $\qquad$ <br> .. WASHER, (AP) | $1$ |  |  |
| /5-99 | GF326 | 27193 | - SWITCH, Internal/External, ................................ | 1 |  | PAFZZN |
|  |  |  | S2, Slide, DPDT |  |  |  |
| /5-100 | MS35233-14 | 96906 | - SCREW, Machine, Pan Head,........... $440 \times 5 / 16 \text { (AP) }$ | 2 |  | PAOZZN |
|  | MS 35338-78 | 81349 | - WASHER, Split Lock, No. 4 (AP) ....... | 2 |  | PAOZZN |
|  | MS15795-303 | 96906 | - WASHER, Flat No. 4 (AP) ................ | 2 |  | PAOZZN |
| /5-101 | 1488.4 | 83330 | - PLUG, Solder, No. 4 ....................... | 1 |  | PAFZZN |
| /5-102 | MS35233-14 | 96906 | - SCREW, Machine, Pan Head $\qquad$ $4-40 \times 5 / 16(A P)$ | 1 |  | PAOZZN |
|  | MS35649-44 | 81349 | - NUT, Hex, 440 (AP) | 1 |  | PAOZZN |
|  | MS35338-78 | 81349 | - WASHER, Split Lock, No. 4 (AP)........ | 1 |  | PAOZZN |
| $/ 5-103$ | UG-606/U | 81349 | - CONNECTOR, JI, N-BNC | $1$ |  | PAOZZN |
| $/ 5-104$ |  |  | - NUT, (AP) <br> - WASHER (AP) | $\begin{aligned} & 1 \\ & 1 \end{aligned}$ |  |  |
| /5-105 | KC19-110 | 91836 | - CONNECTOR, J2, BNC .................... | 1 |  | PAOZZN |
|  |  |  | Bulkhead |  |  |  |
| /5-106 |  |  | - NUT, (AP) .................................... | 1 |  |  |
|  |  |  | - WASHER, (AP) ............................ | 1 |  |  |
| /5-107 | 3483-1000 | 75037 | - CONNECTOR, J3,D <br> Subminiature, 25 Pin | 1 |  | XA |
| /5-108 | MS35233-14 | 96906 | - SCREW, Machine, Pan Head $440 \times 5 / 16(A P)$ | 2 |  | PAOZZN |
|  | MS35649.44 | 81349 | - NUT, flex, 440, (AP) ........................ | 2 |  | PAOZZN |
|  | MS35338-78 | 81349 | - WASHER, Split Lock, No 4 (AP) ....... | 2 |  | PAOZZN |
|  | MS 15795-303 | 96906 | - WASHER, Flat, No. 4 (AP) ............... | 2 |  | PAOZZN |
| /5-109 | D110277 | 71468 | - LATCH, Connector, J3 .................... | 2 |  | PAOZZN |
| /5-110 | 8072-NP | 73743 | - NUT, Ground Lug, Knurled, 10-32 | 1 |  | PAFZZN i |
| /5-111 | MS15795 | 81349 | - WASHER, Flat, No 10 .................... | 2 |  | PAOZZN |
| /5-112 | MS35234-65 | 96906 | - SCREW, Ground Lug, Machine Pan Head, $10-32 \times 3 / 4$ | 1 |  | PAOZZN |
| /5-113 | MS35650-104 MS35338-81 | $\begin{aligned} & 81349 \\ & 81349 \end{aligned}$ | - NUT, Hex, 10-32 (AP) $\qquad$ <br> - WASHER Split Lock No.10(AP) | $\begin{aligned} & 1 \\ & 1 \end{aligned}$ |  | $\begin{aligned} & \text { PAOZZN } \\ & \text { PAOZZN } \end{aligned}$ |

FIGURE 7-1, RECEIVER ASSEMBLY RADIO R-2174 (P)/URR (Cont.)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ | UNITS PER ASSY | USABLE <br> ON CODE | $\begin{aligned} & \text { SMR } \\ & \text { CODE } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |
|  | MS24641-1 | 96906 | -SCREW, Self tapping, No. 2 (AP) | $2$ |  | $\begin{aligned} & \text { XBOZZN } \\ & \text { PAOZZN } \end{aligned}$ |
|  | 08468 | 23386 | -GUSSET, Right ................................. | 1 |  | XBOZZN |
|  | MS3523343 | 96906 | -SCREW, Machine, Pan Head, ................ $8-32 \times 3 / 8 \text { (AP) }$ | 4 |  | PAOZZN |
|  | 5806-10-1 | 86928 a | -WASHER, Spring, No. 8 (AP) ................ | 4 |  |  |
|  | FCS-10 | 95987 | -CLAMP, Cable | $10$ |  | PAOZZN |
|  | MS35249-20 | 96906 | -SCREW, Machine, Flat Head $440 \times 1 / 4(A P)$ | 1 |  | PAOZZN |
|  | 051-075-0000 | 98291 | -CONNECTOR, J5 and J6, $\qquad$ SMB-SMB | 2 |  | PAFZZN |
|  | BSF-IBBGP102M | 04222 | -CAPACITOR, Feed thru, $\qquad$ C1 thru C9, . 001 uf $\pm 20 \%$ | 9 |  | PADZZN |



Figure 7-2. Low Pass Filter Assembly, Module, A1

FIGURE 7-2. LOW PASS FILTER ASSEMBLY, A1

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ | $\begin{aligned} & \text { UNITS } \\ & \text { PER } \\ & \text { ASSY } \end{aligned}$ | USABLE ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-2- 08385 | 23386 |  | - LOW PASS FILTER | 1 |  | PAFLDT |
| -1 | 08470 | 23386 | .. CASE, Low Pass Filter Module ......... | 1 |  | XBOZZN |
| -2 | MS35233-13 | 96906 | - •SCREW, Machine, Pan Head, .......... $4-40 \times \%(A P)$ | 1 |  | PAOZZN |
|  | MS35338-78 | 81349 | -• WASHER, Split Lock, $\qquad$ No. 4 (AP) | 1 |  | PAOZZN |
|  | MS35233-23 | 96906 | -• WASHER, Flat, No. 04 (AP) .............. | 1 |  |  |
| -3 | 08475 | 23386 | .. LOW PASS FILTER $\qquad$ ASSEMBLY, AI AI, Circuit Card | 1 |  | PADZZN |
| -4 | CMROSF111GODR | 81349 | ... CAPACITOR,Cl andC5, $\qquad$ Mica, 110 pf, 500 WVDC, $\pm 2 \%$ | 2 |  | PADZZN |
| -5 | 08477-1 | 23386 | ... COİL, LI, RF Variable ...................... | 1 |  | PADZZN |
| -6 | CMR05CIOOGODR | 81349 | ... CAPACITOR, C2, Mica $\qquad$ 10 pf, 500 WVDC, $\mathrm{t} 5 \%$ | 1 |  | PADZZN |
| -7 | CMRO5F151GODR | 81349 | ... CAPACITOR, C3, Mica, $\qquad$ 150 pf, 500 WVDC, +2\% |  |  | PADZZN |
| -8 | 08477-2 | 23386 | ... COIL, L2, RF Variable .................... | 1 |  | PADZZN |
| -9 | CMR05E750GODR | 81349 | ... CAPACITOR, C4, Mica $\qquad$ 75 pf, 500 WVDC, $\pm 5 \%$ | 1 |  | PADZZN |
| -10 | 08477-3 | 23386 | ... COIL, L3, RF Variable | $1$ |  |  |
| -11 | CMRO05F910GODR | $81349$ | ... CAPACITOR, C6, Mica $\qquad$ 91 pf, 500 WVDC, $\pm 2 \%$ | $1$ |  | PADZZN |
| -12 | CMRO5FI21GODR | 81349 | ... CAPACITOR, C7, Mica $\qquad$ 120 pf, 500 WVDC, $\pm 2 \%$ | 1 |  | PADZZN |
| -13 | 084774 | 23386 | ... COIL, L4, RF Variable .................... | 1 |  | PADZZN |
| -14 | CMR05E430GODR | 81349 | ... CAPACITOR, C8, Mica 43 pf, 500 WVDC, $\pm 2 \%$ | 1 |  | PADZZN |
| -15 | CMR05E820GODR | 81349 | ... CAPACITOR, C9, Mica $\qquad$ 82 pf, 500 WVDC, if $2 \%$ | 1 |  | PADZZN |
| -16 | 08480 | 23386 | ... CABLE ASSEMBLY, W1 ................. | 1 |  | MOOZZN |
| -17 | MS18034-4 | 96906 | ... TYRAP, Cable .............................. | 1 |  | PAOZZN |
| $-18$ | 08479-4 | $23386$ | ... CABLE ASSEMBLY, W2 | $1$ |  | MOOZZN |
| -19 | MS18034-4 |  | ... TYRAP, Cable |  |  |  |

Change 17-31


Figure 7-3. First Mixer Assembly, Circuit Card, A2

FIGURE 7-3. FIRST MIXER ASSEMBLY, CIRCUIT CARD, A2


FIGURE 7-3. FIRST MIXER ASSEMBLY, CIRCUIT CARD, A2 (Cont.)

|  <br> INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ |  | USABLE <br> ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-3 |  |  |  |  |  |  |
| -24 | CMR05F910GODR | 81349 | -• CAPACITOR, C24, Mica, ................... 91 pf, 500 WVDC, + 2\% | 1 |  | PADZZN |
| -25 | 08477-2 | 23386 | -• COIL ASSEMBLY, L11,RF,................ Variable | 1 |  | PADZZN |
| -25a | 08477-6 | 23386 | .. COIL ASSEMBLY,L11,RF,................. Variable | 1 | A | PADZZN |
| -26 | 08404 | 23386 | -• FILTER, FLI, 40.455.MHz ................. | 1 |  | PADZZN |
| -27 | MS35649-44 | 81349 | -• NUT, Hex, 4-40(AP) ........................ | 2 |  | PAOZZN |
|  | MS35338-78 | 81349 | -• WASHER, Splt Lock,No. 4 (AP) | 2 |  | PAOZZN |
|  | MS15795-303 | 96906 | -• WASHER, Flat, No.4(AP)................ | 1 |  | PAOZZN |
| -28 | 1488-4 | 83330 | -• LUG, cable, No. 4........................... | 1 |  | PAOZZN |
| -29 | CMR05E430GODR | 81349 | -. CAPACITOR,C26,Mica,43 pf 500 WVIC, $+2 \%$ | 1 |  | PADZZN |
| -30 | 08477-5 | 23386 | .. COIL ASSEMBLY, L12, RF................ Variable | 1 |  | PADZZN |
| -30a | 08477-7 | 23386 | -• COIL ASSEMBLY, L12, RF................ Variable | 1 | A | PADZZN |
| -31 | CMR05E680GODR | 81349 | -• CAPACITOR,C27,Mica,68 $\qquad$ pf 500 WVDC, + 2\% | 1 |  | PADZZN |
| -32-33 | 08504 | 23386 | -. MIXER BOX ASSEMBLY,.................. <br> UL, First Mixer | 1 |  | PADZZN |
|  | MS35649-44 | 81349 | -. NUT, Hex, 4-40(AP)........................ | 4 |  | PAOZZN |
| -33 | $\begin{aligned} & \text { MS35388-78 } \\ & \text { (AP) } \end{aligned}$ | 81349 | -• WASHER, Splt Lock,No.4................ | 4 |  | PAOZZN |
|  | MS15795-303 | 96906 | -• WASHER, Flat, No.4(AP). ............... | 4 |  | PAOZZN |
| -34 | MS39014101-1593 | 81349 | -. CAPACITOR,CII,C15,C18. C28 ,C43,Ceramic, 1 uf, 50 WVDC, + 20\% | 5 |  | PALZZN |
| -35 | RCR20G150JS | 81349 | -• ESISTOR,R6,Composition................. <br> 15 ohms, +5\%, 1/2 Watt | 1 |  | PADZZN |
| -36 | 2N4126 PNP | 04713 | -• TRANSISTOR,Q1,Silicon,. ............... | 1 |  | PAIZZN |
| -37 | LT1OK131 | 81349 | $\begin{aligned} & \text { •• CHOKE,L5,L7,LS,L9,LIO,.................. } \\ & \text { RF, } 10 \text { uH, }+10 \% \end{aligned}$ | 5 |  | PADZZN |
| -38 | MS39014102-1419 | 81349 | -. CAPACITOR,C17,C20,Cer,. .............. 1 uf, 500 WVDC, + 20\% | 2 |  | PADZZN |
| -39 | IN916B | 07623 | -• DIOLE, CR1 and CR2 ..................... | 2 |  | PADZZN |
| -40 | RLO7S562G | 81349 | .. RESISTOR,R2,Fixed Film.................. <br> $5.6 \mathrm{~K},+2 \%, 1 / 4$ Watt | 1 |  | PADZZN |
| -41 | K1935 | 13103 | -. HEAT SINK, Transistor $\qquad$ Q3 and Q4 | 2 |  | PAOZZN |
| -42 | 2N5160 | 04713 | .. TRANSISTOR,Q3,Silicon $\qquad$ High Power, PNP | 1 |  | PADZZN |
| -43 | MS35233-12 | 96906 | -• SCREW, Machine, Pan Head. $\qquad$ 4-40 x 3/16(AP) | 1 |  | PAOZZN |
|  | MS35338-78 | 81349 | -• WASHER, Splt Lock,No. 4 (AP) | 1 |  | PAOZZN |
|  | MS15795-303 | 96906 | -. WASHER, Flat, No.4(AP)................. | 1 |  | PAOZZN |
| -44 | 69011-1058 | 78912 | -. PAD, Transistor, Q3 ....................... | 1 |  | PAOZZN |
| -45 | JANTX2N3866 | 81349 | .. TRANSISTOR,Q4, Silicon, $\qquad$ High Power, PNP | 1 |  | PAOZZN |

FIGURE 7-3. FIRST MIXER ASSEMBLY, CIRCUIT CARD, A2 (Cont.)



Figure 7-4. Second Mixer Assembly, Circuit Card, A3

FIGURE 7-4. SECOND MIXER ASSEMBLY, CIRCUIT CARD, A3

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ | $\begin{gathered} \text { UNITS } \\ \text { PER } \\ \text { ASSY } \end{gathered}$ | USABLE <br> ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7.4- | 08191 | 23386 | - •SECOND MIXER ASSEMBLY, $\qquad$ Circuit Card,A3 | 1 |  | PAOLDT |
| -1 | CMR05E820GODR | 81349 | -.- CAPACITOR, C18, Mica, 82 pf, 500 WVDC, $\pm 2 \%$ | 1 |  | PADZZN |
| -2 | CMROS5F331GOD中 | 81349 | -•CAPACITOR,C19, Mica, 330 pf, ......... 500 WVDC, $\pm 20 \%$ | 1 |  | PADZZN |
| -3 | 08522 | 23386 | -•COIL,L5,L6,L7,L8,RF <br> Variable | 4 |  | PADZZN |
| -4 | 051-351-000-220 | 98291 | -• CONNECTOR, J2, RF, SMB ............ | 1 |  | PAOZZN |
| -5 | CY15C102M | 71590 | - CAPACITOR, C1, C2, C5, C6, C8,C10,CII,C12,C14,C20, C21, C22, C23, Mica, 1000 pf, 50 WVDC, $+20 \% 0$ | 3 |  | PADZZN |
| -6 | RL07S103G | 81349 | -. RESISTOR, R17, R36, Fixed Film, 10 K ohms, $\pm 2 \%$, Y4 watt | 2 |  | PADZZN |
| -7 | RL07S470G | 81349 | - RESISTOR, R30, R31, R32, R40, R44, R45, Fixed Film, 47 ohms, $\pm 2 \%$, Y4 watt | 6 |  | PADZZN |
| -8 | RL07S10IG | 81349 | -• RESISTOR, R2, R3, R20, R21 R27, R35, Fixed Film, 100 ohms, $\pm 2 \%$, Y4 watt | 6 |  | PADZZN |
| -9 | CMROSE680GODR | 81349 | -. CAPACITOR, C15, C16, C17, ........... Mica, 68 pf, 500 WVDC, $\pm 2 \%$ | 3 |  | PADZZN |
| -10 | RL07S102G | 81349 | -. RESISTOR, R19, R29, R33, Fixed Film, 1 K ohms, $\pm 2 \%$, 1/4 watt | 3 |  | PADZZN |
| -11 | MC1496L | 04713 | -• INTEGRATED CIRCUIT, U3, <br> Second Mixer | 1 |  | PADZZN |
| -12 | T362A105M035AS | 31433 | - CAPACITOR, C7, C30 Tantalum, 1 uf, 35 WVDC, $\pm 20 \%$ | 2 |  | PADZZN |
| -13 | RL07S561G | 81349 | -• RESISTOR, R15, Fixed Film, 560 ohms, $\pm 2 \%, 1 / 4$ watt | 1 |  | PADZZN |
| -14 | CA3046E | 02735 | .. INTEGRATED CIRCUIT, U1, ............ Transistor Array | 1 |  | PADZZN |
| -15 | KS8379 | 04713 • | - DIODE, CRI, Pin ............................ | , |  | PADZZN |
| -16 | RL07S 104G | 81349 | -. RESISTOR, R9, Fixed Film 100 K ohms, $\pm 2 \%$, Y4 watt | 1 |  | PADZZN |
| -17 | RL07S223G | 81349 | - RESISTOR, R5, R6, RI1 $\qquad$ <br> Fixed Film, 22K ohms, $\pm .2 \%$, 1/4 watt | 3 |  | PADZZN |
| -18 | RL07S472G | 81349 | -. RESISTOR, R7, R23, Fixed $\qquad$ Film, 4.7 K ohms, $\pm 2 \%, 1 / 4$ watt | 2 |  | PADZZN |
| -19 -20 | RL07S221G RL07S122G | 81349 81349 | -• RESISTOR, R1, R10, R13 $\qquad$ <br> R18, Fixed Film, 220 ohms, $\pm 2 \%, 1 / 4$ watt | 4 |  | PADZZN PADZZN |
| -20 | RL07S122G | 81349 | Film, 1.2K ohms, $\pm 2 \%, 1 / 4$ watt | 2 |  | PADZZN |

FIGURE 7-4. SECOND MIXER ASSEMBLY, CIRCUIT CARD, A3 (Cont.)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | 1234567 DESCRIPTION | UNITS PER ASSY | USABLE <br> ON CODE | $\begin{array}{\|c} \text { SMR } \\ \text { CODE } \end{array}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-4. |  |  |  |  |  |  |
| -21 | D155E220DO | 81349 | -. CAPACITOR,C31, C32, Mica, 22 pf, 500 WVDC, $\pm 2 \%$ | 2 |  | PADZZN |
| -22 | MS39014101-1593 | 81349 | -. CAPACITOR, C24, thru C27, Ceramic, 0.1 uf, 50 WVDC, $\pm 20 \%$ | 4 |  | PADZZN |
| -23 | LT100K133 | 81349 | -• CHOKE, LI, L3, L9, RF $15 \mathrm{uH}, \pm 10 \%$ | 3 |  | PADZZN |
| -24 | 08530 | 23386 | - TRANSFORMER ASSEMBLY, ............ | 1 |  | PADZZN |
| -25 | RLR07C390GR | 81349 | -•RESISTOR, R34, Fixed Film, 39 ohms, $\pm 2 \%$, V4 watt | 1 |  | PADZZN |
| -26 | RLR07C100GR | 81349 | -•RESISTOR, R16, R37, Fixed Film, 10 ohms, $\pm 2 \%, 1 / 4$ watt | 2 |  | PADZZN |
| -27 | RL07S270G | 81349 | -•RESISTOR, R14, Fixed Film, 27 ohms, $\pm 2 \%$, Y4 watt | 1 |  | PADZZN |
| -28 | RL07S222G | 81349 | -. RESISTOR, R8, R12, R28 Fixed Film, 2.2K ohms, $\pm 2 \%$, V4 watt | 3 |  | PADZZN |
| -29 | CA324E | 02735 | .. INTEGRATED CIRCUIT, $\qquad$ U2, 324 Quad Operational Amplifier | 1 |  | PADZZN |
| -30 | T362B156M020AS | 31433 | .. CAPACITOR,C13,C28,C29, Tantalum, 15 uf, 20 WVDC, 20\% | 3 |  | PADZZN |
| -31 | LTIOK012 | 81349 |  | 1 |  | PADZZN |
| -32 | RL07S152G | 81349 | -. RESISTOR, R43, Fixed Film, 1.5 K ohms, $\pm 2 \%, 1 / 4$ watt | 1 |  | PADZZN |
| -33 | MC1733CP | 04713 | INTEGRATED CIRCUIT <br> U4, Operational Amplifier | 1 |  | PADZZN |
| -34 | MS18034-4 | 96906 |  | 1 |  | PAOZZN |
| -35 | 08479-3 | 23386 | . $\quad . \quad$ CABLE, W2, 455 kHz Output ........... | 2 |  | MOOZZN |
| -36 | RL07S331G | 81349 | -. RESISTOR, R41, R42, Fixed Film, 330 ohms, $\pm 2 \%$, ' 4 watt | 2 |  | PADZZN |
| -37 | C320C103M1UIC1 | 31433 | -- CAPACITOR, C4, Ceramic, 0.01 uf, 50 WVDC, $\pm 20 \%$ | 1 |  | PADZZN |
| -38 | RLR07C224GR | 81349 | .. RESISTOR, R4, Fixed Film, 220 K ohms, $\pm 2 \%$, V4 watt | 1 |  | PADZZN |
| -39 | 08499-5 | 23386 | - $\cdot$ CONNECTOR, JI ........................... | 1 |  | PAOZZN |
| -40 | RL07S393G | 81349 | -. RESISTOR, R22, Fixed Film, 39 K ohms, $\pm 2 \%$, ' 4 watt | 1 |  | PADZZN |
| -41 -42 | JAN2N918 7717-HWHT | 81349 13103 | . $\quad . \quad$ TRANSISTOR, Q3, NPN .......................... | 1 |  | PADZZN PADZZN |
| -43 | RJSOFW501 | 81349 | -. RESISTOR, R26, Variable $\qquad$ <br> 500 ohms, $\pm 20 \%$ | 1 |  | PADZZN |
| -44 | RL07S560G | 81349 | -• RESISTOR, R25, Fixed Film, ............ 56 ohms, $\pm 2 \%$, /4 watt | 1 |  | PADZZN |
| -45 | 08503 | 23386 | .. TRANSFORMER, T1, RF ................. Variable | 1 |  | PADZZN |
| -46 | CMRO5E330GODR | 81349 | -. CAPACITOR, C9, Mica, 33 pf, 550 WVDC, $\pm 2 \%$ | 1 |  | PADZZN |
| -47 | U310 | 17896 | -. TRANSISTOR, QI, Q2, Field Effect | 2 |  | PADZZN |

Change $1 \quad 7-38$
T.O. 31R2-2URR-251

FIGURE 7-4. SECOND MIXER ASSEMBLY, CIRCUIT CARD, A3 (Cont.)



Figure 7-5. Main IF/AF Assembly, Circuit Card, A4 (Sheet 1 of 2) 7-40

FIGURE 7-5. MAIN IF/AF ASSEMBLY, CIRCUIT CARD, A4

| FIGURE \& INDEX No. | PART NUMBER | FSCM | 1234567 DESCRIPTION | $\begin{aligned} & \text { UNITS } \\ & \text { PER } \\ & \text { ASSS } \end{aligned}$ | USABLE ON CODE | $\begin{aligned} & \text { SMR } \\ & \text { CODE } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7.5./ | 08465 | 23386 | - MAIN IF/AF ASSEMBLY, $\qquad$ Circuit Card, A4 | 1 |  | PAOLDT |
| /1,2-1 | RL07S472G | 81349 | -. RESISTOR, R1 thru R7, $\qquad$ RS0, R92, R126, R136, R137, Fixed Film, 4.7K ohms, $\pm 2 \%, 1 / 4$ watt | 12 |  | PADZZN |
| /1,2-2 | Not Used |  | -. CAPACITOR, C1 thruC12, C48, C55, C90, C109, C110 | 17 |  | PADZZN |
| /1, 2-3 | 051-351-0000-220 | 98291 | .. CONNECTOR, J1, J3 thru J6, $\qquad$ RF, 700209 | 5 |  | PAOZzN |
| /1-4 | NS-441-B1 | 06776 | -. SOCKET, Filter Pin, 0,040 $\qquad$ <br> diameter | 14 |  | XA |
| /1-5 | NS432-100 | 06776 | -• SOCKET, Filter Pin, 0.100 $\qquad$ diameter | 14 |  | XA |
| /1,2-6 | MS39014101-1593 | 81349 | -. CAPACITOR,C13 throughCl9, C21 thru C31, C33, C34, C36, C38 thru C40, C45, C49, C50, C54, C56, C57, C60 thru C67, C72, C75 thru C79, C81, C82, C94, C100, C101,C103,C104,C111,, C112, C120, C121, ceramic, 0.1 uf, 50 WVDC, $\pm 20 \%$ | 55 |  | PADZZN |
| /1,2-7 | RL07S103G | 81349 | .. RESISTOR, R16 thru R22, <br> R24, R33, R36, R40, R53, R57, R58, R71, R86, R88, R89, R130, R132, R135, Fixed Film, 10K ohms, $\pm 2 \%$, $1 / 4$ watt | 21 |  | PADZZN |
| /1,2-8 | IN916B | 07623 | .. DIODE, CR1 thru CR16, CR18.......... thru CR31 | 30 |  | PADZZN |
| /1,2-9 | RL07S102G | 81349 | .. RESISTOR, R9 thru R15, R23........... <br> R28, R55, R62, R66, R68, <br> R70, R72, R77 thru R79, <br> R85, R87, R102, R144, <br> R145, Fixed Film, IK ohms, $\pm 2 \%, 1 / 4$ watt | 23 |  | PADZZN |
| /1, 2-10 | 2N5089 | 04713 | .. TRANSISTOR, Q1, Q2, Q6 $\qquad$ thru Q10, NPN, High Power | 7 |  | PADZZN |
| /1, 2-11 | RL07SIOIG | 81349 | .. RESISTOR, R25, R60 $\qquad$ R141, Fixed Film, 100 ohms, $\pm 2 \%, 1 / 4$ watt | 3 |  | PADZZN |
| /1,2-12 | RL07S222G | 81349 | -. RESISTOR, R34, R51, R101, $\qquad$ <br> R128, Fixed Film, 2.2K ohms, $\pm 2 \%$, $1 / 4$ watt | 4 |  | PADZZN |
| /1,2-13 | RL07S104G | 81349 | .. RESISTOR, R37, R95 $\qquad$ <br> R96, R117, R122, R123, <br> R125, R129, R140, R150, <br> R151, Fixed Film, 100 K ohms,$\pm 2 \%$, $1 / 4$ watt. | 11 |  | PADZZN |

FIGURE 7-5. MAIN IF/AF ASSEMBLY, CIRCUIT CARD, A4 (Cont.)


FIGURE 7-5. MAIN IF/AF ASSEMBLY, CIRCUIT CARD, A4 (Cont.)



Figure 7-5. Main F/AF Assembly, Circuit Card, A4 (Sheet 2 of 2)

FIGURE 7-5. MAIN IF/AF ASSEMBLY CIRCUIT CARD, A4 (Cont.)

|  <br> INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ |  | USABLE <br> ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-5- |  |  |  |  |  |  |
| /2-55 | Not Used |  | -• DIODE, CR17, Not Used .................. | 1 |  | PADZZN |
| /2-56 | M38510/05101BCA | 81349 | .. INTEGRATEDCIRCUIT $\qquad$ <br> U9, Flip Flop | 1 |  | PADZZN |
| /2-57 | CD40109BF3 | 54590 | .. INTEGRATED CIRCUIT $\qquad$ U3, US, U16, Level Translator | 3 |  | PADZZN I |
| /2-58 | 08531 | 23386 | . . COIL ASSEMBLY, L3, RF................. | 1 |  | PADZZN |
| /2-59 | CMR06F332GODR | 81349 | - CAPACITOR, C68, Mica, $3300 \mathrm{pf}, 500$ WVDC, $\pm 2 \%$ | 1 |  | PADZZN |
| /2-60 | CD4053BD/3 | 02735 | .. INTEGRATEDCIRCUIT $\qquad$ U 19, Multiplexer/Demultiplexer | 1 |  | PADZZN |
| /2-61 | C320C152K2R5C1 | 31433 | - CAPACITOR, C88, Ceramic, $\qquad$ 1500 pf, 100 WVDC, $\pm 10 \%$ | 1 |  | PADZZN |
| /2-62 | RL07S822G | 81349 | -. RESISTOR, R127, $\qquad$ <br> Fixed Film. 8.2K ohms, $\pm 2 \%$. $1 / 4$ watt | 1 |  | PADZZN |
| /2-63 | DM I SF821F0300- | 72136 | -. CAPACITOR, C89, Mica $\qquad$ WV4CR 820 pf, 300 WVDC, $\pm 2 \%$ | 1 |  | PADZZN |
| /2-64 | CA1458E | 02735 | .- INTEGRATEDCIRCUIT, $\qquad$ U22, U28, Dual Operational Amplifier | 2 |  | PADZZN |
| /2-65 | RJ50FW502 | 81349 | .. RESISTOR, R119, Variable,.............. 5K ohms | 1 |  | PADZZN |
| /2-66 | RL07S683G | 81349 | -. RESISTOR, R120, Fixed $\qquad$ <br> Film, 68K ohms, $\pm 2 \%$, 1/4 watt | 1 |  | PADZZN |
| /2-67 | RL07S393G | 81349 | -. RESISTOR, R118, Fied $\qquad$ <br> Film, 39K ohms, $\pm 2 \%$, $1 / 4$ watt | 1 |  | PADZZN |
| /2-68 | DTZ-10 | 71590 | -. CAPACITOR,C84,Ceramic $\qquad$ $10 \mathrm{pf}, \pm 5 \%$ | 1 |  | PADZZN |
| /2-69 | SN17524N | 01295 | .. INTEGRATED CIRCUIT, $\qquad$ <br> U21, 8 Bit Buffer <br> Multiplying | 1 |  | PADZZN |
| /2-70 | T362AI05M035AS | 31433 | -. CAPACITOR, C86, C87 $\qquad$ <br> C114, Cl 116, Tantalum, <br> 1.0 uf, 35 WVDC, <br> $\pm 20 \%$ | 4 |  | PADZZN |
| /2-71 | M38510/11201BCB | 02735 | .- INTEGRATED CIRCUIT $\qquad$ U24, Quad Voltage Comparator | 1 |  | PADZZN |
| /2-72 | 3432-2003 | 75037 | -• CONNECTOR, J2, ............................. <br> Ribbon, 40 way | 1 |  | PADZZN |
| /2-73 | CMR05FIOIGODR | 81349 | -. CAPACITOR,C71,Mica, 100 pf, 500 WVDC, $\pm 2 \%$ | 1 |  | PADZZN |

FIGURE 7-5. MAIN IF/AF ASSEMBLY CIRCUIT CARD, A4 (Cont.)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ |  | USABLE <br> ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & 7-5- \\ & / 2-74 \end{aligned}$ | MC1357P | 04713 | .- INTEGRATEDCIRCUIT, ................... U18, FM Detector | 1 |  | PADZZN |
|  |  |  |  |  |  |  |
| /2-75 | MC1496L | 04713 | .. INTEGRATED CIRCUIT <br> U20, Product Detector | 1 |  | PADZZN |
| /2-76 | CD22100F | 02735 |  | 1 |  | PADZZN |
|  |  |  | -. INTEGRATED CIRCUIT U25, $4 \times 4$ Crosspoint Switch |  |  |  |
| /2-77 | ULN2278B | 56289 | .. INTEGRATED CIRCUIT $\qquad$ U26, Audio Amplifier, 2 watt | 1 |  | PADZZN |
|  |  |  |  |  |  |  |
| /2-78 | T362C686M015AS | 31433 | - CAPACITOR, C107, C118, C119, Tantalum, 68 uf, 15 WVDC, $\pm 20 \%$ | 3 |  | PADZZN |
|  |  |  |  |  |  |  |
| /2-79 | LT10K012 | 81349 | -• CHOKE, L4, L5, 100 uH..................... $\pm 5 \%$ | 2 |  | PADZZN |
|  |  |  |  |  |  |  |
| /2-80 | 3431-2003 | 75037 | - CONNECTOR, J8. <br> Ribbon, 34 way <br> -• TRANSFORMER ASSEMBLY, T1 | 1 |  | PADZZN |
| /2-81 | 08535 | 23386 |  | 1 |  | PADZZN |
| /2-82 | MS3564944 <br> MS35338-78 | $\begin{aligned} & 81349 \\ & 81349 \end{aligned}$ | -• NUT, Hex, No. 4 (AP) $\qquad$ <br> -• WASHER, Split Lock, $\qquad$ <br> No. 4 (AP) | 22 |  | $\begin{aligned} & \text { PAOZZN } \\ & \text { PAOZZN } \end{aligned}$ |
|  |  |  |  |  |  |  |
|  | MS15795-303 MS35233-14 | $\begin{aligned} & 96906 \\ & 96906 \end{aligned}$ | - WASHER, Flat, No. 4, (AP) <br> -. SCREW, Machine, Pan Head, $4-40 \times 5 / 16(\mathrm{AP})$ |  |  | $\begin{array}{\|l} \text { PAOZZN } \\ \text { PAOZZN } \end{array}$ |
|  |  |  |  | $\begin{aligned} & 2 \\ & 2 \end{aligned}$ |  |  |
| /2-83 | 3429-2003 | 75037 | -• CONNECTOR, J7. <br> Ribbon, 26 way | 1 |  | PADZZN |
| /2-84 | RN55D1400F | 81349 | -. RESISTOR, R142, Fixed $\qquad$ <br> Film, 140 ohms, $\pm 2 \%$, 14 watt |  |  | PADZZN |
| /2-85 | DD-391 | 71590 | - CAPACITOR, C11S, C1 17. Ceramic, 390 pf, 500 WVDC, $\pm 10 \%$ <br> INTEGRATED CIRCUIT | 2 |  | PADZZN |
| /2-86 | MC3340L | 04713 | INTEGRATED CIRCUIT $\qquad$ U30, U31, Electronic Attenuator | 2 |  | PADZZN |
| /2-87 | 3071FE221T016SF | 30983 | -. CAPACITOR, C99, C105, $\qquad$ C108, Electrolytic, 220 uf, 16 WVDC, $-10+50 \%$. <br> -• INTEGRATED CIRCUIT $\qquad$ <br> U27, Voltage Regulator, $\pm 15$ volts, 7812 <br> -. NUT, Hex, No. 4 (AP) $\qquad$ <br> - WASHER, Split Lock. $\qquad$ $\text { No. } 4 \text { (AP) }$ <br> -. WASHER, Flat, No. 4 (AP)1 $\qquad$ <br> -. SCREW, Machine, Pan Head, $\qquad$ $4-40 \times 5 / 16 \text { (AP) }$ | 3 |  | PADZZN |
| /2-88 | M38510/10708BEA. | 81349 |  |  |  | PADZZN |
| /2-89 | MS3564944 MS35338-78 <br> MS1 5795-303 MS35233-14 | $\begin{aligned} & 81349 \\ & 81349 \\ & \\ & 96906 \\ & 96906 \end{aligned}$ |  |  |  | $\begin{aligned} & \text { PAOZZN } \\ & \text { PAOZZN } \end{aligned}$ |
|  |  |  |  | 1 |  | PAOZZN |
|  |  |  |  | $1$ |  | $\begin{aligned} & \text { PAOZZN } \\ & \text { PAO77N } \end{aligned}$ |
|  |  |  |  | 1 |  | PAOZZN |

Change 1 7-47


Figure 7-6. ISB Assembly, Circuit Card, A5

FIGURE 7-6. ISB ASSEMBLY, CIRCUIT CARD, A5

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | 1234567 DESCRIPTION | UNITS PER ASSY | USABLE <br> ON CODE | $\begin{aligned} & \text { SMR } \\ & \text { CODE } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-6 | 08390 | 23386 | - ISB ASSEMBLY, Circuit Card, A5. | 1 |  | PAOLDT |
| -1 | Not Used |  | .. TRANSISTOR, Q2, Q4 Q5, Q9, Not Used | 4 |  |  |
| -2 | Not Used |  | .. RESISTOR, R31, R33, R36, R40, R42, R44, R46, R47, R50, R62, R80, Not Used | 11 |  |  |
| -3 | Not Used |  | - CONNECTOR, J2, Not Used ........... | 1 |  |  |
| -4 | Not Used |  | .. CAPACITOR, C20, C22, C27, C29, Not Used | 4 |  |  |
| -5 | CMR06F152GODR | 81349 | -. CAPACITOR,C16,C19, Mica, $1500 \mathrm{pf}, 500$ WVDC, $\pm 2 \%$ | 2 |  | PADZZN |
| -6 | M39014101-1593 | 81349 | .- CAPACITOR, C1,C3, C5, C7, C8, C10, C12 thru C15, C21, C23, C28, C30, C31,C33 thru C37, C45, C54, Ceramic, 0.1 uf, 50 WVDC, $\pm 20 \%$ | 22 |  | PADZZN |
| -7 | 08485 | 23386 |  | 2 |  | PADZZN |
| -8 | CMROS5E820GOD中 | 81349 | .. CAPACITOR,C18,Mica, 82 pf, 500 WVDC, $\pm 2 \%$ | 1 |  | PADZZN |
| -9 | RL07S102G | 81349 | .. RESISTOR, R12, R30, R39,.............. <br> R48, R51, R53, R59, Fixed Film, 1 K ohms, $\pm 2 \%, 1 / 4$ watt | 7 |  | PADZZN |
| -10 | RLO7S 103G | 81349 | -• RESISTOR, R10, R11, <br> R14, R29, R34, R35, <br> R43, R56, R63, Fixed <br> Film, 10K ohms, $\pm 2 \%$, <br> $1 / 4$ watt | 9 |  | PADZZN |
| -11 | RL07S471G | 81349 | .. RESISTOR, R8, R64, Fixed $\qquad$ Film, 470 ohms, $\pm 2 \%$, $1 / 4$ watt | 2 |  | PADZZN |
| -12 | 2N5089 | 04713 | .. TRANSISTOR, Q1, Q3, $\qquad$ Q6 thru Q8, NPN, High Power | 5 |  | PADZZN |
| -13 | T362B156K020AS | 31433 | -. CAPACITOR, C9, C26 C53, Tantalum, 15 uf, 20 WVDC, $\pm 20 \%$ | 3 |  | PADZZN |
| -14 | RL07S470G | 81349 | -. RESISTOR, R16, R22 Fixed Film, 47 ohms, $\pm 2 \%, 1 / 4$ watt | 2 |  | PADZZN |
| -15 -16 | RL07S472G RL07S332G | 81349 81349 | -• RESISTOR, R27, R72, R73, R78, Fixed Film, 4.7 K ohms, $\pm 2 \%, 1 / 4$ watt | 5 |  | PADZZN PADZZN |
| -16 | RL07S332G | 81349 | -• RESISTOR, R6, R65 thru R67, R69, Fixed Film, 3.3 K ohms, $\pm 2 \%$, $1 / 4$ watt | 5 |  | PADZZN |

FIGURE 7-6. ISB ASSEMBLY, CIRCUIT CARD, A5 (Cont.)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ | UNITS PER ASSY | USABLE <br> ON CODE | $\begin{aligned} & \text { SMR } \\ & \text { CODE } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & 7-6- \\ & -17 \end{aligned}$ | RL07S223G | 81349 | -• RESISTOR, RI, R5, R26; $\qquad$ <br> R32, R37, R41, R54, R58, <br> Fixed Film, 22K ohms, <br> $\pm 2 \%, 1 / 4$ watt | 8 |  | PADZZN |
| -18 | RL07S222G | 81349 | -• RESISTOR, R7, R25, R81, <br> Fixed Film, 2.2K ohms, $\pm 2 \%, 1 / 4$ watt | 3 |  | PADZZN |
| -19 | RL07S473G | 81349 | -• RESISTOR,R9, R13, R15 $\qquad$ R28, Fixed Film, 47K ohms, $\pm 2 \%$, 4 watt | 4 |  | PADZZN |
| -20 | RJ50FW202 | 81349 | -. RESISTOR, R19 $\qquad$ <br> Variable, 2 K ohms | 1 |  | PADZZN |
| -21 | UA757DC | 07263 | -• INTEGRATED CIRCUIT, $\qquad$ U6, Operational Amplifier | 1 |  | PADZZN |
| -22 | RL07S681G | 81349 | -• RESISTOR, R17, R24, $\qquad$ Fixed Film, 680 ohms, $\pm 2 \%$, /4 watt | 2 |  | PADZZN |
| -23 | C320C103MIUICI | 31433 | -• CAPACITOR. C17, C24, C25, ........... Ceramic, 0.01 uf, 50 WVDC, $\pm 20 \%$ | 3 |  | PADZZN |
| -24 | ETIO1X025A5 | 30983 | -. CAPACITOR, C4, C50,..................... <br> Electrolytic, 100 uf, <br> 25 WVDC, $-10 \pm 50 \%$ | 2 |  | PADZZN |
| -25 | 051-351-0000-220 | 98291 | -• CONNECTOR, J, J3,RF ................. | 2 |  | PADZZN |
| -26 | RL07S10IG | 81349 | -. RESISTOR, R21, Fixed <br> Film, 100 ohms, $\pm 2 \%$, 1/ watt | 1 |  | PADZZN |
| -27 | RL07S563G | 81349 | -• RESISTOR, R3, Fixed $\qquad$ <br> Film, 56 K ohms, $\pm 2 \%$, /4 watt | 1 |  | PADZZN |
| -28 | RL07S123G | 81349 | -. RESISTOR, R4, R52, Fixed $\qquad$ Film, 12 K ohms, $\pm 2 \%$, /4 watt | 2 |  | PADZZN |
| -29 | RJ50FW503 | 81349 | - R RESISTOR, R23,Variable $\qquad$ 50K ohms | 1 |  | PADZZN |
| -30 | LM78LIZAWC | 27014 | .. INTEGRATED CIRCUIT $\qquad$ <br> U11, $\pm 12$ volt regulator | 1 |  | PADZZN |
| -31 | ET471X025A01 | 30983 | - CAPACITOR, C52 $\qquad$ <br> Electrolytic, 470 uf, <br> 25 WVDC,-10+100\% | 11 |  | PADZZN |
| -32 | RL07S391G | 81349 81349 | -• RESISTOR, R61, Fixed $\qquad$ <br> Film, 390 ohms, $\pm 2 \%$, <br> $1 / 4$ watt | 11 |  | PADZZN |
| -33 | RL07S220G | 81349 | -. RESISTOR, R83, Fixed <br> Film, 22 ohms, $\pm 2 \%$, <br> $1 / 4$ watt | 1 |  | PADZZN |

FIGURE 7-6. ISB ASSEMBLY, CIRCUIT CARD, A5 (Cont.)


FIGURE 7-6. ISB ASSEMBLY, CIRCUIT CARD, A5 (Cont.)

| FIGURE \& INDEX No. | PART NUMBER | FSCM | 12334567 DESCRIPTION | $\begin{aligned} & \text { UNITS } \\ & \text { PER } \\ & \text { ASSY } \end{aligned}$ | USABLE <br> ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-6- |  |  |  |  |  |  |
| -51 | M38510/05101BEA | 81349 | -• INTEGRATED CIRCUIT $\qquad$ <br> U4, $\pm 12$ volt regulator <br> .. INTEGRATED CIRCUIT, $\qquad$ <br> U5, U9, Quad Bilateral <br> Switch <br> - RESISTOR, R85, Fixed $\qquad$ <br> Film, IO ohms, $\pm 2 \%$, <br> $1 / 4$ watt <br> -• CABLE, WI, Ribbon $\qquad$ | 1 |  |  |
| -52 | CD4066BEX | 02735 |  | 2 |  | PADZZN |
|  |  |  |  |  |  |  |
| -53 | RLR07CIO0GR | 81349 |  | 1 |  | PADZZN |
|  |  |  |  |  |  |  |
| -54 | 08486 | 23386 |  |  |  | $\begin{array}{\|l\|} \hline \text { XA } \\ \text { PADZZN } \end{array}$ |
| -55 | RL07S] 53G | 81349 | -. RESISTOR, R45, R70 <br> R71, Fixed Film, 15K ohms, $\pm 2 \%$, V\% watt | 3 |  |  |
|  |  |  |  |  |  |  |
| -56 | RN5SD1400F | 81349 | .- RESISTOR, R82, R84 $\qquad$ <br> Fixed Film, 140 ohms, <br> $\pm 2 \%, 1 / 4$ watt | 2 |  | PADZZN |
|  |  |  |  |  |  |  |
| -57 | RL07S104G | 81349 | .. RESISTOR, R76, R77 <br> Fixed Film, 100K ohms, $\pm 2 \%$, /4 watt <br> -. CAPACITOR, C48, C49 | 2 |  | PADZZN |
|  |  |  |  |  |  |  |
| -58 | T362C686M015AS | 31433 |  | 2 |  | PADZZN |
|  |  |  | -. CAPACITOR, C48, C49 Tantalum, 68 uf, 15 WVDC, $\pm 20 \%$ |  |  |  |
| -59 | LM1877N-9 | 27014 | -. INTEGRATED CIRCUIT U12, Audio Amplifier, 2 watt, dual <br> -. CAPACITOR, C40, C47 | 1 |  | PADZZN |
|  |  |  |  |  |  |  |
| -60 | 3071FE221T0168F | 30983 |  | 2 |  | PADZZN |
|  |  |  | -. CAPACITOR, C40, C47,................... Electrolytic, 220 uf, 16 WVDC, $-10 \pm 50 \%$ |  |  |  |
| -61R | RLO7S682G | 81349 | -. RESISTOR, R57, Fixed <br> Film, 6.8 K ohms, <br> $\pm 2 \%, 1 / 4$ watt | 1 |  | PADZZN |
|  |  |  |  |  |  |  |
| -62 | RLR07C105GR | 81349 |  |  |  | PADZZN |
|  |  |  | -. RESISTOR, R86, Fixed $\qquad$ <br> Film, 1 meg ohms, <br> $\pm 2 \%, 1 / 4$ watt |  |  |  |
| -63 | CD4013BE | 02735 | .. INTEGRATED CIRCUIT .................. | 2 |  | PADZZN |
|  |  |  | U7, U8, Quad Latch |  |  |  |
| -64 | RL07S183G | 81349 | -• RESISTOR, R49, Fixed $\qquad$ <br> Film, 18 K ohms, $\pm 2 \%$, | 1 |  | PADZZN |
|  |  |  | 1/4 watt |  |  |  |
| -65 | 08535 | 23386 | .. TRANSFORMER ASSEMBLY, .......... T1, T2, Audio | 2 |  | PADZZN |
| -66 | MS35649-44 | 81349 | . . NUT, Hex, 4-40 (AP) ....................... | 2 |  | PAOZZN |
|  | MS35338-78 | 81349 | .. WASHER, Split Lock $\qquad$ <br> No. 4 (AP) | 2 |  |  |
|  | MS15795-303 | $96906$ | -. WASHER, Flat, No 4 (AP) |  |  | $\begin{aligned} & \text { PAOZZN } \\ & \text { PAOZZN } \end{aligned}$ |
|  | MS35233-14 | $96906$ | - SCREW, Machine, Pan Head, $\qquad$ $440 \times 5 / 16 \text { (AP) }$ | $2$ |  |  |
| -67 | M38510/10708BEA | 81349 | .. INTEGRATED CIRCUIT <br> U13, $\pm 15$ volt regulator | 1 |  | PADZZN |

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FIGURE 7-6. ISB ASSEMBLY, CIRCUIT CARD, AS (Cont.)



Figure 7-7. Serial Asynchronous Interface Assembly, Circuit Card, A6AI

FIGURE 7-7. SERIAL ASYNCHRONOUS INTERFACE ASSEMBLY,CIRCUIT CARD, A6A1

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ | UNITS PER ASSY | USABLE <br> ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-7- | 08391 | 23386 | -• ASYNCHRONOUS $\qquad$ INTERFACE ASSEMBLY, Circuit Card, A6A1 | 1 |  | PAFLDT |
| -1 | MS90376-16Y | 96906 | -• COVER, Dust $\qquad$ Connector WIJI | 1 |  | PAOZZN |
| -2 | M83723-02R-1626N | 81349 | -••CONNECTOR, WIJ1 $\qquad$ 26 Pin | 1 |  | PAFZZN |
| -3 | MS3564944 | 81349 | - . . NUT, Hex, 4-40 (AP) ...................... | 5 |  | PAOZZN |
|  | MS35338-78 | 81349 | -. . WASHER, Split Lock,' $\qquad$ No. 4 (AP) | 5 |  | PAOZZN |
|  | MS15795-303 | 96906 | ... WASHER, Flat $\qquad$ <br> No. 4 (AP) | 5 |  | PAOZZN |
|  | MS35233-14 | 96906 | ... SCREW, Machine, $\qquad$ Pan Head, $440 \times 5 / 16$ (AP) | 5 |  | PAOZZN |
| -4 | 08493 | 23386 | -••CABLE ASSEMBLY, $\qquad$ W1 | 1 |  | XA |
| -5 | 08471 | 23386 | -••PLATE, Connector $\qquad$ mounting, J1 | 1 |  | XA |
| -6 | 22M40 | 13257 | -• NUT, Hex, Elastic Stop, $\qquad$ 4-40 (AP) | 2 |  | PAOZZN |
|  | MS35233-15 | 96906 | -••SCREW, Machine, $\qquad$ Pan Head, $440 \times 3 / 8$ (AP) | 2 |  | PAOZZN |
| -7 | SIP-8-223 | 91637 | -•• INTEGRATED CIRCUIT, U1, U10, Resistor Network, 22K, 8 pin | 2 |  | PADZZN |
| -8 | C320C103M1U1C1 | 31433 | - . CAPACITOR, C4, $\qquad$ <br> C7, C8, C9, C11 thru <br> C18, Ceramic, 0.01 uf, <br> 50 WVDC, $\pm 20 \%$ | 12 |  | PADZZN |
| -9 | AM26LS32CN | 01295 | -•• INTEGRATED CIRCUIT $\qquad$ <br> U2, Quad differential line receiver | 1 |  | PADZZN |
| -10 | AM26LS30PC | 34335 | -•• INTEGRATED CIRCUIT, $\qquad$ <br> U3, Quad line driver | 1 |  | PADZZN |
| -11 | CMR05F301GODR | 81349 | -•• CAPACITOR, C10, C19 $\qquad$ Mica, 300 pf, 500 WVDC, $\pm 5 \%$ | 2 |  | PADZZN |
| -12 | MIL-W-3861/5-22 | 81349 | - . - WIRE, Links, LKI1, $\qquad$ LK3, LK4, LK5, Buss No. 22 | AR |  | PAOZZN |
| -13 | MIL-I- 22129C-22 | 81349 | ... SLEEVING, Links, LK1, LK3, LK4, LK5, PTFE tubing No. 22 | AR |  | PAOZZN |
| -14 | 08490 | 23386 | •• CRYSTAL, Y, ................................... 4.9152 MHz | 1 |  | PAOZZN |
| -15 | AH5016C-5 | 27014 | -•• INTEGRATED CIRCUIT, $\qquad$ <br> U4, Baud Rate Generator | 1 |  | PADZZN |

FIGURE 7-7. SERIAL ASYNCHRONOUS INTERFACE ASSEMBLY,CIRCUIT CARD, A6A1 (Cont.)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ | UNITS PER ASSY | USABLE <br> ON CODE | $\begin{aligned} & \text { SMR } \\ & \text { CODE } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-7- |  |  |  |  |  |  |
| -16 | COM6402P | 53848 | -•• INTEGRATED CIRCUIT, .................. U9, UART | 1 |  | PADZZN |
| -17 | CD4503BCN | 27014 | -•• INTEGRATED CIRCUIT, $\qquad$ <br> U 11 thru U14, Hex three state buffer | 4 |  | PADZZN |
| -18 | LM7905CN | 27014 | -•• INTEGRATED CIRCUIT, $\qquad$ <br> U5, -5 volts regulator | 1 |  | PADZZN |
| -19 | T362A685M035AS | 31433 | -•• CAPACITOR,C5,C6, <br> Tantalum, 6.8 uf, 35 WVDC, $\pm 20 \%$ | 2 |  | PADZZN |
| -20 | M38510/05001BCB | 81349 | -•• INTEGRATED CIRCUIT, $\qquad$ <br> U6, UI5, Quad 2 input <br> NAND | 2 |  | PADZZN |
| -21 | M38510/30107BEB | 81349 | -•• INTEGRATEDCIRCUIT, U16, Quad D Flip Flop | 1 |  | PADZZN |
| -22 | M38510/3003BCB | 81349 | -•• INTEGRATEDCIRCUIT, <br> U7, Hex inverter | 1 |  | PADZZN |
| -23 | CD4085BF | 02735 | -••INTEGRATED CIRCUIT, $\qquad$ <br> U8, Dual AND/OR <br> inverter gate | 1 |  | PADZZN |
| -24 | M38510/06203BEA | 81349 | -•• INTEGRATED CIRCUIT, $\qquad$ U17, Dual 4 input NOR gate | 1 |  | PADZZN |
| -25 | 4310R-101-223 | 32997 | . INTEGRATED CIRCUIT, <br> U18, Resistor network, 22K, 10 pin | 1 |  | PADZZN |
| -26 | CD4529BCN | 27014 | -.. INTEGRATED CIRCUIT, $\qquad$ U19, Dual 4 channel data select | 1 |  | PADZZN |
| -27 | T362B226K015AS | 31433 | -••CAPACITOR, C3 $\qquad$ <br> Tantalum, 22 uf, <br> 15 WVDC, $\pm 20 \%$ | 1 |  | PADZZN |
| -28 -29 | MS39014101-1593 | 81349 | -••CAPACITOR, CI $\qquad$ Ceramic, 0.1 uf, 50 WVDC, $\pm 20 \%$ | 1 |  | PADZZN |
| -29 -30 | MS39014/2-1356 08491 | $81349$ | -••CAPACITOR, C2 <br> Ceramic, 022 uf, <br> 50 WVDC, $\pm 20 \%$ | 1 |  | PADZZN |
| -30 | 08491 | 23386 | 50 pin, 2 section | 1 |  | PADZZN |



Figure 7-8. Microcomputer Assembly, Circuit Card, A6A2
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FIGURE 7-8. MICROCOMPUTER ASSEMBLY, CIRCUIT CARD, A6A2

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ |  | USABLE <br> ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-8- | 4100011-501 | 23386 | -• MICROCOMPUTER $\qquad$ ASSEMBLY, Circuit Card, A6A2 (with EPROM ROM set) | 1 |  | PAOLDT |
| 7-1/1-40 | 08392 | 23386 | -. MICROCOMPUTER $\qquad$ <br> ASSEMBLY, Circuit Card A6A2 (without EPROM, ROM set) | 1 |  |  |
| -1 | 65000-036 | 22526 | -• CONNECTOR, J2 $\qquad$ <br> 34 pin | 1 |  | PADZZN |
| -2 | MS39014101-1593 | 81349 | -.. CAPACITOR, C3 thru $\qquad$ <br> C10, C18, Ceramic, <br> 0.1 uf, 50 WVDC, <br> $\pm 20 \%$ | 9 |  | PADZZN |
| -3 | MK3850 | 50088 | -•• INTEGRATED CIRCUIT, $\qquad$ <br> U1, Central Processor Unit | 1 |  | PADZZN |
| -4 | MK3853 | 50088 | - .• INTEGRATED CIRCUIT, $\qquad$ U2, System Memory Interface | 1 |  | PADZZN |
| -5 | M38510/30005BCB | 81349 | -•• INTEGRATED CIRCUIT, $\qquad$ U10, Triple 3 input NAND | 1 |  | PADZZN |
| -6 | DTZ-15 | 71590 | -•• CAPACITOR, C1,C2,........................ <br> Ceramic, 15 pf, Non <br> Polarized, $\pm 5 \%$ | 2 |  | PADZZN |
| -7 | 08487 | 23386 | - . - CRYSTAL, Y1 <br> 2 MHz | 1 |  | PADZZN |
| -8 | M38510/30702BEB | 49956 | -.. INTEGRATEDCIRCUIT, $\qquad$ <br> U3, Decoder, <br> Demultiplexer | 1 |  | PADZZN |
| -9 | M38510/31004BEA | 81349 | -.. INTEGRATED CIRCUIT, $\qquad$ <br> U4, octal, 3 state transceiver | 1 |  | PADZZN |
| -10 | 08449 | 23386 | -•• INTEGRATED CIRCUIT, $\qquad$ U5, U6, U14, Firmware, EPROM, ROM, 1 set of 3 | 1 |  | PADZZN |
| -11 | A23-2023Z | K1935 | -••SOCKET,XU5,XU6 $\qquad$ XU14, integrated circuit | 3 |  | PADZZN |
| -12 | M38510/30301BCB | 81349 | -•• INTEGRATED CIRCUIT, ................... <br> U11, Quad 2 input NOR | 1 |  | PADZZN |
| -13 | M38510/3003BCB | 81349 | -••INTEGRATED CIRCUIT, U12, Hex inverter | 1 |  | PADZZN |
| -14 | SN74S374N | 01295 | -•• INTEGRATED CIRCUIT, $\qquad$ <br> U13, octal tri state <br> D flip flop | 1 |  | PADZZN |
| -15 | DS2SD | 24446 | -••BATTERY, BT1, Nickel Cadmium, 2.4 VDC | 1 |  | PADZZN |
| -16 | PS101L | 34649 | -.. INTEGRATED CIRCUIT $\qquad$ <br> U7, U8, Random access memory | 2 |  | PADZZN |
| -17 | RL07S103G | 81349 | -••RESISTOR, R3, R8, $\qquad$ <br> Fixed Film, 10K ohms, $\pm 2 \%, 1 / 4$ watt | 2 |  | PADZZN |

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FIGURE 7-8. MICROCOMPUTER ASSEMBLY, CIRCUIT CARD, A6A2 (Cont.)

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FIGURE 7-8. MICROCOMPUTER ASSEMBLY, CIRCUIT CARD, A6A2 (Cont.)

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NOTE: Refer to Parts List for differences in the R-2174A Receiver.


Figure 7-9. First L.O. Synthesizer Assembly, Circuit Card, A7 (Sheet I of 2)

FIGURE 7-9. FIRST L.O. SYNRHESIZER ASSEMBLY, CIRCUIT CARD, A7

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ |  | USABLE <br> ON CODE | $\begin{aligned} & \text { SMR } \\ & \text { CODE } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-9 | 08361 | 23386 | - FIRST LOCAL OSCILLATOR. SYNTHESIZER ASSEMBLY, Circuit Card, A7 | 1 |  | PAOLDT |
| 7-9 | 09134 | 23386 | - FIRST LOCAL OSCILLATOR. SYNTHESIZER ASSEMBLY, Circuit Card, A7 | 1 | A | PAOLDT |
| /1,2-1 | C320C103M1U1C1 | 31433 | -• CAPACITOR, C5, C7, C9, <br> C11, C13 thru C34, C39, <br> C44, C47, C48, C55, C85, <br> Ceramic, 0.01 uf, 100 <br> WVDC, $\pm 20 \%$ | 32 |  | PADZZN |
| /1,2-2 | CD4094BEX | 02735 | -. INTEGRATED CIRCUIT, $\qquad$ LB thru U14, 8 bit register, shift and store | 7 |  | PADZZN |
| /1,2-3 | RLO7SIO3G | 81349 | .. RESISTOR, R3, R50, R62,............... <br> R73, 81, R82, R83,Fixed <br> Film, 10K ohm, $\pm 2 \%, 1 / 4$ <br> Watt | 7 |  | PADZZN |
| /1-4 | CD40174BCN | 27014 | -. INTEGRATED CIRCUIT,UI8,. Hex D flip flop | 1 |  | PADZZN |
| /1-5 | 750-61-100K | 32997 | .. INTEGRATED) CIRCUIT,U16, ........... Transistor Array, 1OK ohm 6 pin | 1 |  | PADZZN |
| /1-6 | N82S83N | 18324 | .. INTEGRATED CIRCUIT,U15,............ 4 bit BCD adder | 1 |  | PADZZN |
| /1-7 | M38510/05701BCB | 81349 | .. INTEGRATED CIRCUIT,U20,............ U21, 18 bit static shift register | 2 |  | PADZZN |
| /1,2-8 | MS39014101-1593 | 81349 | -. CAPACITOR, C35, C37, $\qquad$ C42, C51, C59, C62 thru C77, C93, C96, Ceramic, 0.1 uf, 100 WVLC, $\pm 20 \%$ | 23 |  | PAZZN |
| /1,2-9 | T362A685M035AS | 31433 | - CAPACIT'OR, C1 thru C4, 06, C8, C12, C36, C38, C45, C46, C49, C50, C57, 060, 0C61, C79, C83, C84, C94, Tantalum, 6.8 uf, 35 WVIC, $\pm 20 \%$ | 20 |  | PADZZN |
| /1,2-10 | RL07S332G | 81349 | -. SISOR, F63, R54, R59. F69, F84, Fixed Film,3.3 K ohms, $\pm 2 \%, 1 / 4$ Watt | 5 |  | PADZZN |
| /1,2-11 | 1N916B | 07623 | -• DIOL, CR, CR2, CR5 thru CR16, CR19, CR20, Silicon | 16 |  | PADZZN |
| /1-12 | C312C222M1U1CA | 31433 | - CAPACITOR, M O, Ceramic. ............ |  |  | PADZZN |
| /1-13 | MILW-3861/9S-24 | 81349 | -- WIRE, Buss No. 24 ........................ | AR |  | PAOZZN |
| /1-14 | MIL, 1-22129C-24 | 81349 | -• TLBNG, r-TEFP No 24.................... | AR |  | PAOZZN |
| /1-15 | CA339E | 02735 | .. INTEGRATED CIRCUIT, U34 ............ Quad Voltage Comparator | 1 |  | P4DZZN |

FIGURE 7-9, FIRST L.O. SYNTHESIZER ASSEMBLY, CIRCUIT CARD, A7 (Cont.)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | 1234567 DESCRIPTION | $\begin{gathered} \hline \text { UNITS } \\ \text { PER } \\ \text { ASSY } \end{gathered}$ | USABLE <br> ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{array}{\|l\|} \hline 7-9- \\ 11,2-16 \end{array}$ | RL07S102G | 81349 | -• RESISTOR, R25, R61 <br> R63, R68, R70, R90, R95, R97, Fixed Film, I K ohms, $\pm 2 \%$, V4 watt | 8 |  | PADZZN |
| /1-17 | RL07S154G | 81349 | -. RESISTOR, R80, Fixed $\qquad$ <br> Film, 150 K ohms, $\pm 2 \%$, <br> V4 watt | 1 |  | PADZZN |
| /1-18 | C320C473MSUICA | 31433 | .. CAPACITOR, C89, Ceramic, $\qquad$ 0.047 uf, 100 WVDC, $\pm 20 \%$ | 1 |  | PADZZN |
| /1-19 | R107S104G | 81349 | .. RESISTOR, R77, $\qquad$ <br> Fixed Film, 100K ohms, <br> $\pm 2 \%$, '4 watt | 1 |  | PADZZN |
| 11-20 | RJR50FP201 | 81349 | -. RESISTOR, R5, Variable $\qquad$ <br> 200 ohms | 1 |  | PADZZN |
| /1,2-21 | RL07S151G | 81349 | -. RESISTOR, R4, R91 $\qquad$ Fixed Film, 150 ohms, $\pm 2 \%, 1 / 4$ watt | 2 |  | PADZZN |
| /1-22 | M38510/10707YB | 81349 | .. INTEGRATEDCIRCUIT U39, $\pm 12$ volt regulator | 1 |  | PADZZN |
| /1-23 | MS3564944 | 81349 | -. NUT, Hex, 440 (AP) ......................... | 1 |  | PAOZZN |
|  | MS35338-78 | 81349 | -. WASHER, Split Lock $\qquad$ <br> No. 4 (AP) | 1 |  | PAOZZN |
|  | MS15795-303 | 96906 | -. WASHER, Flat, No. 4 (AP) | 1 |  | PAOZZN |
|  | MS35233-14 | 96906 | -. SCREW, Machine, <br> Pan Head, 4-40 x 5/16 (AP) | 1 |  | PAOZZN |
| /1-24 | T362AI05M035AS | 31433 | -. CAPACITOR, C86, C87 $\qquad$ Tantalum, I uf, 35 WVDC, $- \pm 10 \%$ | 2 |  | PADZZN |
| /1,2-25 | RL07S471G | 81349 | -. RESISTOR, R24, R34 $\qquad$ <br> R35, R38, R40 thru R44, <br> R46, R48, R49, R55, R58, <br> R86, Fixed Film, 470 ohms, <br> $\pm 2 \%, 1 / 4$ watt | 15 |  | PADZZN |
| /1-26 | JANIN757A | 81349 | .. DIODE,CR18, $\qquad$ <br> Zener, 9.1 volts | 1 |  | PADZZN |
| /1-27 | CD4503BD/883 | 27014 | .. INTEGRATED CIRCUIT <br> U2, Hex 3 state buffer | 1 |  | PADZZN |
| /1, 2-28 | RL07S473G | 81349 | -. RESISTOR, R1, R87, R89, Fixed Film, 47K ohms, $\pm 2 \%, 1 / 4$ watt | 3 |  | PADZZN |
| /1-29 | CD4070BCN | 27014 | .. INTEGRATED CIRCUIT <br> U7, Quad exclusive OR gate | 1 |  | PADZZN |
| /1,2-30 | M38510/30106BEB | 81349 | -. INTEGRATED CIRCUIT $\qquad$ <br> U3, U5, U6, 6 bit D flip flop | 3 |  | PADZZN |
| /1-31 | M38510/30102BCB | 81349 | .. INTEGRATED CIRCUIT $\qquad$ <br> U19, Dual D flip flop | 1 |  | PADZZN |

FIGURE 7-9, FIRST L.O. SYNTHESIZER ASSEMBLY, CIRCUIT CARD, A7 (Cont.)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ |  | USABLE <br> ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & 7-9- \\ & / 1-32 \end{aligned}$ | M38510/30301BCB | 81349 | -. INTEGRATED CIRCUIT <br> U4, Quad 2 input NOR gate | 1 |  | PADZZN |
| /1-33 | CD40108BEX | 02735 | -. INTEGRATED CIRCUIT $\qquad$ U22, $4 \times 4$ multiport register | 1 |  | PADZZN |
| /1,2-34 | RL07S222G | 81349 | -. RESISTOR, R17, R52, $\qquad$ R56, Fixed Film, 2.2K ohms, $\pm 2 \%, 1 / 4$ watt | 3 |  | PADZZN |
| /1,2-35 | RL07S821G | 81349 | -. RESISTOR, RIS, R30, R33, R37, R47, Fixed Film, 820 ohms, $\pm 2 \%$, V4 watt | 5 |  | PADZZN |
| /1-36 | RL07S472G | 81349 | -. RESISTOR, R51 $\qquad$ <br> Fixed Film;4.7K ohms, $\pm 2 \%, 1 / 4$ watt | 1 |  | PADZZN |
| /1-37 | DAC20CQ | 31148 | -. INTEGRATED CIRCUIT $\qquad$ U23, Digital to analog converter | 1 |  | PADZZN |
| /1,2-38 | T362C336M025AS | 31433 | - CAPACITOR, C41, C56 $\qquad$ <br> Tantalum, 33 uf, 25 WVDC, $\pm 10 \%$ | 2 |  | PADZZN |
| /1,2-39 | MIL-S-19500/291B | 81349 | -. TRANSISTOR, Q6, Q7, Q8, Q11, Silicon, PNP, ZN4126 | 4 |  | PADZZN |
| /140 | LM145BN | 27014 | -. INTEGRATED CIRCUIT $\qquad$ U33, Dual operational amplifier | 1 |  | PADZZN |
| /1-41 | RL07S563G | 81349 | -. RESISTOR, R45 $\qquad$ Fixed Film, 56K ohms, $\pm 2 \%, / 4$ watt | 1 |  | PADZZN |
| /1,2-42 | JANTX2N2222A | 81349 | -. TRANSISTOR, Q9, Q10, Low power, NPN, 2N4124 | 2 |  | PADZZN I |
| /1-43 | RL07S392G | 81349 | -. RESISTOR, R57, Fixed $\qquad$ Film, 3.9K ohms, $\pm 2 \%, 4$ watt | 1 |  | PADZZN |
| /1,2-44 | RL07S123G | 81349 | -. RESISTOR, R67, R79 $\qquad$ Fixed Film, 12K ohms, $\pm 2 \%$, /4watt | 2 |  | PADZZN |
| /1-45 | CMR05E270GODR | 81349 | -. CAPACITOR,C80 $\qquad$ Mica, 27 pf, 100 WVDC, $\pm 5 \%$ | 1 |  | PADZZN |
| /1-46 | CMR05FIO1GODR | 81349 | -. CAPACITOR,C78, $\qquad$ Mica, 100 pf, 100 WVDC, $\pm 5 \%$ | 1 |  | PADZZN |
| /1, 247 | RL07S220G | 81349 | -. RESISTOR, R9, R22 $\qquad$ R60, Fixed Film, 22 ohms, $\pm 2 \%, 1 / 4$ watt | 3 |  | PADZZN |

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NOTE: Refer to Parts List for differences in the R-2174A Receiver.


Figure 7-9. First L.O. Synthesizer Assembly, Circuit Card, A7 (Sheet 2 of 2) CHANGE 2

FIGURE 7-9_FIRST L.O. SYNTHESIZER ASSEMBLY, CIRCUIT CARD, A7 (Cont)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | 12344567 DESCRIPTION | UNITS PER ASSY | USABLE on Code | $\begin{aligned} & \text { SMR } \\ & \text { CODE } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & 7-9- \\ & 11-48 \end{aligned}$ | C280MCH/A68K | 30983 | -• CAPACITOR, C81 $\qquad$ <br> Polyester, 0.001 uf, 400 WVDC, $- \pm 10 \%$ | 1 |  | PADZZN |
| /1-49 | C280MCH/A47K | 30983 | -• CAPACITOR, C82, $\qquad$ Polycarbonate, 0.047 uf, 100 WVDC, $\pm 10 \%$ | 1 |  | PADZZN |
| /1-50 | RL07S393G | 81349 | -•RESISTOR, R64 $\qquad$ <br> Fixed Film, 39K ohms, <br> $\pm 2 \%$, V4 watt | 1 |  | PADZZN |
| /1-51 | AD518JH883B | 24355 | -• INTEGRATEDCIRCUIT $\qquad$ U35, Operational amplifier analog devices | 1 |  | PADZZN |
| /1, 2-52 | RL07S153G | 81349 | -•RESISTOR, R65, R74 thru R76, Fixed Film, 15 K ohms, $\pm 2 \%, 1 / 4$ watt | 4 |  | PADZZN |
| /1-53 | CS13BE156M | 81349 | - CAPACITOR,C88, $\qquad$ Tantalum, $15 \mathrm{uf}, 20 \mathrm{WVDC}$, $\pm 20 \%$ | 1 |  | PADZZN |
| /2-54 | LTIOK129 | 81349 | - CHOKE, L1 thru L4A, Fixed RF, 6.8 uH , $\pm 10 \%$ | 4 |  | PADZZN |
| /2-55 | 3492-1002 | 75037 | -• CONNECTOR, J1,PCB, <br> Right angle, 20 way | 1 |  | PADZZN |
| /2-56 | RLR07C00GR | 81349 | -•RESISTOR, R2, R12, R28, R72, R92, R94, Fixed Film, 10 ohms, $\pm 2 \%, 1 / 4$ watt | 6 |  | PADZZN |
| /2-57 | M38510/054-OIBEA | 81349 | - •INTEGRATED CIRCUIT <br> U17, 4 bit full adder | 1 |  | PADZZN |
| /2-58 | M38510/05202BCB | 81349 | -•INTEGRATED CIRCUIT $\qquad$ U1, Quad 2 input NOR gate | 1 |  | PADZZN |
| /2-59 | 051-351-0000-220 | 98291 | - • CONNECTOR, J2, J3, J7, PCB, Coax, SMB, male | 3 |  | PADZZN |
| /2-60 | CMR05E470GODR | 81349 | - CAPACITOR, C103, C104, Mica, 47 pf, 500 WVDC, $\pm 5 \%$ | 2 |  | PADZZN |
| /2-61 | M38510/31505BEA | 81349 | -• INTEGRATED CIRCUIT, $\qquad$ U30, U31, Decode up-down counter | 2 |  | PADZZN |
| /2-62 | RLO7S221G | 81349 | -• RESISTOR, R31, R32, R36, Fixed Film, 220 ohms, $\pm 2 \%, 1 / 4$ watt | 3 |  | PADZZN |
| /2-63 | T362C157K006AS | 31433 | - CAPACITOR, C53, <br> Tantalum, 150 uf, 6 WVDC, $\pm 20 \%$ | 1 |  | PADZZN |
| /2-64 | F10231PC | 07263 | - - INTEGRATED CIRCUIT $\qquad$ U26, U28, Dual D flip flop | 2 |  | PADZZN |

FIGUBE 7-9, FIRST L.O. SYNTHESIZER ASSEMBLY, CIRCUIT CARD, A7 (Cont)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | 1234567 DESCRIPTION | $\begin{aligned} & \text { ARD, A7 } \\ & \text { UNITS } \\ & \text { PER } \\ & \text { ASSY } \end{aligned}$ | USABLE <br> ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-9- |  |  |  |  |  |  |
| /2-65 | F10211PC | 07263 | -•INTEGRATED CIRCUIT $\qquad$ <br> U32, Dual NOR gate | 1 |  | PADZZN |
| /2-66 | LM1458N | 27014 | - - INTEGRATED CIRCUIT $\qquad$ U37, Dual operational amplifier | 1 |  | PADZZN |
| /2-67 | CM06FD472JN3 | 81349 | - CAPACITOR, C91, Mica, 4700 pf, 100 WVDC, $\pm 5 \%$ | 1 |  | PADZZN |
| /2-68 | AD201A | 24355 | - - INTEGRATED CIRCUIT, U36, Analog switch | 1 |  | PADZZN |
| /2-69 | RL07S124G | 81349 | -• RESISTOR, R88, $\qquad$ Fixed Film, 120K ohms, $\pm 2 \%, 1 / 4$ watt | 1 |  | PADZZN |
| /2-70 | RL07S822G | 81349 | -•RESISTOR, R66, Fixed $\qquad$ <br> Film, 8.2K ohms, <br> $\pm 2 \%, 1 / 4$ watt | 1 |  | PADZZN |
| /2-71 | RL07S272G | 81349 | -•RESISTOR, R78, Fixed $\qquad$ Film, 2.7K ohms, $\pm 2 \%, 1 / 4$ watt | 1 |  | PADZZN |
| /2-72 | CMR05F222GODR | 81349 | - - CAPACITOR, C92 <br> Mica, 2200 pf, <br> 100 WVDC, $\pm 5 \%$ | 1 |  | PADZZN |
| /2-73 | CY15C102M | 71590 | - CAPACITOR, C40, C43, C52, C54, C95, Ceramic, 0.001 uf, 100 WVDC, $\pm 20 \%$ | 5 |  | PADZZN |
| /2-74 | CA3140BT/3 | 32293 | - • INTEGRATED CIRCUIT $\qquad$ U40, Operational amplifier | 1 |  | PADZZN |
| /2-75 | RL07S681G | 81349 | -•RESISTOR, RI1, R93, $\qquad$ Fixed Film, 680 ohms, $\pm 2 \%, 1 / 4$ watt | 2 |  | PADZZN |
| /2-76 | C280MCH/AIM | 30983 | - CAPACITOR, C97 $\qquad$ <br> Polycarbonate, 1.0 uf, 100 WVDC, $\pm 10 \%$ | 1 |  | PADZZN |
| /2-77 | MKC-1860-533/06 | 52763 | - - CAPACITOR, C98, <br> Polycarbonate, 3.3 uf, 63 WVDC, $\pm 10 \%$ | 1 |  | PADZZN |
| /2-78 | 2N4921 | 04713 | - •TRANSISTOR, Q4 $\qquad$ High Power, NPN, 2N4921 | 1 |  | PADZZN |
| /2-79 | 08670 | 23386 | - - WASHER, Q4, $\qquad$ Insulator, modified | 1 |  | MOOZZN |
| /2-80 | 120-5 | 05820 | - COMPOUND, Thermal, Wakefield No. 128 | AR |  | PAOZZN |
| /2-81 | RLO7S470G | 81349 | -•RESISTOR, R18 $\qquad$ Fixed Film, 47 ohms, $\pm 2 \%, 1 / 4$ watt |  |  | PADZZN |

FIGURE 7-9_FIRST L.O. SYNTHESIZER ASSEMBLY. CIRCUIT CARD A7 (Cont)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | 1234567 DESCRIPTION | UNITS PER ASSY | USABLE <br> ON CODE | $\begin{array}{\|c} \text { SMR } \\ \text { CODE } \end{array}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-9 |  |  |  |  |  |  |
| /2-82 | M38510/10201BEA | 81349 | -• INTEGRATED CIRCUIT, $\qquad$ U24, Voltage regulator | 1 |  | PADZZN |
| /2-83 | SK2-1/4-1ROJ | 52783 | - • ESISTOR, R19, $\qquad$ Composition, 1 ohm, $5 \%$, 1/4 Watt | 1 |  | PADZZN |
| /2-84 | JAN2N2369A | 81349 | - • TRANSISTOR, Q5, Q13, Silicon, NPN, 2N2369 | 2 |  | PADZZN |
| /2-85 | 69011-1058 | 78912 | - - PAD,Transistor,Q5,Q13,................... | 2 |  | PALEZN |
| /2-86 | RL07S501G | 81349 | -•RSISTOR,R16,Variable, 500 ohms | 1 |  | PADZZN |
| /2-87 | M38510/31506BEA | 81349 | -• INTEGRATED CIRCUIT, U29, Binary up-down counter | 1 |  | PADZZN |
| /2-88 | 11C90DCQR | 07263 | -• INTEGRATED CIRCUIT, .................... U27, Divider | 1 |  | PADZZN |
| /2-89 | RL07S820G | 81349 | -. RESISTOR, R71, Fixed $\qquad$ <br> Film, 82 ohm, $\pm 2 \%, 1 / 4$ Watt | 1 |  | PADZZN |
| /2-90 | CMIO5DIOOGODR | 81349 | - - CAPACITOR, C58, CO11 $\qquad$ Mica, 10 pf., 500 WVLC, $\pm 1 / 2 \mathrm{pf}$ | 2 |  | PADZZN |
| /2-90a | CMRD5E820GODR | 81349 | - CAPACITOR, C101, Mica,.................. $82 \mathrm{pf}, \pm 2 \%, 500$ WVDC | 1 | A | PADZZN |
| /2-91 | RL07SIO1G | 81349 | -• RE SISTOR, R6, R29, R85,. Fixed Film, 100 ohms, $\pm$ $2 \%, 1 / 4$ Watt | 3 |  | PADZZN |
| /2-92 | RL07S331G | 81349 | - - RESISTOR, R21, R27, $\qquad$ <br> Fixed Film, 330 ohms, $\pm$ $2 \%, 1 / 4$ Watt | 2 |  | PADZZN |
| /2-93 | RL07S152G | 81349 | - $\cdot$ RESISTOR, RIO, R26, Fixed Film, 1.5 K ohms, $\pm$ $2 \%, 1 / 4$ Watt | 2 |  | PALMZN |
| /2-94 | JAN2N918 | 81349 | - •TRANSISTOR,Q12,Silicon, $\qquad$ Low Power, NPN | 1 |  | PADZZN |
| /2-95 | 69011-1058 | 78912 | - P PAD, Transistor, Q12, ..................... | 1 |  | PADZZN |
| /2-96 | MI LS-19500/428 | 81349 | -•TRANSISTOR, Q1, Field..................... | 1 |  | PADZZN |
| /2-97 | 69011-1058 | 78912 | - P PAD, Transistor, Q1 ....................... | 1 |  | PALZZN |
| /2-98 | RLD7S680G | 81349 | -. RESISTOR, R7, RS, R14, R23, Fixed Film, 68 ohms, $\pm 2 \%, 1 / 4$ Watt | 4 |  | PADZZN |
| /2-99 | 08304 | 23386 | - - COIL ASSMBLY, L5, Local.................. oscillator, air wound | 1 |  | PADZZN |
| /2-99a | 09255 | 23386 | - . COIL ASSEMBLY, L5,Local................. oscillator, air wound | 1 | A | PADZZN |
| /2-100 | LKV6522B | 52673 | -• DIOEE, CR3, CR4, $\qquad$ <br> Varactor | 2 |  | PADZZN |

FIGURE 7-9 FIRST L.O. SYNTHESIZER ASSEMBLY, CIRCUIT CARD, A7 (Cont)


NOTE: Refer to Parts List for differences in the R-2174A Receiver.


Figure 7-10. Second L.O./BFO Synthesizer Assembly, Circuit Card, A8

FIGURE 7-10 SECOND LOCAL OSCILLATOR SYNTHESIZER ASSEMBLY, CIRCUIT CARD, A8

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | 1234567 DESCRIPTION | UNITS PER ASSY | USABLE <br> ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-10 | 08387 | 23386 | -SEOOND LOCAL OSCILLATOR ........... SYNTHESIZER ASSEMBLY, Circuit Card, A8 | 1 |  | PAODDR |
| 7-10 | 09632 | 23386 | - SECOOND LOCAL OSCILLATOR......... SYNTHESIZER ASSEMBLY, Circuit Card, A8 | 1 | A | PAODDT |
| -1 | RL07S821G | 81349 | -•RESISTOR, R22, R35, R76, R77, Fixed Film, 820 ohms, $\pm 2 \%, 1 / 4$ Watt | 4 |  | PADZZN |
| -2 | LT10K020 | 81349 | -• INDIUCOR, L6, RF, 220 $\mathrm{uH}, \pm 10 \%$ | 1 |  | PAEZZN |
| -3 | CMR05F151G0DR | 81349 | - - CAPACITOR, C28, C64, Mica, 150 pf, 500 WVDC,. $\pm 2 \%$ | 2 |  | PAEZZN |
| -4 | CM06FI21.JN3 | 81349 | -• CAPACITOR, 063, C65, Mica, 820 pf, 500 WVLUC,. $\pm 5 \%$ | 2 |  | PADZZN |
| -5 | MS39014101-1593 | 81349 | - • CAPACITOR, C3, C4, C9, ................... <br> C11, C12, C13, C15 thru <br> C19, C21, C22, C24, C27, <br> C30, C39, C43, C44, C45, <br> C47, 048, 062, C66 thru <br> C71, C73, C74, C75, <br> Ceramic, 0.1 uf, 50 WVLC $\pm 20 \%$ | 32 |  | PADZZN |
| -6 | RL07S331G | 81349 | -•ESISTOR, R11, R27, R69 <br> - R70, Fixed Film, 330 ohm $\pm 2 \%, 1 / 4$ Watt | 4 |  | PADZZN |
| -6a | Not Used |  | - R70 .............................................. | , | A |  |
| -7 | C320C103M1U1C1 | 31433 | -•CAPACITOR, C5, C8, C14, <br> C32, C42, C52, C53, C58 <br> thru 061, C72, C78, C80, <br> Cer., 0.01 uf, 50 WVDC <br> $\pm 20 \%$ | 14 |  | PADZZN |
| -8 | 11C90DCQR | 07263 | -• INTEGRATED CIRCUIT, U19............. <br> Divide by 10/11, <br> Prescaler | 1 |  | PADZZN |
| -9 | 051-351-0000-220 | 98291 | -•CONNECTOR, J1 thru J4,.................. SMB, Fush on | 4 |  | PADZZN |
| -10 | M38510/3031BCB | 81349 | - • INTEGRATED CIRCUIT, LB,.............. <br> U9, U20, Dual decade counter | 3 |  | PAOODT |
| -11 | M38510/30301BCB | 81349 | -• INTEGRATED CIRCUIT, U12, Quad 2 NOR gate | 1 |  | PADZZN |
| -12 | LT10K128 | 81349 | -• INDUCTOR, L7, LS, RF $5.6 \mathrm{uH}, \pm 10 \%$ | 2 |  | PADEZN |
| -13 | T362A685M035AS | 31433 | - CAPACITOR, C1, C23, C49. C54, C79, C81, Tantalum, 6.8 uf, 35 WVDC, $\pm 20 \%$ | 6 |  | PADZZN |
| -14 | M38510/31503BEA | 81349 | .. INTEGRATED CIRCUIT, U14............. <br> thru U18,Synchronous <br> 4 bit counter | 5 |  | PADZZN |

FIGURE 7-10 SECOND LOCAL OSCILLATOR SYNTHESIZER ASSEMBLY, CIRCUIT CARD, A8 (Cont.)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | 1234567 DESCRIPTION | $\begin{aligned} & \text { UNITS } \\ & \text { PER } \\ & \text { ASSY } \end{aligned}$ | USABLE ON CODE | $\begin{aligned} & \text { SMR } \\ & \text { CODE } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-10 |  |  |  |  |  |  |
| -15 | 3429-1002 | 75037 | - - CONNECIOR, J5, Control/ <br> Power, right angle, 26 pin | 1 |  | PADZZN |
| -16 | Not Used |  | - - CONNECTOR, J6, J7, J8, ................... <br> Not Used | 3 |  |  |
| -16a | 051-351-0000-220 | 98291 | . . CONNECTOR, J8, SMB, ................... Push on | 1 | A | PAEZZN |
| -17 | LT10K012 | 81349 | - • INDUCTOR, L1, RF, 100 $\qquad$ $\mathrm{uH}, \pm 10 \%$ | 1 |  | PADZZN |
| -18 | C320C473M5U1CA | 31433 | - CAPACITOR, C6, C7, <br> Ceramic, 0.047 uf, 50 <br> WVDC, $\pm 20 \%$ | 2 |  | PADZZN |
| -19 | RL07S182G | 81349 | -. RESISTOR, R19, R21, $\qquad$ <br> R75, Fixed Film, 1.8K <br> ohms, $\pm 2 \%, 1 / 4$ Watt | 3 |  | PADZZN |
| -20 | CMROSF101G0DR | 81349 | - - CAPACITOR, C2, C34, C35, C37, C46, C56, C57, Mica, 100 pf, 500 WVLC, $\pm 5 \%$ | 7 |  | PAEZZN |
| -21 | MIL-S-19500/428 | 81349 | - TRANSISTOR, Q18, Q19, $\qquad$ <br> Field effect | 2 |  | PADZZN |
| -22 | 7717-HWHT | 13103 | - PAD, Transistor, Q18, $\qquad$ Q19 | 2 |  | PADZZN |
| -23 | RLD7S681G | 81349 | - . RESISTOR, R6, R33, R34 R38, R59, 163, R67,Fixed Film, 68 ohms, $\pm 2 \%, 1 / 4$ Watt | 7 |  | PADZZN |
| -24 | CMR05E560G0DR | 81349 | - . CAPACITOR, C55, Mica,.................... 56 pf, 500 WVIC, $\pm 5 \%$ | 1 |  | PADZZN |
| -25 | RL07S102G | 81349 | - . RESISTOR, R10, RU3, R23. <br> R41, R43, R44, R45, R48, <br> R49, R54, 157, R58, R74, <br> R84, Fixed Film,, 1K ohm, <br> $\pm 2 \%, 1 / 4$ Watt | 14 |  | PADZZN |
| -26 | RLR07C564GR | 81349 | -. RESISTOR, 66, 68 $\qquad$ Fixed Film, 560 K ohms, $\pm$ 2\%, 1/4 Watt | 2 |  | PALZZN |
| -27 | RL07S472G | 81349 | - FESISTOR, R7, R24, R25, $\qquad$ R72, Fixed Film, 4.7K ohms, $\pm 2 \%, 1 / 4$ Watt | 4 |  | PAZLZN |
| -28 | JAN2N2369A | 81349 | - TRANSISTOR, Q1 thru Q6, Q10, Q11, Q20, Switching NPN | 9 |  | PALEZN |
| -29 | 7717-HWHT | 13103 | - PAD, Transistor, Q1 $\qquad$ thru Q6, Q10, Q11, Q20 | 9 |  | PADZZN |
| -30 | 4N28 | 07263 | .. INTEGRATED CIRCUIT, ................... <br> U21, opto isolator | 1 |  | PADZZN |
| -30a | Not Used |  | .. INTEGRATED CIRCUIT, ................... <br> U21, opto isolator | 1 | A | PADZZN |

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| FIGURE \& INDEX NO. | PART NUMBER | FSCM | 1234567 DESCRIPTION | $\begin{aligned} & \text { UNITS } \\ & \text { PER } \\ & \text { ASSY } \end{aligned}$ | USABLE <br> ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & 7-10- \\ & -31 \end{aligned}$ | RL07S103G | 81349 | -•RESISTOR, R28, R29, R36, R47, R61, R73, R78, Fixed Film, 10K ohms, $\pm 2 \%, 1 / 4$ watt | 7 |  | PADZZN |
| -32 | LT100K138 | 81349 | - • INDUCTOR, L3, L5 <br> RF, $33 \mathrm{uH}, \pm 5 \%$ | 2 |  | PADZZN |
| -33 | RLR07CI00GR | 81349 | -•RESISTOR, R12, R26 $\qquad$ <br> Fixed Film, 10 ohms, $\pm 2 \%, 1 / 4$ watt | 2 |  | PADZZN |
| -34 | RLR07C330GR | 81349 | - RESISTOR, R9, Fixed $\qquad$ <br> Film, 33 ohms, $\pm 2 \%$, <br> 1/4 watt | 1 |  | PADZZN |
| -35 | RL07S470G | 81349 | - RESISTOR, R8, R14 $\qquad$ R32, R83, Fixed Film, 47 ohms, $\pm 2 \%, 1 / 4$ watt | 4 |  | PADZZN |
| -36 | 08289 | 23386 | . - OSCILLATOR, Y1 Oven controlled crystal | 1 |  | PADZZN |
| -37 | 1N916B | 07623 | - DIODE, CR2, CR3 CR5, Silicon | 3 |  | PADZZN |
| -38 | 2N3904 | 04713 | -•TRANSISTOR, Q7, Q9 Q15, Q17, Low power, NPN | 4 |  | PADZZN |
| -39 | 08525 | 23386 | -• INDUCTOR, L4 <br> RF, Variable | 1 |  | PADZZN |
| -40 | MIL-S-19500/436 | 81349 | -• DIODE, CR6, CR7, Varicap | 2 |  | PADZZN |
| -41 | RL07S183G | 81349 | -•RESISTOR, R65 <br> Fixed Film, 18K ohms, <br> $\pm 2 \%, 1 / 4$ watt | 1 |  | PADZZN |
| -42 | C280MCH/A100K | 30983 | - - CAPACITOR, C26, C51 <br> Polycarbonate, 0.1 uf, 100 WVDC, $\pm 20 \%$ | 2 |  | PADZZN |
| -43 | C280MCH/A470K | 30983 | - . CAPACITOR, C50 $\qquad$ <br> Polycarbonate, 0.47 uf, 100 WVDC, $\pm 20 \%$ | 1 |  | PADZZN |
| -44 | 2N4126 | 04713 | -•TRANSISTOR, Q8, Q16, Low power, PNP | 2 |  | PADZZN |
| -45 | RL07S332G | 81349 | -•RESISTOR, R18, R20, R37, R62, Fixed Film, 3.3 K ohms, $\pm 2 \%, 1 / 4$ watt | 4 |  | PADZZN |
| -46 | RLO7S222G | 81349 | .. RESISTOR, R5, R39 $\qquad$ R60, R64, Fixed Film, 2.2 K ohms, $\pm 2 \%, 1 / 4$ watt | 4 |  | PADZZN |
| -47 | M38510/31004BCB | 27014 | .. INTEGRATED CIRCUIT $\qquad$ U13, Quad 2 input AND | 1 |  | PADZZN |
| -48 | MLL-W-3861/S-24 | 81349 | - - WIRE, LK1, LK2 <br> Links, Buss No. 24 | AR |  | PAOZZN |
| -49 | MIL-I-22129C-24 | 81349 | -•TUBING, LK1, LK2 <br> PTFE, No. 24 | AR |  | PAOZZN |



FIGURE 7-10 SECOND LOCAL OSCILLATOR SYNTHESIZER ASSEMBLY, CIRCUIT CARD,A8 (Cont.)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | 1234567 DESCRIPTION | $\begin{aligned} & \text { UNITS } \\ & \text { PER } \\ & \text { ASSY } \end{aligned}$ | USABLE ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & 7-10- \\ & -69 \end{aligned}$ | MIL-S-19500/383 | 81349 | - DIODE, CR4, | 1 |  | PADZZN |
| -70 | CY15C102M | 71590 | Varicap, 22 pf <br> - CAPACITOR,C36, $\qquad$ <br> C40, C87, Ceramic, <br> 0.001 uf, 50 WVDC, <br> $\pm 20 \%$ | 3 |  | PADZZN |
| -71 | DM10115N | 27014 | - • INTEGRATED CIRCUIT U22, Line Receiver | 1 |  | PADZZN |
| -72 | RL07S101G | 81349 | - - RESISTOR, R53 $\qquad$ Fixed Film, 100 ohms, $\pm 2 \%, 14$ watt | 1 |  | PADZZN |
| -73 | CMR05C120GODR | 81349 | - CAPACITOR,C20, Mica, 12 pf, 500 WVDC, $\pm 5 \%$ | 1 |  | PADZZN |
| -74 | RL07S560G | 81349 | -•RESISTOR, R46, $\qquad$ Fixed Film, 56 ohms, $\pm 2 \%, 1 / 4$ watt | 1 |  | PADZZN |
| -75 | RL07S221G | 81349 | - •RESISTOR, R52, Fixed $\qquad$ <br> Film, 220 ohms, $\pm 2 \%$, <br> 1/4 watt | 1 |  | PADZZN |
| -76 | CMR05E680G0DR | 81349 | - - CAPACITOR, C33, C41 $\qquad$ Mica, 68 pf, 500 WVDC, $\pm 2 \%$ | 2 |  | PADZZN |
| -77 | CS13BE335M | 81349 | - CAPACITOR, C25, <br> Tantalum, 3.3 uf, <br> 15 WVDC, $\pm 10 \%$ | 1 |  | PADZZN |
| -78 | LT4K081 | 81349 | - • INDUCTOR, L2, <br> RF, $1 \mathrm{uH}, \pm 10 \%$ | 1 |  | PADZZN |
| -79 | CMR05E820G0DR | $81349$ | * CAPACITOR, C76, Mica, 82 pf, 500 WVDC, $\pm 2 \%$ | 1 |  | PADZZN |
| -80 | 08522 | 23386 | - • INDUCTOR, L9, L10, Variable | 2 |  | PADZZN |
| -81 | CMR05F331G0DR | $81349$ | - - CAPACITOR, C77, <br> Mica, 330 pf, 500 WVDC, $\pm 2 \%$ | 1 |  | PADZZN |
| -82 | MS39014102-1419 | 81349 | .. CAPACITOR, C75, Ceramic, .............. 1 uf, 50 WVDC, $\pm 20 \%$ | 1 |  | PADZZN |



Figure 7-11. Receiver Control Assembly, Circuit Card, A9 (Sheet 1 of 2)
T.O. 31R2-2URR-251

FIGURE 7-11_RECEIVER CONTROL_ASSEMBLY, CIRCUIT CARD, A9

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | 12334567 DESCRIPTION | UNITS PER ASSY | USABLE ON CODE | $\begin{array}{\|c} \text { SMR } \\ \text { CODE } \end{array}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-11- | 08388 | 23386 | - RECEIVERCONTROL $\qquad$ ASSEMBLY, Circuit Card, A9 | 1 |  | PAFLDT |
| /1,2-1 | HBLB17S-5 | 09922 | -• CONNECTOR,J2,J3, Switch Panel, 17 contact | 2 |  | PADZZN |
| /1, 2-2 | SIP-8-103 | 91637 | -. INTEGRATED CIRCUIT, $\qquad$ <br> U1, US, Resistor array, <br> 10K ohms, 7 resistors | 2 |  | PADZZN |
| /1, 2-3 | 4310R-101-103 | 32997 | - • INTEGRATEDCIRCUIT, $\qquad$ U2, U6, U30, Resistor array, 10K ohms, 9 resistors | 3 |  | PADZZN |
| /1,2-4 | C320C103M1IJ1C1 | 31433 | - - CAPACITOR, CI thru C7, C9 thru C13, C16, C19 thru C24, Ceramic, 0.01 uf, 50 WVDC, $\pm 20 \%$ | 19 |  | PADZZN |
| /1-5 | 08536 | 23386 | - CABLE ASSEMBLY, WI .................. | 1 |  | XA |
| /1-6 | 3402-0000T | 75037 | - CONNECTOR, P1 <br> 34 contact | 1 |  | XA |
| /1-7 | 3414-6034 | 75037 | - CONNECTOR, <br> 35 contact | 1 |  | XA |
| /1,2-8 | CD4503BCN | 27014 | - - INTEGRATED CIRCUIT U7, U8, U9, U23, U24, U25, U42, U43, Hex 3 state buffer | 8 |  | PADZZN |
| /1-9 | CD4066BEX | 02735 | - • INTEGRATED CIRCUIT $\qquad$ <br> U27, U28, Quad <br> bilateral switch | 2 |  | PADZZN |
| /1-10 | CD4051BEMJ/883B | 02735 | .. INTEGRATED CIRCUIT, $\qquad$ U29, 8 channel analog multiplexer | 1 |  | PADZZN |
| /1-11 | 08573 | 23386 | - CONNECTOR, J8, 4 contact | 1 |  | PADZZN |
| /1, 2-12 | RL07S103G | 81349 | - - RESISTOR, R2, R6, $\qquad$ R27, R28, Fixed Film, 10 K ohms, $\pm 2 \%, 1 / 4$ watt | 4 |  | PADZZN |
| /1-13 | JAN2N2369A | 81349 | - •TRANSISTOR, Q1 Switching, NPN | 1 |  | PADZZN |
| /1-14 | RL07S102G | 81349 | - •RESISTOR, R3, R14, R15, Fixed Film, IK ohms, $\pm 2 \%, 1 / 4$ watt | 3 |  | PADZZN |
| /1-15 | MC14508BBJBS | 04713 | - - INTEGRATED CIRCUIT, <br> U46, Dual 4 bit latch | 1 |  | PADZZN |
| /1,2-16 | CD4514BF/3 | 02735 | - . INTEGRATED CIRCUIT, $\qquad$ U35, U47, address decoder | 2 |  | PADZZN |
| /1-17 | DTZ-20 | 71590 | - CAPACITOR, C14, C15 $\qquad$ Ceramic, 20 pf, non polarized, $\pm 5 \%$ | 2 |  | PADZZN |
| /1-18 | RL07S333G | 81349 | -•RESISTOR, R4, R5 $\qquad$ Fixed Film 33K ohms, $\pm 2 \%, 14$ watt | 2 |  | PADZZN |

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FIGURE 7-11 RECEIVER CONTROL_ASSEMBLY, CIRCUIT CARD, A9 (Cont.)

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | 12334567 DESCRIPTION | UNITS PER ASSY | USABLE <br> ON CODE | $\begin{array}{\|c} \text { SMR } \\ \text { CODE } \end{array}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & 7-11- \\ & 11-19 \end{aligned}$ | CD4528BF | 27014 | -• INTEGRATED CIRCUIT, <br> U55, one shot multivibrator | 1 |  | PADZZN |
| /1-20 | M38510/05204BCD | 81349 | - - INTEGRATED CIRCUIT $\qquad$ U48, Triple 3 input NOR gate | 1 |  | PADZZN |
| /1-21 | CD4069UBE | 02735 | - - INTEGRATED CIRCUIT, U56, Hex inverter | 1 |  | PADZZN |
| /1-22 | M38510/17001BCD | 27014 | - . INTEGRATEDCIRCUIT $\qquad$ <br> U57, Quad 2 input AND | 1 |  | PADZZN |
| /1-23 | RL07S101G | 81349 | -• RESISTOR, R7, Fixed Film, 100 ohms, $\pm 2 \%, 1 / 4$ watt | 1 |  | PADZZN |
| /1-24 | 3428-2002 | 70537 | - - CONNECTOR, J6 <br> 20 contact | 1 |  | PADZZN |
| /1-25 | 08270 | 23386 | - - INTEGRATED CIRCUIT, $\qquad$ U3, Liquid crystal display, Frequency | 1 |  | PBOZZN |
| /1-26 | WB-11-55-G | 06776 | .. SOCKET, XU3A <br> 11 contact | 4 |  | XA |
| /1-27 | WB-20-55-G | 06776 | . . SOCKET, XU3B, <br> 20 contact | 2 |  | XA |
| /1,2-28 | CD4056BD | 02735 | - - INTEGRATED CIRCUIT, $\qquad$ U10 thru U18, U31 thru U34, Display drivers, 7 segment | 13 |  | PADZZN |
| /1-29 | HP50824468 | 28480 | .. DISPLAY, DS5, $\qquad$ Light emitting diode/ resistor assembly | 1 |  | PADZZN |
| $\begin{aligned} & 11-30 \\ & / 1,2-31 \end{aligned}$ | $\begin{aligned} & \text { MIL---22129C-20 } \\ & \text { CD4054BD } \end{aligned}$ | $\begin{aligned} & 81349 \\ & 02735 \end{aligned}$ | -•SLEEVING, DS5, $\qquad$ <br> -• INTEGRATED CIRCUIT $\qquad$ U19 thru U22, U36 thru U40, U49, Display drivers, 7 segment | $\begin{gathered} \text { AR } \\ 10 \end{gathered}$ |  | $\begin{aligned} & \text { PAOZZN } \\ & \text { PADZZN } \end{aligned}$ |
| /1-32 | CD40N28BCN | 27014 | - - INTEGRATED CIRCUIT $\qquad$ <br> U50, Decimal decoder | 1 |  | PADZZN |
| /1,2-33 | RL07S392G | 81349 | . - RESISTOR, R8, R9, $\qquad$ R24, R25, Fixed Film, 3.9K ohms, $\pm 2 \%, 1 / 4$ watt | 4 |  | PADZZN |
| /1-34 | RL07S113G | 81349 | -•RESISTOR, R10, R11 Fixed Film, 11K ohms, $\pm 2 \%, 1 / 4$ watt | 2 |  | PADZZN |
| /1-35 | CK60BX101K | 81349 | -. CAPACITOR,C17,C18 Ceramic, 100 pf, non polarized, $\pm 10 \%$ | 2 |  | PADZZN |
| /1-36 | CA339E | 02735 | -• INTEGRATED CIRCUIT $\qquad$ U58, Quad voltage comparator | 1 |  | PADZZN |


| FIGURE \& INDEX No. | $\begin{aligned} & \text { PART } \\ & \text { NUMBER } \end{aligned}$ | FSCM | 1234567 DESCRIPTION | $\begin{aligned} & \text { UNTITS } \\ & \text { PER } \\ & \text { ASSY } \end{aligned}$ | USABLE <br> ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & 7-11- \\ & 11-37 \end{aligned}$ | RLR07C105GR | 81349 | -• RESISTOR, R12, R13, $\qquad$ <br> Fixed Film, 1 meg ohms, <br> $\pm 2 \%, 1 / 4$ watt | 2 |  | PADZZN |
| /1-38 | RL07S104G | 81349 | .. RESISTOR, R16, R17 <br> Fixed Film, 100K ohms, $\pm 2 \%, 1 / 4$ watt | 2 |  | PADZZN |
| /1-39 | RL07S223G | 81349 | - . RESISTOR, R18, R19 R29 thru R32, Fixed Film, 22K ohms, $\pm 2 \%$, 1/4 watt | 6 |  | PADZZN |
| /1-40 | RL07S181G | 81349 | -• RESISTOR, R20, R21, Fixed Film, 180 ohms, $\pm 2 \%$, 14 watt | 2 |  | PADZZN |
| /141 | OPB706A | 07374 | -• INTEGRATED CIRCUIT $\qquad$ U59, U60, Optical transducer | 2 |  | PADZZN |
| /1-42 | 08473 | 12697 | .- RESISTOR, R22, R23, Potentiometer, audio line, 25 K ohms | 2 |  | PADZZN |
| /2-43 | 08526 | 23386 | .. INTEGRATED CIRCUIT, $\qquad$ <br> U4, Liquid crystal display mode | 1 |  | PBOZZN |
| /2-44 | WB-12-55-G | 06776 | .. SOCKET, XU4A <br> 12 contact | 2 |  | XA |
| /2-45 | WB-14-55-C | 06776 | .. SOCKET, XU4B, <br> 14 contact | 2 |  | XA |
| /2-46 | 3432-2002 | 75037 | .- CONNECTOR, J5 <br> IF, 40 contact | 1 |  | PADZZN |
| /2-47 | 3429-2002 | 75037 | -• CONNECTOR, J4, J7, $\qquad$ BFO, Power, 26 contact | 2 |  | PADZZN |
| /2-48 | RL07S153G | 81349 | .. RESISTOR, R26 Fixed Film, 15K ohms, $\pm 2 \%, 1 / 4$ watt | 1 |  | PADZZN |
| /2-49 | CD4076BD | 27014 | - INTEGRATED CIRCUIT, U61 thru U64, 8 bit addressable latch | 4 |  | PADZZN |
| /2-50 | 2N3906 | 04713 | -• TRANSISTOR, Q2 <br> Low power, PNP, <br> 2N3906 | 1 |  | PADZZN |
| /2-51 | MS39014101-1593 | 81349 | - CAPACITOR, C8, <br> Ceramic, 0.1 uf, 50 WVDC, $\pm 20 \%$ | 1 |  | PADZZN |
| /2-52 | RL07S203G | 81349 | .. RESISTOR, R1 Fixed Film, 20K ohms, $\pm 2 \%, 1 / 4$ watt | 1 1 |  | PADZZN |
| /2-53 | CD4047BEMJ/883B | 02735 | - • INTEGRATED CIRCUIT $\qquad$ <br> U26, Monostable, <br> Astable multivibrator | 1 |  | PADZZN |



Figure 7-11. Receiver Control Assembly, Circuit Card, A9 (Sheet 2 of 2)

FIGURE 7-11 RECEIVER CONTROL_ASSEMBLY, CIRCUIT CABD, A9 (Cont.)



Figure 7-12. Power Supply Assembly, Module, A10 (Sheet 1 of 2)

FIGURE 7-12. POWER SUPPLY ASSEMBLY, MODULE, A10

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | $\begin{array}{llllllll}1 & 2 & 3 & 4 & 5 & 6 & 7 & \text { DESCRIPTION }\end{array}$ | $\begin{aligned} & \text { UNITS } \\ & \text { PER } \\ & \text { ASSY } \end{aligned}$ | USABLE <br> ON CODE | $\begin{gathered} \text { SMR } \\ \text { CODE } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7-12 | 08389 | 23386 | -• POWERSUPPLYASSEMBLY.............. Module, A10 | 1 |  | PAOODT |
| /1-1 | 08518 | 23386 | - COVER, Power Supply ..................... | 1 |  | XB |
| /1-2 | $\begin{aligned} & \text { 3186EA12340 } \\ & \text { 25AMA2 } \end{aligned}$ | 30983 | - CAPACITOR, C1, $\qquad$ Electrolytic, 12,500 uf, 25 WVDC, $-10+75 \%$ | 1 |  | PAOZZN |
| /1-3 | MS3564948 | 81349 | - NUT, Hex, 6-32 (AP) ....................... | 3 |  | PAOZZN |
|  | MS35338-79 | 81349 | - WASHER, Split Lock $\qquad$ <br> No. 6 (AP) | 3 |  | PAOZZN |
|  | MS15795-305 | 81349 | -. WASHER', Flat, $\qquad$ No. 6 (AP) | 3 |  | PAOZZN |
|  | AN507C-632-6 | 81349 | - SCREW, Machine, Flat Head $\qquad$ $6-32 \times 3 / 8 \text { (AP) }$ | 3 |  | PAOZZN |
| /1-4 | 458648 | 56289 | - CLAMP, Capacitor, C1 ..................... | 1 |  | PAOZZN |
| /1-5 | MS3564948 | 81349 | - WASHER, Split Lock, ...................... | 1 |  | PAOZZN |
|  | MS35338-79 | 81349 | - WASHER, Split Lock $\qquad$ No. 6 (AP) | 1 |  | PAOZZN |
|  | MS15795-305 | 81349 | - WASHER, Flat, No. 6 (AP) | 1 |  | PAOZZN |
|  | AN507C-632-6 | 81349 | - . SCREW, Machine $\qquad$ <br> Pan Head, 6-32 x 3/8 (AP) | 1 |  | PAOZZN |
| /1-6 | $\begin{aligned} & \text { 3186BA522U0 } \\ & \text { 40AM } \end{aligned}$ | 30983 | - CAPACITOR, C4, C7 Electrolytic, 5200 uf, 40 WVDC, $-10+75 \%$ | 2 |  | PAOZZN |
| /1-7 | MS35649-48 | 81349 | - . NUT, Hex, 6-32 (AP) ...................... | $2$ |  | PAOZZN |
|  | MS35338-79 | 81349 | - WASHER, Split Lock <br> No. 6 (AP) | $2$ |  | PAOZZN |
|  | MS15795-305 <br> AN507C-632-6 | $81349$ | -•WASHER, Flat, No. 6 (AP) $\qquad$ | $2$ |  | PAOZZN |
|  | AN507C-632-6 | $81349$ | -•SCREW, Machine, $\qquad$ <br> Flat Head, 6-32 x 3/8 (AP) | $2$ |  | PAOZZN |
| /1-8 | VR3 | 97244 | -•CLAMP, Capacitor, C4, C7 | 2 |  | PAOZZN |
| /1-9 | MS35649-48 | $81349$ | - • NUT, Hex, 6-32 (AP) | $1$ |  | PAOZZN |
|  | MS35338-79 | $81349$ | - WASHER, Split Lock $\qquad$ No. 6 (AP) | $1$ |  | PAOZZN |
|  | MS15795-305 | 81349 | - WASHER, Flat <br> No. 6 (AP) | 1 |  | PAOZZN |
|  | AN507C-632-6 | 81349 | - - SCREW, Machine $\qquad$ Pan Head, 6-32 x 3/8 (AP) | 1 |  | PAOZZN |
| /1-10 | 08517 | 23386 | - TRANSFORMER, T $\qquad$ Power | 1 |  | PAOZZN |
| /1-11 | MS35649-84 | 81349 | - . NUT,Hex,8-32(AP) .......................... | 4 |  | PAOZZN |
|  | MS35338-80 | 81349 | - WASHER, Split Lock $\qquad$ No. 8 (AP) | 4 |  | PAOZZN |
|  | MS15795-307 | 81349 | - WASHER, Flat. <br> No. 8 (AP) | 4 |  | PAOZZN |
| /1-12 | MS35234-10 | 96906 | - - SCREW, Machine, Flat Head, $\qquad$ $8-32 \times 1 / 2$ (AP) | 4 |  | PAOZZN |

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FIGURE 7-12. POWER SUPPLY ASSEMBLY, MODULE A10 (Cont.)


FIGURE 7-12. POWER SUPPLY ASSEMBLY, MODULE A10 (Cont

| FIGURE \& INDEX NO. | PART NUMBER | FSCM | 1234567 DESCRIPTION | UNITS PER ASSY | USABLE ON CODE | $\begin{array}{\|c} \text { SMR } \\ \text { CODE } \end{array}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & 7-12- \\ & / 1-33 \end{aligned}$ | MS3564944 | 81349 |  |  |  | PAOZZN |
|  | MS35338-78 | 81349 81349 | -• WASHER, Split Lock, <br> No. 4 (AP) | 1 |  | PAOZZN |
|  | MS15795-303 | 96906 | - . WASHER, Flat, No. 4 (AP) .............. | 1 |  | PAOZZN |
|  | MS35233-14 | 96906 | ... SCREW, Machine <br> Pan Head $4-40 \times 5 / 16$ (AP) | 1 |  | PAOZZN |
| /1-34 | 08592 | 23386 | ... PRINTED WIRING BOARD, A1OA2 | 1 |  | XA |
| /1-36 | MS35649-48 | 81349 | -• NUT, Hex, 6-32 (AP) ...................... | 2 |  | PAOZZN |
|  | MS35338-79 | 81349 | -• WASHER, Split Lock, $\qquad$ <br> No. 6 (AP) | 2 |  | PAOZZN |
|  | MS15795-305 | 81349 | - W WASHER, Flat, No. 6 (AP) ................. | 2 |  | PAOZZN |
| /1-37 | 1488-6 | 83330 | - LUG, Solder, No. 6 ......................... | 1 |  | PAOZZN |
| /1-40 | 5608-10 | 86928 | $\cdots$ WASHER, Shoulder insulating............ | 2 |  | PAFZZN |
| /1-43 | MS35649-48 | 81349 | - NUT, Hex, 6-32 (AP) ........................ | 2 |  | PAOZZN |
|  | MS35338-79 | 81349 | -• WASHER, Split Lock, $\qquad$ No. 6 (AP) | 2 |  | PAOZZN |
|  | MS15795-305 | 81349 | -. WASHER, Flat, No. 6 (AP) ................ | 2 |  | PAOZZN |
| /1-44 | 1488-6 | 83330 | - . LUG, Solder, No. 6 (AP) ........... | 1 |  | PAOZZN |
|  | MS35233-30 | 96906 | - . SCREW, Machine $\qquad$ Pan Head, 6-32 $\times 1 / 2$ (AP) | 2 |  | PAOZZN |
| /1-45 | 6015B | 13103 | - HEAT, Sink, Regulator | 1 |  | PAOZZN1 |
| /1-46 | 1488-4 | 83330 | -•LUG, Solder, No. 4 ............ | 2 |  | PAOZZN |
| /1-47 | MS3564944 | 81349 | - NUT, Hex, 4-40 (AP) .......... | 1 |  | PAOZZN |
|  | MS35338-78 | 81349 | - WASHER, Split Lock, $\qquad$ <br> No. 4 (AP) | 1 |  | PAOZZN |
|  | MS35233-14 | 96906 | - . SCREW, Machine, Pan Head ............. 4-40 x 5/16 (AP) | 1 |  | PAOZZN |
| /1-48 | LM7915C | 27014 | - - INTEGRATED CIRCUIT, <br> U3, - 15 volt regulator | 1 |  | PAOZZN |
| /1-49 | MS3564948 | 81349 | -• NUT, Hex, 6-32 (AP) ........ | 2 |  | PAOZZN |
|  | MS35338-79 | 81349 | - . WASHER, Split Lock $\qquad$ <br> No. 6 (AP) | 2 |  | PAOZZN |
|  | MS15795-305 | 81349 | - WASHER, Flat, No. 6 (AP) | 2 |  | PAOZZN |
|  | 1488-6 | 83330 | -•LUG, Solder, No. 6 | $1$ |  | PAOZZN |
| /1-51 | 5210-2 MS35233-30 | 86928 96906 | -•INSULATOR, Regulator $\qquad$ <br> SCREW Machin Pan Head | $\begin{aligned} & 1 \\ & 2 \end{aligned}$ |  | PAOZZN |
|  | MS35233-30 | 96906 | -• SCREW, Machine, Pan Head $\qquad$ $6-32 \times 1 / 2(\mathrm{AP})$ | 2 |  | PAOZZN |
| /1-52 | A22-2003 | K1935 | - . - COVER, Regulator ........................ | 1 |  | PAOZZN |
| /1-53 | 5608-10 | 86928 | $\cdots$ WASHER, Shoulder insulating ........... | 2 |  | PAOZZN |
| /1-54 | 14884 | 83330 | - LUG, Solder, No. 4 ......................... | 1 |  | PAOZZN |
| /1-55 | MS3564944 MS35338-78 | 81349 81349 | • ${ }^{\text {a }}$ NUT, Hex, 4-40(AP) ....................... | 1 |  | PAOZZN |
|  | MS35338-78 | 81349 | -•WASHER, Split Lock, $\qquad$ <br> No. 4 (AP) | 1 |  | PAOZZN |
|  | MS35233-14 | 96906 | -•SCREW, Machine, <br> Pan Head, $4-40 \times 5 / 16$ (AP) | 1 |  | PAOZZN |
| /1-56 | T310A105M035AS | 31433 | - CAPACITOR,C2,C3, C5, C6, C8, Tantalum, 1 uf, 35 WVDC, $\pm 20 \%$ | 5 |  | PAOZZN |

FIGURE 7-12. POWER SUPPLY ASSEMBLY, MODULE, A10 (Cont)



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Figure 7-12. Power Supply Assembly, Module, A10 (Sheet 2 of 2)

FIGURE 7-12. POWER SUPPLY ASSEMBLY, MODULE A10 (Cont.)


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FIGURE 7-12. POWER SUPPLY ASSEMBLY, MODULE, A10 (Cont.)


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Figure 7-13. Circuit Card Assembly, LCD Lampboard

FIGURE 7-13, CIRCUIT CARD ASSEMBLY, LCD LAMP BOARD



Figure 7-13.1 Circuit Card Assembly, LCD/LED Board

FIGURE 7-13.1. CIRCUIT CARD ASSEMBLY, LCD-LED BOARD


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| PART NUMBER | FIGURE AND INDEX NO. | QTY PER END ITEM | PART NUMBER | FIGURE AND INDEX NO. | QTY PER END ITEM |
| :---: | :---: | :---: | :---: | :---: | :---: |
| AD201A | 7-9-68 | 1 | CMD5CD150G03 | 7-4-53 | 1 |
| AL518JH883B | 7-9-51 | 1 | CMRO5E270GODR | 7-9-45 | 1 |
| ALV7524JN | 7-5-69 | 1 | CMRO5E300GODR | 7-3-67 | 1 |
| AH5016C-5 | 7-7-15 | 1 | CMRO5E330GODR | 7-3-10 | 2 |
| AM26LS30PC | 7-7-10 | 1 | CMRO5E430GODR | 7-2-14 | 2 |
| AM26LS32CN | 7-7-9 | 1 | CMRO5E470GODR | 7-9-60 | 2 |
| AN507C-632-6 | 7-12-3 | 7 | CMRO5E560GODR | 7-10-24 | 1 |
| A1004AL | 7-3-41 | 2 | CMRO5E680GODR | 7-3-31 | 6 |
| A22-2003 | 7-12-39 | 2 | CMRO5E750GODR | 7-2-9 | 1 |
| A23-2023Z | 7-8-11 | 3 | CMRO5E820GODR | 7-2-15 | 7 |
| BSF-1BBGP102M | 7-1-129 | 9 | CMRO5FIO1GODR | 7-3-24 | 11 |
| B52600F003 | 7-10-61 | 1 | CMRO5F111GODR | 7-2-4 | 2 |
| CA-11-STL-T3WW | 7-11-26 | 4 | CMRO5F121GODR | 7-2-12 | 1 |
| CA-12-STL-T3WW | 7-11-44 | 2 | CMRO5F151GODR | 7-2-7 | 5 |
| CA-14-STL-T3WW | 7-11-45 | 2 | CMRO5F241GODR | 7-3-68 | 1 |
| CA-20-STL-T3WW | 7-11-27 | 2 | CMFO5F271GODR | 7-3-64 | 1 |
| CA1458E | 7-5-64 | 2 | CMRO5F301GODR | 7-7-11 | 6 |
| CA3046E | 7-4-14 | 3 | CMRO5F331GODR | 7-4-2 | 2 |
| CA3140BT/3 | 7-9-74 | 1 | CMRO5F910GODR | 7-2-11 | 2 |
| CA324E | 7-5-29 | 6 | CMRO6F152GODR | 7-5-22 | 4 |
| CCR06CG222GR | 7-6-34 | 1 | CMRO6F222GODR | 7-9-72 | 1 |
| CCR06CG472GR | 7-6-35 | 1 | CMRO6F332GODR | 7-5-59 | 1 |
| CD22100F | 7-5-76 | 1 | CMRO6F472GODR | 7-9-67 | 1 |
| CD40N28BCN | 7-5-44 | 2 | CM96FD821JN3 | 7-10-4 | 2 |
| CD40107BEX | 7-11-55 | 3 | COM6402P | 7-7-16 | 1 |
| CD40108BEX | 7-9-33 | 1 | CS13BE156M | 7-9-53 | 1 |
| CD40109BEX | 7-5-57 | 3 | CS13BE335M | 7-10-77 | 1 |
| CD4013BE | 7-5-42 | 7 | CYL1400-250 | 7-1-59 | 1 |
| CD40174BMT | 7-9-4 | 1 | CY15C102M | 7-3-1 | 32 |
| CD4047BEMJ/883B | 7-11-53 | 1 | C280MCH/AIM | 7-9-76 | 1 |
| C4051BEMJ/883B | 7-11-10 | 1 | C280MCH/A100K | 7-10-42 | 2 |
| C4053BEX | 7-5-60 | 1 | C280MCH/A47K | 7-9-49 | 1 |
| CD4054BD | 7-11-31 | 10 | C280MCH/A470K | 7-10-43 | 1 |
| CD4056BD | 7-11-28 | 13 | C280MCH/A68K | 7-9-48 | 2 |
| CD4066BEX | 7-5-39 | 6 | C312C222M1U1CA | 7-9-12 | 1 |
| CD4069UBE | 7-11-21 | 1 | C312C479K2G5CA | 7-9-107 | 1 |
| CD4076BD | 7-11-49 | 4 | C320C103M1U1C1 | 7-3-5 | 98 |
| CD4085BF | 7-11-23 | 1 | C320C152K2R5C1 | 7-5-61 | 1 |
| CD4094BEX | 7-9-2 | 7 | C320C223M1U1CA | 7-5-53 | 2 |
| CD4503BCN | 7-7-17 | 13 | C320C473M5U1CA | 7-9-18 | 3 |
| CD4508BD/3 | 7-11-15 | 1 | DAC-20CQ | 7-9-37 | 1 |
| CD4514BD/3 | 7-11-16 | 2 | DD-391 | 7-5-85 | 2 |
| CD4516BCN | 7-11-58 | 1 | DKV6522B | 7-9-100 | 2 |
| CD4528BF | 7-11-19 | 1 | DM10115N | 7-10-71 | 1 |
| CD4529BCN | 7-7-26 | 1 | DM74LSOONAT | 7-10-64 | 1 |
| CK60BX101K | 7-11-35 | 2 | DM74LS390N | 7-10-10 | 3 |
| CMRO5ROGODR | 7-10-67 | 1 | DS3SD | 7-8-15 | 1 |
| CMRO5C100GODR | 7-2-6 | 3 | DTZ-10 | 7-5-68 | 1 |
| CMRO5C120GODR | 7-10-73 | 1 | DTZ-15 | 7-8-6 | 2 |
| CM05CD180G03 | 7-3-62 | 1 |  |  |  |

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\(\left.$$
\begin{array}{|l|l|l|l|l|l|}\hline & \begin{array}{ll}\text { FIGURE AND } \\
\text { INDEX NO. }\end{array} & \begin{array}{c}\text { QTY PER } \\
\text { END ITEM }\end{array}
$$ \& \& <br>
PART NUMBER \& PART NUMBER \& FIGURE AND <br>

INDEX NO.\end{array}\right]\)| QTY PER |
| :---: |
| END ITEM |

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| PART NUMBER | FIGURE AND INDEX NO. | QTY PER END ITEM | PART NUMBER | FIGURE AND INDEX NO. | QTY PER END ITEM |
| :---: | :---: | :---: | :---: | :---: | :---: |
| MV1634 | 7-10-69 | 1 | RLR07C330GR | 7-6-37 | 3 |
| MV1650 | 7-10-40 | 2 | RLR07C334GR | 7-5-54 | 2 |
| M24308/2-1 | 7-12-68 | 1 | RLR07C390GR | 7-4-25 | 2 |
| M38510/05001BCB | 7-7-20 | 2 | RLR07C564GR | 7-10-26 | 2 |
| M38510/05101BEA | 7-5-56 | 4 | RL07S101G | 7-3-14 | 16 |
| M38510/05202BCB | 7-9-58 | 1 | RL07S102G | 7-3-6 | 63 |
| M38510/05204BCB | 7-11-20 | 1 | RL07S103G | 7-3-8 | 53 |
| M38510/054-01BEA | 7-9-57 | 1 | RL07S104G | 7-4-16 | 18 |
| M38510/05701BCB | 7-9-7 | 2 | RL07S113G | 7-11-34 | 2 |
| M38510/06203BEA | 7-7-24 | 1 | RL07S122\& | 7-3-49 | 5 |
| M38510/10201BEA | 7-9-82 | 2 | RL07S123G | 7-5-26 | 6 |
| M38510/1070\B | 7-9-22 | 1 | RL07S124G | 7-8-35 | 2 |
| M38510/10708BEA | 7-5-88 | 2 | RL07S151G | 7-9-21 | 2 |
| M38510/10708Bl-C | 7-12-42 | 1 | RL07S152G | 7-4-32 | 12 |
| M38510/11201BCB | 7-5-71 | 3 | RL07S153G | 7-5-31 | 15 |
| M38510/17001BCD | 7-11-22 | 1 | RL07S154G | 7-9-17 | 1 |
| M38510/30001BCB | 7-10-60 | 1 | RL07S181G | 7-11-40 | 2 |
| M38510/30005BCD | 7-8-5 | 1 | RL07S182\& | 7-10-19 | 3 |
| M38510/3003BCB | 7-7-22 | 2 | RL07S183G | 7-5-35 | 5 |
| M38510/30102BCB | 7-9-31 | 4 | RL07S203G | 7-11-52 | 1 |
| M38510/30106BEB | 7-9-30 | 3 | RL07S220G | 7-6-33 | 4 |
| M38510/30107BEB | 7-7-21 | 1 | RL07S221G | 7-3-7 | 12 |
| M38510/30301BCB | 7-8-13 | 3 | RL07S222G | 7-3-50 | 18 |
| M38510/30702BEB | 7-8-8 | 1 | RL07S223G | 7-4-17 | 26 |
| M38510/30901BEA | 7-10-50 | 2 | RL07S270G | 7-4-27 | 1 |
| M38510/31004BCB | 7-8-9 | 2 | RL07S271G | 7-8-27 | 1 |
| M38510/31501BCB | 7-10-51 | 1 | RL07S272G | 7-9-71 | 1 |
| M38510/31503BEA | 7-10-14 | 5 | RL07S273G | 7-5-30 | 2 |
| M38510/31505BEA | 7-9-61 | 2 | RL07S331G | 7-4-36 | 8 |
| M38510/31506BEA | 7-9-87 | 1 | RL07S332G | 7-5-49 | 22 |
| M83723-02R-1626N | 7-7-2 | 1 | RL07S333G | 7-6-36 |  |
| NS-432-100 | 7-5-5 | 14 | RL07S391G | 7-6-32 | 2 |
| NS-441-B1 | 7-5-4 | 14 | RL07S392G | 7-9-43 | 6 |
| N82S83N | 7-9-6 | 1 | RL07S393G | 7-4-40 | 5 |
| N8 8026 | 7-9-105 | 2 | RL07S470G | 7-3-56 | 21 |
| P5101L | 7-8-16 | 3 | RL07S471G | 7-5-46 | 25 |
| R-2174(P)/URR | Reference | 1 | RL07S4726 | 7-4-18 | 24 |
| R-2174A(P)/URR | Reference | 1 | RL07S473G | 7-5-28 | 17 |
| RB-67-1SK-7 | 7-1-68 | 2 | RL07S560G | 7-4-44 | 2 |
| RCR20G150JS | 7-3-35 | 1 | RL07S561G | 7-4-13 | 1 |
| RCR20G181JS | 7-13.1/1-2 | 6 | RL07S562G | 7-3-40 | 1 |
| RWR50FP201 | 7-9-20 | 1 | RL07S563G | 7-6-27 | 2 |
| RJ50FW103 | 7-5-37 | 1 | RL07S680G | 7-9-98 | 4 |
| RJ50FW-202 | 7-5-19 | 2 | RL07S681G | 7-5-25 | 13 |
| RJ50FW501 | 7-4-43 | 2 | RL07S6820 | 7-5-27 | 5 |
| RJ50FW502 | 7-3-55 | 2 | RL07S683G | 7-5-65 | 2 |
| RJ50FW503 | 7-5-23 | 2 | RL07S820G | 7-9-89 | 1 |
| RLRO7C100GR | 7-3-9 | 17 | RL07S822G | 7-9-35 | 9 |
| RLRO7C105GR | 7-5-52 | 4 | RL07S823G | 7-8-36 | 1 |
| RLRO7C184GR | 7-5-34 | 1 | RN55D1400F | 7-5-84 | 3 |
| RLRO7C224GR | 7-4-38 | 1 | SIPL8-103 | 7-11-2 | 2 |

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|  | FIGURE AND <br> INDEX NO. | QTY PER <br> END ITEM |  | FIGURE AND |
| :--- | :--- | :--- | :--- | :--- | :--- |
| PART NUMBER | PART NUMBER | QTY PER <br> INDEX NO. |  |  |
| SIP8-223 ITEM |  |  |  |  |

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|  | FIGURE AND <br> INDEX NO. | QTY PER <br> END ITEM |  | FIGURE AND | QTY PER |
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| PART NUMBER | PART NUMBER | END ITEM |  |  |  |

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| PART NUMBER | FIGURE AND INDEX NO. | QTY PER END ITEM | PART NUMBER | FIGURE AND INDEX NO. | QTY PER END ITEM |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 27VX-10-B07A-5B | 7-1-85 | 2 |  |  |  |
| 30-350 | 7-1-70 | 2 |  |  |  |
| 3050HJ122U050JM | 7-12-13 | 1 |  |  |  |
| 3071FE221T016SF | 7-5-87 | 5 |  |  |  |
| 3100-25 | 7-1-61 | 1 |  |  |  |
| 3186BA522U040AM | 7-12-6 | 2 |  |  |  |
| 3186EA1234025AMA2 | 7-12-2 | 1 |  |  |  |
| 3402-0000T | 7-11-6 | 1 |  |  |  |
| 3414-6034 | 7-11-7 | 1 |  |  |  |
| 3428-2002 | 7-11-24 | 1 |  |  |  |
| 3429-1002 | 7-10-15 | 1 |  |  |  |
| 3429-2002 | 7-11-47 | 2 |  |  |  |
| 3429-2003 | 7-5-83 | 1 |  |  |  |
| 3431-2002 | 7-8-39 | 1 |  |  |  |
| 3431-2003 | 7-5-80 | 1 |  |  |  |
| 3432-2002 | 7-11-46 | 1 |  |  |  |
| 3432-2003 | 7-5-72 | 1 |  |  |  |
| 3483-1000 | 7-1-107 | 1 |  |  |  |
| 3492-1002 | 7-9-55 | 1 |  |  |  |
| 350244 | 7-12-20 | 1 |  |  |  |
| 37039 | 7-10-68 | 1 |  |  |  |
| 4N28 | 7-10-30 | 1 |  |  |  |
| 4310R-101-103 | 7-11-3 | 3 |  |  |  |
| 4310R-101-223 | 7-7-25 | 1 |  |  |  |
| 4459-M07-F09 | 7-1-94 | 4 |  |  |  |
| 4586-48 | 7-12-4 | 1 |  |  |  |
| 5210-2 | 7-12-38 | 2 |  |  |  |
| 5608-10 | 7-12-40 | 4 |  |  |  |
| 5610-280-032 | 7-1-56 | 4 1 |  |  |  |
| 5710-54-25 | 7-1-59 | 2 |  |  |  |
| 5710-94-015 | 7-1-63 | 2 |  |  |  |
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## CHAPTER 8. CIRCUIT DIAGRAMS

## 8-1. INTRODUCTION

This chapter contains circuit diagrams and Interconnection diagrams applicable to the R-2174(P)/URR Radio Receiver. A listing of all diagrams, showing Figure number and Page number for each, is shown for quick reference to any diagram.

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Figure 5-2. Overall, Simplified Functional, Block Diagram



Figure 6-2. Typical Signal Levels R-2174 (P)/URR HF Radio Receiver


Figure 6-8B. Main IF/AF A4 Printed Circuit
6-29/6-30

NOTE: Refer to Parts List for differences in the $\mathrm{R}-2174 \mathrm{~A}$ Receiver.


Figure 6-12B. First LO Synthesizer A7 Printed Circuit
CHANGE 2 6-41/6-42


Figure 6-14B. Receiver Control A9 Printed
Circuit
6-47/6-48


Figure 8-1. Overall, Simplified Functional, Block Diagram

## NOTES:

## (UNLESS OTHEPWISE NOTED)

1. RESISTOR VALUES ARE IN OHMS $1 / 4$

WATT $K=1,000 \mathrm{M}=1,000,000$
2. CAPACITOR 'VALUES ONE OR GREATER

ARE IN PICOFFARADS, LESS THAN ONE
ARE IN MICROFARADS
3. INDUCTANCE VALUES ONE OR GREATER

ONE ARE IN MILLIHENRIES
4. PARTIAL REFERENCE DESIGNATIONS

Shown; FOR COMPLETE DESIGNATION
PREFIX WITH UNIT AND OR ASSEMBLY
BESIGN.


Figure 8-2. Schematic Diagram, RF Low
Pass Filter, A1


Figure 8-3. Schematic Diagram, 1st Mixer,


8-9/8-10
LESS THAN ONE ARE ENE MILLE GRENRTER ARE IN MICROHENRIES
4.
PARTIAL REFERENCE DESIGNATIONS ARE SHOWN: FOR COMPLETE
4. PARTAA REFERENE DESIGNTONR RAE SHOWN: FOR COMPLETE

6. FOR ISB OPRRATION CONNECTLINKI TO ISE.

| 1.c. no. | device | GNo | +15V(8) | +15V(C) | -15V |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 47 | 324 | II |  | 4 |  |
| U14 | 324 | 11 |  | 4 |  |
| U17 | 324 | 11 |  | 4 |  |
| U9 | 4013 | 7 |  | 14 |  |
| UII | 4066 | 7 |  | 14 |  |
| U12 | 4066 | 7 |  | 14 |  |
| U19 | 4053 | 6,7,8 |  | 16 |  |
| 422 | 1458 |  |  | 8 | 4 |
| 424 | 339 | 12 |  | 3 |  |
| 028 | 1458 |  | 8 |  | 4 |
| 429 | 1458 | 4 | 8 |  |  |



Figure 8-5. Schematic Diagram, Main IF/AF, A4 (Sheet 2 of 6)


Figure 8-5. Schematic Diagram, Main IF/AF, A4 (Sheet 3 of 6)


Figure 8-5. Schematic Diagram, Main IF/AF, A4 (Sheet 4 of 6)



Figure 8-5. Schematic Diagram, Main IF/AF, A4 (Sheet 6 of 6 )
8-21/8-22


Figure 8-6. Schematic Diagram, ISB, A5 (Optional) (Sheet 1 of 2)



Figure 8-7. Schematic Diagram, Microcomputer Assembly, A6


Figure 8-8. Schematic Diagram, Serial Asynchronous
Interface, A6A1 (Optional) (Sheet 1 of 2)
8-29/8-30


Figure 8-8. Schematic Diagram, Serial Asynchronous Interface, A6A1 (Optional) (Sheet 2 of 2)


Figure 8-9. Schematic Diagram, Microcomputer, A6A2 (Sheet 1 of 2)


Figure 8-9. Schematic Diagram, Microcomputer, A6A2 (Sheet 2 of 2)

Hotes: (UNLESS OTHERWISE NOTED)

1. Resistor values are in ohms; $\mathrm{K}=1,000 ;$ m $=1,000,000$.
2. CAPACITOR VALUES ONE OR GREATER ARE IN PICOFARADS
3. Linouctance values one or greater are in microhenries,
4. LNOCTANCE VALUES ONE OR GLEATER A

- partial reference designations are shown ; for COMPLETE DESIGN
ASSEMELY DESIGN




Figure 8-10. Schematic Diagram, 1st LO Synthesizer, A7 (Sheet 2 of 5)



PATENTED DEVICE
MANUFACTURED UNDER U.S. PATENT NOS. 4,204,174
NDER U.S. PA
and $3,555,446$
Figure 8-10. Schematic Diagram, 1st LO



. RESISTOR VALUES AAE WO OHMS $1 / 4$ CAPACITOR VALUES ONE OR GRE ATER
ARE IN PCOFARASO, LESS THAN ONE
 3. LNDECTANE VALUES ONE OR GREATE CNE ARE IN MILLIHENRIES



Figure 8-11. Schematic Diagram
2nd LO/BFO Synthesizer, A8


Figure 8-11. Schematic Diagram
2nd LO/BFO Synthesizer, A8
Sheet 2 of 3)
8-49/8-50


Figure 8-11. Schematic Diagram
2nd LO/BFO Synthesizer, A8
Sheet 3 of 3)
8-51/8-52


Figure 8-11.1 Schematic Diagram
$2^{\text {nd }}$ LO/BFO Synthesizer, A8
CHANGE 28 -52.1/8-52.2


Figure 8-11.1 Schematic Diagram
$2^{\text {nd }}$ LO/BFO Synthesizer, A8
(Sheet 2 of 3)
CHANGE $28-52.3 / 8-52.4$


Figure 8-11.1 Schematic Diagram,
$2^{\text {nd }}$ LO/BFO Synthesizer,
A8 (Sheet 3 of 3)
CHANGE 2 8-52.5/8-52.6


NOTES: UNLESS OTHERWISE NOTED


 3. INDUCTANCE NALLES OHE O
4. ONE ARE IN MLLLIIENELIES

5. DESSIGN USED-UEG, C24, R32, O2, E8, Js, WI, OS4

| $A$ | $E$ | $A 3$ |
| :---: | :---: | :---: |
| ST. | SH. | SSH |
| REVISON STATUS OF SHEETS |  |  |



Figure 8-12. Schematic Diagram, Receiver
Control, A9 (Sheet 2 of 3)
8-55/8-56


Figure 8-12. Schematic Diagram, Receive

## NOTES:

(UNLESS OTHERWISE NOTED)

1. RESISTORVALUES ARE IN OHMS I/4 WATT $K=1,000 \mathrm{M}=1,000,000$
2. CAPACITOR VALUES ONE OR GREATER ARE IN PICOFARADS, LESS THAN ONE ARE IN MICROFARADS
3. INDUCTANCE VALUES ONE OR GREATER ARE IN MICROHENRIES, LESS THAN ONE ARE IN MILLIHENRIES
4. PARTIAL REFERENCE DESIGNATIONS

ARE SHOWN; FOR COMPLETE DESIGNATION, PREFIX WITH UNIT AND OR ASSEMBLY DESIGN.


Figure 8-13. Schematic Diagram, Liquid
Crystal Display Lamps
8-59/8-60
T.O. 31R2-2URR-251


Figure 8-13.1 Schematic Diagram,
Liquid Crystal Display LED
CHANGE 2 8-60.1/8-60.2


Figure 8-14. Schematic Diagram,
Power Supply, A10
8-61/8-62



Figure 8-15. Schematic Diagram
Interconnection Diagram


# The Metric System and Equivalents 

## Lincer Monourn

1 centimeter $=10$ millimeters $=.39$ inch
1 decimeter $=10$ centimeters $=3.94$ inches
1 meter $=10$ decimeters $=39.37$ inches
1 dekameter $=10$ meters $=32.8$ feet
1 hectometer $=10$ dekameters $=328.08$ feet
1 kilometer $=10$ hectometers $=3,280.8$ feet

## Weighte

1 centigram $=10$ milligrams $=.15$ grain
1 decigram $=10$ centigrams $=1.54$ grains
$1 \mathrm{gram}=10$ decigram $=.035$ ounce
1 dekagram $=10$ grams $=.35$ ounce
1 hectogram = 10 dekagrams $=3.52$ ounces
1 kilogram $=10$ hectograms $=2.2$ pounds
1 quintal $=100$ kilograms $=220.46$ pounds
1 metric ton $=10$ quintals $=1.1$ short tons

1 centiliter $=10$ milliters $=.34$ fl. ounce
1 deciliter $=10$ centiliters $=3.38$ fl. ounces
1 liter $=10$ deciliters $=33.81$ fl. ounces
1 dekaliter $=10$ liters $=2.64$ gallons
1 hectoliter $=10$ dekaliters $=26.42$ gallons
1 kiloliter $=10$ hectoliters $=264.18$ gallons

## Equare Mcecure

1 sq . centimeter $=100 \mathrm{sq}$. millimeters $=.155 \mathrm{sq}$. inch
1 sq . decimeter $=100$ sq. centimeters $=15.5$ sq. inches
1 sq . meter (centare) $=100 \mathrm{sq}$. decimeters $=10.76$ sq. feet
1 sq. dekameter (are) $=100 \mathrm{sq}$. meters $=1,076.4$ sq. feet
1 sq . hectometer (hectare) $=100 \mathrm{sq}$. dekameters $=2.47 \mathrm{acres}$
1 sq. kilometer $=100$ sq. hectometers $=.386$ sq. mile

Cubic Moenure
1 cu. centimeter $=1000 \mathrm{cu}$. millimeters $=.06 \mathrm{cu}$. inch 1 cu . decimeter $=1000 \mathrm{cu}$. centimeters $=61.02 \mathrm{cu}$. inches 1 cu. meter $=1000 \mathrm{cu}$. decimeters $=35.31 \mathrm{cu}$. feet

## Approximate Conversion Factors

| To chenge | To | Muthiply by | To change | To | Multiply by |
| :---: | :---: | :---: | :---: | :---: | :---: |
| inches | centimeters | 2.540 | ounce-inches | newton-meters | . 007062 |
| feet | meters | . 305 | centimeters | inches | . 394 |
| yards | meters | . 914 | meters | feet | 3.280 |
| miles | kilometers | 1.609 | meters | yards | 1.094 |
| square inches | square centimeters | 6.451 | kilometers | miles | . 621 |
| square feet | square meters | . 093 | square centimeters | square inches | . 155 |
| square yards | square meters | . 836 | square meters | square feet | 10.764 |
| square miles | square kilometers | 2.590 | square meters | square yards | 1.196 |
| acres | square hectometers | . 405 | square kilometers | square miles | . 386 |
| cubic feet | cubic meters | . 028 | square hectometers | acres | 2.471 |
| cubic yards | cubic meters | . 765 | cubic meters | cubic feet | 35.315 |
| fluid ounces | milliliters | 29,573 | cubic meters | cubic yards | 1.308 |
| pints | liters | . 473 | milliliters | fluid ounces | . 034 |
| quarts | liters | . 946 | liters | pints | 2.113 |
| gallons | liters | 3.785 | liters | quarts | 1.057 |
| ounces | grams | 28.349 | liters | gallons | . 264 |
| pounds | kilograms | . 454 | grams | ounces | . 035 |
| short tons | metric tons | . 907 | kilograms | pounds | 2.205 |
| pound-feet | newton-meters | 1.356 | metric tons | short tons | 1.102 |
| pound-inches | newton-meters | . 11296 |  |  |  |

Temperature (Exact)

| ${ }^{\circ} \mathrm{F}$ | Fahrenheit <br> temperature | 5/9 (after <br> subtracting 32) | Celsius <br> temperature | ${ }^{\circ} \mathrm{C}$ |
| :--- | :--- | :--- | :--- | :--- |

PIN: 048248-000

